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# Aural servo: towards an alternative approach to sound localization for robot motion control

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**THÈSE / UNIVERSITÉ DE RENNES 1**  
*sous le sceau de l'Université Bretagne Loire*

pour le grade de

**DOCTEUR DE L'UNIVERSITÉ DE RENNES 1**

*Mention : Traitement du signal et telecommunications*

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présentée par

**Aly Magassouba**

préparée à l'unité de recherche IRISA – UMR6074  
Institut de Recherche en Informatique et Système Aléatoires  
ISTIC

**Aural servo:  
towards an alternative  
approach  
to sound localization  
for robot motion control**

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Eventually I dedicate this thesis to my family which gave me the strength to come this far since I left my homeland. Now I can proudly look ahead and move forward....





# Synthèse

*Est-ce que les robots doivent nécessairement localiser les sources sonores afin d'interagir avec leur environnement ?*

Cette question a été le fil central de cette thèse. En effet, dans un contexte d'interactions homme-robot, un robot doté d'un sens de l'ouïe fiable est essentiel. L'information auditive en robotique est généralement exploitée par des capacités tels que la localisation de source, la séparation de sources, le traitement de la parole et l'analyse de scène acoustique, qui définissent le sujet de *l'audition robotique*. Dans cette thématique, une attention particulière a été accordée à la localisation de source, principalement parce qu'elle est la première étape de l'interaction auditive.

Les techniques de localisation de source, en robotique, s'inspirent des capacités auditives des mammifères et plus particulièrement de celles des humains. L'approche typique de la localisation de source en robotique consiste à transposer les connaissances acquises en psychoacoustique et en physiologie dans un système d'audition artificiel dont sont extraits des indices sonores. Ces méthodes supposent généralement que les capteurs et les sources d'intérêt sont statiques. Les indices sonores tels que la différence interaurale en temps et la différence interaurale en amplitude ou les densités spectrales, fournissent des informations sur la direction d'arrivée du son dans le plan horizontal et le plan vertical. L'efficacité du processus de localisation dépend uniquement de la précision des indices sonores extraits.

Cependant, imiter le système auditif humain se trouve être relativement complexe en environnement réel. D'une part, la perception et plus particulièrement les indices sonores sont influencés par la forme du corps (tête, pavillon auditif, torse ...) et d'autre part par les conditions acoustiques (acoustique de la salle, bruit, réverbération). Ces indices sonores varient d'un individu à un autre et d'un environnement à un autre. Par conséquent, même les systèmes artificiels les plus aboutis ne peuvent être déployés que dans des environnements contrôlés. En environnement réel, qui implique la présence de réverbération, de bruit et de mouvement, la précision des indices sonores se dégrade considérablement, ce qui entraîne des erreurs de localisation.

C'est la raison pour laquelle une nouvelle ligne d'approches qui étudie en profondeur les capacités du système auditif humain émerge : se baser uniquement sur la pertinence et la précision des indices sonores ne suffit pas pour une localisation robuste. Plusieurs analyses des capacités de localisation chez l'humain, telles que celles proposées dans [Pop08] ou [CLH97], montrent une grande variabilité de précision lorsqu'il s'agit d'estimer la direction d'arrivée du son. Dans ces expériences, le processus de localisation démontre une bonne précision pour les stimuli sonores

---

situés en face du sujet alors que les positions plus latérales entraînent des erreurs de localisation plus importantes. D'autres études [ALO90, CTS68] ont rapporté que l'estimation de la distance a tendance à être largement sous-estimée. Néanmoins, malgré ces imprécisions, des actions telles que tourner la tête en direction de la source sonore sont exécutées avec précision et sans effort particulier. Ces résultats tendent à démontrer que la précision des indices sonores n'est pas cruciale pour les interactions basées sur l'ouïe. Cette hypothèse est également corroborée par une étude de [HCM83], qui a montré que des sujets souffrant de perte auditive ont développé des stratégies de localisation basée sur des indices sonores erronés, en utilisant le mouvement de leur tête et leur sens visuel. Leur système auditif s'est adapté à l'utilisation d'indices sonores incorrects.

À partir de ces résultats, on peut alors suggérer que les points-clés du système auditif humain ne sont pas la précision des indices sonores, mais plutôt la capacité d'adaptation et la flexibilité par rapport aux conditions acoustiques comme le montre [HVRVO98]. Un système auditif efficace peut être défini comme un mécanisme qui combine différentes modalités telles que la vision (via des références spatiales), le mouvement et la mémoire des expériences auditives précédentes. L'espace acoustique est ensuite appréhendé dynamiquement et les erreurs potentielles des indices sonores peuvent être compensées par ces modalités. Différentes approches peuvent alors être considérées comme la localisation combinant l'ouïe et la vision, la localisation par apprentissage, ou par le biais de mouvement.

Cette thèse se concentre plus particulièrement sur les stratégies de mouvement liées à la perception auditive, dans un contexte binaural. Ces stratégies ont été explorées par les méthodes d'*audition active*. L'*audition active* consiste à mesurer les indices sonores en générant un mouvement prédéfini du robot afin d'améliorer leur précision. L'étape de localisation est effectuée ensuite par la fusion des mesures prises à partir des différentes positions du robot. Cette approche améliore le processus de localisation, mais fait face à certaines limites. D'abord, les mouvements prédéfinis peuvent limiter ou modifier l'interaction avec un comportement non-naturel du robot. Ensuite, les scènes dynamiques, où le robot et/ou la source sont en mouvement nécessitent des modélisations supplémentaires (généralement la dynamique de la source) et un tracking de sorte que la fusion des mesures reste cohérente. Enfin, plus important encore, l'audition active repose toujours sur la localisation et la précision des indices sonores. Sachant que le robot lui-même et les conditions acoustiques influent sur les indices sonores à chaque position du robot dans la scène, cette approche est limitée par une bonne modélisation des conditions acoustiques, de façon similaire aux techniques de localisation en configuration statique.

En réponse à la question initiale sur la nécessité de la localisation, le travail développé dans cette thèse diffère des approches citées auparavant en définissant une interaction auditive comme une tâche modélisée par une commande référencée capteurs. Cette méthode, que nous appelons asservissement sonore, ne localise pas la source sonore. À la place, on définit une tâche par des conditions de mesures à satisfaire, qui correspondent généralement à une position particulière du robot. Par exemple, une tâche en asservissement sonore consiste à positionner le robot afin de satisfaire des conditions définies par un ensemble de mesures auditives, alors

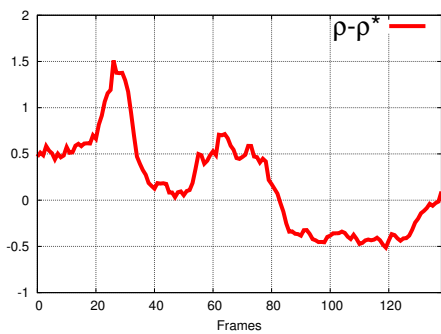
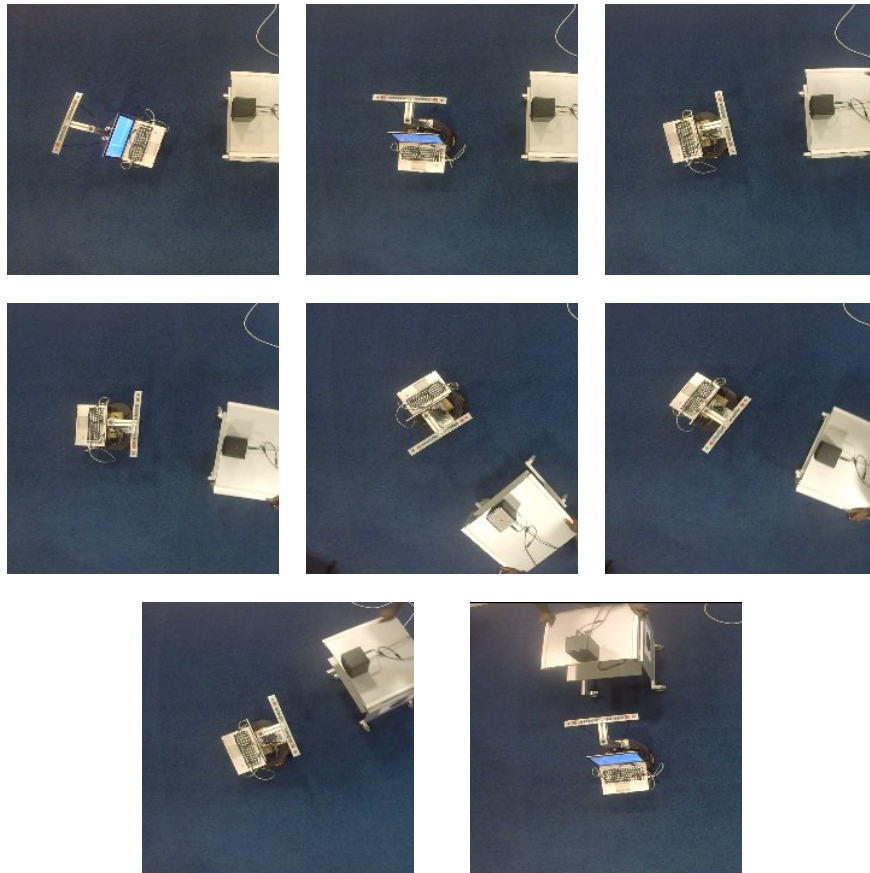
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qu’une approche basée sur la localisation nécessite d’extraire les angles d’azimut et d’élévation et la distance par rapport à la source sonore, avant de déplacer le robot à l’emplacement désiré. Dans l’asservissement sonore, le mouvement du robot est généré en temps réel, à travers une boucle de commande basée sur des indices sonores mesurés à la volée. La commande du robot est obtenue en modélisant la relation entre la dynamique des indices sonores et le mouvement des capteurs auditifs. De cette façon, notre approche repose principalement sur la dynamique des indices sonores et non sur leur valeur intrinsèque, ce qui permet une meilleure tolérance vis-à-vis des conditions acoustiques. Avoir des indices sonores précis n’est alors plus requis. En conséquence, l’étape de localisation est ignorée et la modélisation des conditions acoustiques ou des perturbations de la tête ne sont pas nécessaires. De la même façon, il est possible d’effectuer des tâches basées sur des indices sonores erronés du moment que la dynamique de ces indices reste cohérente. Cette approche appliquée au domaine de l’audition robotique est la contribution essentielle de cette thèse. Le manuscrit est donc centré sur la description de cette approche et comprend la description de la commande référencée capteurs, la modélisation des indices sonores pour la commande et des résultats expérimentaux qui valident la pertinence et l’utilité de cette méthode en environnement réel. Ces expériences sont réalisées dans des configurations dynamiques en considérant des micros en champ libre et des micros intégrés dans des robots humanoïdes.

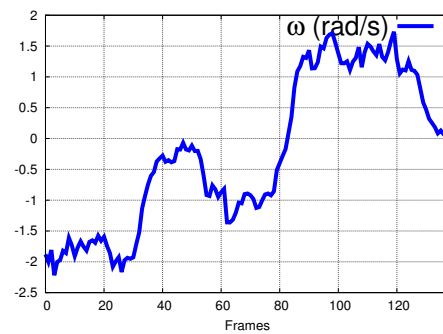
Dans cette thèse, nous avons introduit plusieurs lois de commandes se basant sur la variation des indices sonores par rapport aux mouvements du robot :

- une première basée sur la différence interaurale en temps (ITD),
- une deuxième basée sur la différence interaurale en amplitude (ILD),
- et une dernière utilisant le niveau d’énergie du son.

A partir de ces commandes, nous avons expérimenté différentes tâches de positionnement. Que ce soit avec l’ILD ou avec l’ITD, les tâches qui consistent à orienter le robot vis-à-vis d’une source sonore ont été effectuées dans des conditions réelles (de bruit et de réverbération). De plus, les avantages de ces méthodes ont été soulignés par la résolution des problèmes liés à la perception auditive. Dans le cas de l’ILD, l’ambiguïté avant/arrière est intrinsèquement résolue par le système de contrôle. Dans le cas de l’ITD, nous avons démontré la robustesse du système de contrôle que ce soit en champ lointain ou en champ proche. Finalement, nous avons également prouvé que le positionnement en distance en utilisant le niveau d’énergie peut être effectué avec précision dans des conditions réelles. En outre, ces approches sont adaptées aux scènes dynamiques, qui sont rarement abordées en audition robotique. L’expérience décrite en Figure 1 résume le potentiel de l’asservissement sonore. Dans celle-ci, le robot est capable de faire face à la source sonore alors que celle-ci est initialement présente dans le dos du robot. Par la suite, lorsque la source est déplacée, le robot s’oriente en temps réel vers celle-ci en la maintenant toujours de face. Cette expérience qui s’est déroulée en environnement réel démontre la robustesse de notre approche. Outre le fait de souligner la pertinence de l’asservissement sonore en conditions réelles, ces expériences valident également la polyvalence de l’asservissement



(i) Erreur



(j) Commande

Figure 1: Une expérience illustrant l’asservissement sonore. D’une position quelconque le robot est capable de se retourner vers la source sonore et suivre le mouvement de celle-ci en utilisant deux micros. L’approche utilisée ici est basée sur la différence interaurale en amplitude.

sonore. En effet, ce type d’expérience peut être réalisé, de la même manière, sur des robots humanoïdes où les microphones sont intégrés à la tête. Cette dernière possibilité est illustrée par la Figure 2 où le robot *Romeo* oriente sa tête vers la

source sonore. À notre connaissance, ce travail est l'une des premières tentatives à générer le mouvement d'un robot par rapport à l'information auditive, via une loi de commande explicite.

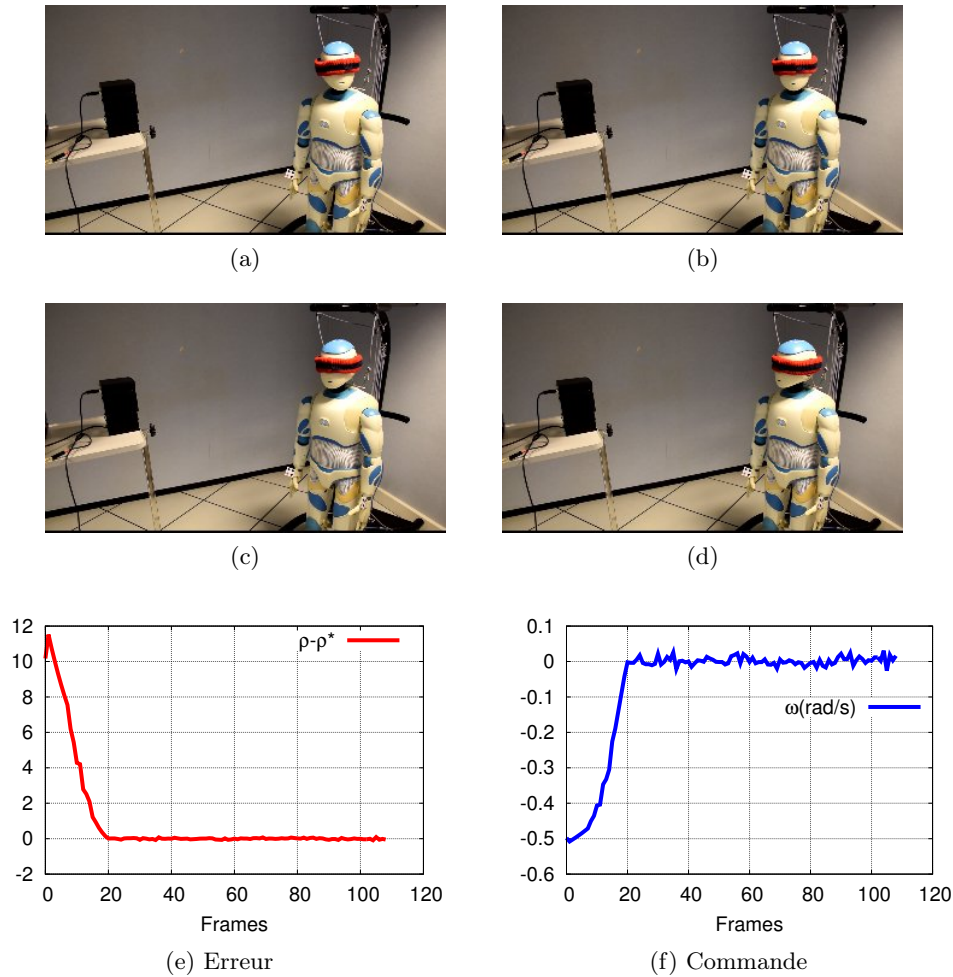


Figure 2: *Romeo* oriente sa tête correctement vers la source sonore en utilisant les indices sonores produits par la différence interaurale en temps.

Si l'on se restreint aux systèmes binauraux, les résultats expérimentaux exposés tout au long de cette thèse ne sont pas réalisables par l'état de l'art de l'audition robotique. Au-delà de ce travail, nous aimerions envisager cette approche comme un cadre général qui peut être appliqué à tout type de robots et de conditions acoustiques. À cette fin, les méthodes présentées dans cette thèse pourraient être améliorées afin de gérer des sources sonores non continues ou le suivi et l'évaluation du nombre de sources actives.

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# Introduction

*Do robots necessarily need to localize sound source(s) in order to engage into interaction?* This question builds up the central thread of this thesis. Indeed, towards the dream of autonomous human-robot interactions, a robot endowed with a reliable sense of hearing is essential. Nowadays, this sense of hearing is generally exploited through auditory capabilities such as sound localization, along with sound separation, speech processing and auditory scene analysis, that build up the topic of *robot audition*. In this topic, particular attention has been given to sound source localization, certainly because it is the first step of auditory-based interaction. The localization step allows to control robot motions for positioning tasks such as facing or approaching a speaker in a scene.

Binaural sound source localization techniques are inspired by the way mammals and more particularly humans sense and perceive sound. Typical approaches for sound localization consist in transposing knowledge in psychoacoustics and physiology into an artificial hearing system that extracts auditory cues. Auditory cues such as the interaural time difference and the interaural level difference or the spectral notches, provide information about the direction of arrival of the sound in the horizontal and the vertical plane. As a result the efficiency of the localization process depends principally on the accuracy of the extracted auditory cues.

However mimicking the human hearing system happens to be complex in realistic configurations. Perception and more particularly auditory cues are influenced by the shape of the body (head, pinnae, torso...) along with acoustic conditions (acoustics of the room, noise, reverberation). The sensitivity to auditory events is variable for each individual and each location. As a consequence, even the most complex artificial auditory systems can be deployed in controlled environments only. Real world configurations, that include dynamic scenes, reverberation, noises, degrade drastically the accuracy of auditory cues and at the same time the localization performance.

Still, new lines of approaches have emerged by studying in depth the capabilities of the human hearing system: relying only on the relevance and precision of auditory cues does not provide enough robustness to realistic environments. Actually a typical analysis of the performance of localization by human listeners as proposed by [Pop08] or [CLH97] shows variable accuracy among individuals in estimating the direction of sound arrival. In these experiments the localization process showed good accuracy for sound stimuli located around the head front sight while eccentric positions of the source lead to more substantial errors. On top of that, other studies [ALO90, CTS68] reported inaccurate distance estimations, which tends to be largely underestimated.

Despite this lack of accuracy, tasks such as head-turn reflex, which consists in turning the head in the direction of the sound source are performed accurately and effortlessly by human listeners. Thus, it could be deduced that accurate auditory cues may not be essential for auditory-based interactions. This assumption is also corroborated by [HCM83], that showed how subjects suffering from hearing loss developed localization strategies based on erroneous auditory cues, by using body motion and visual searching. By contrast, when hearing aids were used in order to recover correct auditory cues, the test subjects localization performance was immediately degraded. Their hearing system was adapted to incorrect auditory cues.

In view of these results, it can be suggested that the key points of the human auditory system are not the accuracy of auditory cues but rather the adaptability and flexibility to acoustic configurations as shown by [HVRVO98]. Actually an efficient hearing system can be defined as a mechanism that combines different input modalities, such as additional visual cues (spatial references), motions and memory of previous auditory experiences. The acoustic space is then apprehended dynamically and potential auditory cues errors can be compensated by the latter modalities. Different paradigms can then be considered such as audio-visual localization, localization through learning-based approaches, or through efficient motion strategies.

This thesis focuses more particularly on motion strategies governed by auditory perception, in a binaural context. Such strategies have been explored in the so-called *active audition* paradigm [NOK00b]. This method consists in measuring the auditory cues while generating a predefined robot motion in order to improve auditory cues accuracy. The localization step is performed afterwards by fusing the measurements taken from different poses of the robot. This approach improves static sound localization but faces some limitations. Active audition aims to localize sound sources. Knowing that robots and acoustic conditions influence the auditory cues, this approach requires a good modeling of the acoustic conditions, similarly to sound source localization in a static configuration. Furthermore, dynamic scenes (*i.e.* moving sensors/ sources) need additional processing such as tracking so that auditory cues measurements remains consistent.

As an answer to the initial question about the necessity of localization, the work developed in this thesis differs from the approaches discussed before by defining an auditory-based interaction as a task modeled in a sensor-based framework. This framework, that we term *aural servo* (AS), does not localize sound sources. A task is characterized by a state of measurements to be satisfied, that generally corresponds to a particular pose of the robot. For instance, in AS, a homing task consists in positioning the robot so as to satisfy given conditions defined by a set of auditory measurements, while a localization-based approach would require to extract the bearing angles and the distance to the sound source, before moving the robot to the desired location. In AS paradigm, robot motion is generated in real time, through a control loop based on auditory cues measured on the fly. Velocity inputs of the robot are obtained by modeling the relationship between the variation of given auditory cues and of the sensors motion. In this way, our approach principally relies on the variation of these cues instead of their intrinsic values. Consequently the localization step is skipped and the modeling of the acoustic conditions or the effect of the

body/head are not required to be modeled. Hence, it is even possible to complete tasks with erroneous auditory cues as long as the variation of these cues remains consistent. Indeed it appears that the variation of auditory cues is more consistent and less sensitive to acoustic conditions than their intrinsic values. This approach applied to robot audition field is the key contribution of this thesis.

The manuscript is thus centered on the description of the AS paradigm, that introduces a sensor-based framework, models auditory cues for the latter framework. Experimental studies validate this approach in real world and dynamic configurations either on free field microphones or microphones embedded on humanoid robots. These experiments start from elementary auditory-based tasks such as head-turn motions, to autonomous navigation tasks.

This manuscript is organized in 6 chapters. Chapter 1 introduces the concept of robot audition. This chapter provides an historical review of the main contributions related to the topic of robot audition. Then, the main applications and contributions, but also the limits of robot audition are summarized.

In Chapter 2, we focus on sound source localization, one of the most prominent topic in robot audition. As a first step, the localization process is described from the human sensing perspective, in order to give an insight of all physiological mechanisms involved in sound localization process. In the following, this chapter gives an overview of localization techniques, and major achievements and applications in robot audition field.

From the limitations underlined in the two previous chapters, in Chapter 3 we introduce the framework that builds up the AS paradigm. Before presenting the theoretical concepts involved in AS, we stress the potential benefits of using a sensor-based control through an example comparing an approach based on localization and an approach based on AS. Thereafter, the essential theoretical aspects used in our approach are exposed.

Chapters 4 and 5 that are at the core of this thesis, propose a full modeling and analysis of auditory cues (Interchannel Level Difference, Interchannel Time Difference, Energy level) applied to the AS paradigm. These approaches are evaluated through simulations that exhibit the benefits and the limitations of such a paradigm. More importantly, all these approaches are validated experimentally on a mobile robot, in real world environments for several tasks and under various contexts. These experiments can be considered as contributions that support the relevance of our approach.

Eventually Chapter 6 emphasizes an essential key point of our approach: the versatility of the AS paradigm, that can then be considered as a general framework. More specifically, in this chapter we apply our approach on humanoid robots despite a free-field modeling. Once again experimental results corroborate the pertinence of AS for real world applications.



# Chapter 1

## Robot audition

This chapter introduces the context of this thesis. Notably we define the concept of robot audition and the related state-of-the-art. The chapter is divided into three main parts.

The first one gives a brief historical review of progress in robotics and more particularly the developments that conducted to endow robots with auditory capabilities. Following this description, we provide an example of an auditory scene that highlights the needs and expectations from robots endowed with the sense of hearing. Eventually, Section 1.1.3 specifies the functionalities and applications ensued from this sense, shaping the robot audition topic.

The second part of this chapter concerns robot audition challenges in order to be deployed in realistic environments. First, we focus on the sound propagation and its properties, which influence the cues extracted from given signals. In a second phase, robotic context is analyzed through of the additional constraints that it introduces.

Finally, the third part of this chapter in Section 1.3 presents the current level of development of robot audition through software platforms, research projects and conferences related to this topic.

### 1.1 The sense of hearing in robotics

#### 1.1.1 From robotic arms to robot audition

Since the seventies, the definition of and the expectations from robots have drastically changed. In the beginning of robotics, which coincided with the boom of industry, robots were considered as efficient tools to expand and develop industrial processes. This period marks the beginning of fully automated tasks designed in order to replace human force for exhausting and tedious works. The first robots were mainly used in controlled environment. Pre-configured robotic arms were programmed for repetitive and tough tasks, such as pick-and-place tasks, palletizing or welding in assembly lines. The environment was static and/or strictly controlled in order to avoid any interruption in working chains and more importantly to avoid accidents in robots working space.



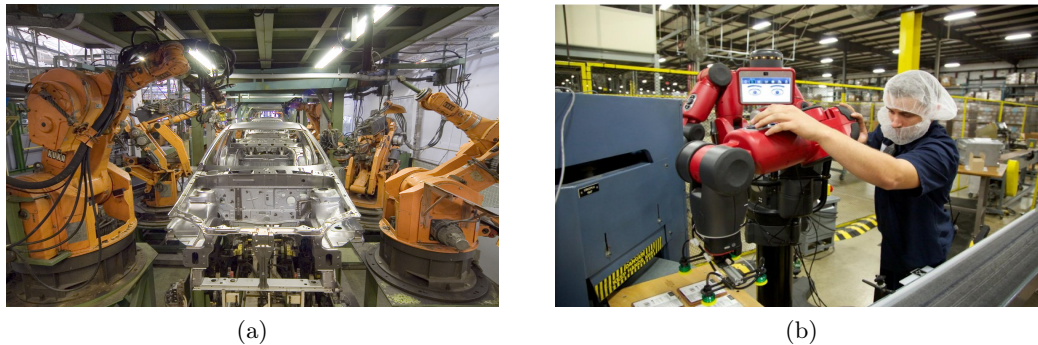


Figure 1.1: Evolution of industrial robots from the last three decades: in (a) Kuka robots on assembly lines in 1983<sup>1</sup>, in (b) Nowadays an assembly line with the robot Baxter (courtesy to Rethink Robotics).

Since that period, robots capabilities have been greatly boosted, although robotic achievements are currently far from the fantasized ideal human-skilled robots. The trigger of this revolution is probably the use of sensors, that made robot more “intelligent” than automated arms. “Intelligent” underlies the concept of advanced skills of sensing, learning and adaptation to unknown and uncontrolled environments. These sensors are from two types and either make robots aware of their internal state or aware of their environment. The internal state of robots is evaluated by proprioceptive sensors that may give measurements about battery level, wheels position, joint position, etc. Among proprioceptive sensors, one can cite encoders, potentiometers, gyroscopes or compasses. These sensors are particularly useful in industrial contexts in order to detect any defect in robots or breakdown in working chains. On the other hand, the environment is observed from exteroceptive sensors. Depending on the context, detecting/analyzing objects or humans and detecting changes in the environment are the main features offered by this type of sensors. Typical exteroceptive sensors include cameras, sonars, lasers, force sensors, microphones, etc. Thus, it is without saying that exteroceptive sensors are greatly used for tasks involving interaction, humans or uncontrolled environments.

Globally, sensors widened the field of application of robotics and democratized the presence of robots in our daily life. In industry, robots tend to be more compliant and to foster human intervention through synergistic tasks (see Figure 1.1b). Besides, the use of robots is no more limited to industrial contexts. Indeed robots are more and more involved in tasks such as exploration, monitoring, surveillance, cleaning, socialization or used as leisure companions. The two latter use cases underline the rise of social robots that are autonomously able to interact and to communicate with humans while being endowed with a social behavior. These robots are particularly used in assisting context such as with diseased children [AVS15] or with elderly people [BHR09].

This wide variety of potential applications allowed to diversify robot structures,

<sup>1</sup> By Mixabest CC BY-SA 3.0, commons.wikimedia.org/w/index.php?curid=9820288

that are not restricted to arms anymore. Indeed, in connection with the sensors that permitted free and autonomous motions, various type of robots emerged: nano/micro robots, wheeled-robots, legged robots, flying robots or humanoid robots among others. In parallel, thanks to a better understanding of the physiological properties of mammals bodies, senses or behaviors, robots tend to become bio-mimetic platforms. For this purpose, roboticists particularly focus on human-like robots, that is, robots endowed with human locomotion capabilities, behavior or sensing skills. The three main senses investigated in robotics are the senses of vision, touch/range sensing and hearing. Among these senses, vision is probably the most prominent and one of oldest subject of study in robotics. Thanks to the astounding progress in electronics, cameras rapidly became embeddable and affordable and subsequently available for robots. On top of that, the support of the long expertise and progress in computer vision field greatly eased knowledge transfer. Robots are nowadays endowed with techniques of detection, recognition, tracking or localization. The sense of touch (*e.g.*, proximity, tactile or force sensing) received more and more attention as cohabitation between robots and humans increased. The main subjects of study concern collision prevention and detection for safety purpose, grasping tasks that lead to interact with objects of various shape and stiffness, or even locomotion tasks.

Eventually the sense of hearing, that is the main thread of this thesis, consists in endowing the robot with capabilities of analyzing acoustic scenes. Such capabilities require functions for localizing, separating, identifying or recognizing sound sources. Robot audition is then an inter-disciplinary field at the confluence of signal processing, acoustics, artificial intelligence and control. Robot audition is a quite “recent” topic that has been extensively studied during the last two decades. However, the first trace of auditory capabilities embedded on a robot dates back from 1973 and the first bipedal robot *WABOT-1* [KOK<sup>+</sup>74] (see Figure 1.2) from Waseda University. This robot was equipped with artificial “ears” for speech recognition and processing and was able to understand simple Japanese commands such as "STOP", "BEGIN", "TURN LEFT". The following series of robots issued from *WABOT-1* were also all equipped with auditory capabilities. Ten years later, *WABOT-2* [KOS<sup>+</sup>87] and in the continuity *WASUBOT*, piano-playing robots, were able to converse with a person with a more evolved speech understanding than *WABOT-1*, and to play instruments accordingly to a singing voice from a pitch analysis.

In parallel, the idea of considering robots as validation platforms, especially in neuro-science, greatly contributed to develop and to democratize the sense of hearing for robots in early nineties. In [Web93] and [Web95], the author designed a mobile robot equipped with two microphones to model the behavior of crickets based on sound localization and recognition. Similarly, in [RWE98] a robot head equipped with two microphones and a camera was used to acknowledge the owl neural hearing model proposed by the authors.

As robotics became a research topic of its own, more contributions were centered on providing robots with auditory capabilities. Auditory capabilities including sound localization and separation based on three microphones were thus proposed in [HOS95]. In the middle of the nineties, the authors of [YA95] proposed one of the first robotic systems endowed with auditory capabilities for a purpose of human-

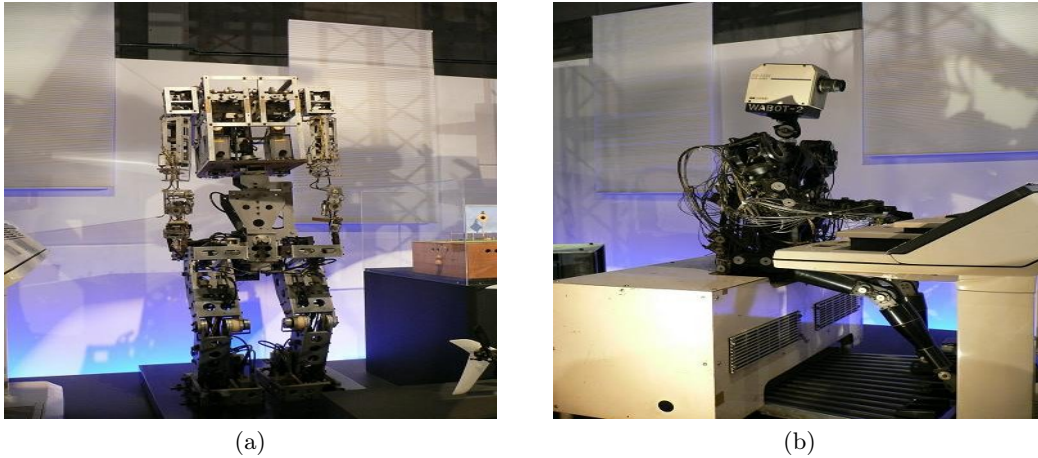


Figure 1.2: The first robots with auditory capabilities developed at Waseda University: (a) *WABOT-1* (1973) a bipedal robot equipped with artificial ears that sustains conversation in Japanese, (b) *WABOT-2* (1984) a piano-playing robot that can accompany a human singer.

robot interaction. The robot *Chaser* was equipped with a heat sensor, ultrasound sensors, infrared sensors, touch sensors but also three microphones. By fusing the measurements of all these sensors, this robot was able to locate and to approach a speaker and to start a dialogue thanks to a speech processing module. *Chaser* was mainly designed in order to validate the concept of *active interface* proposed by the authors. In the continuity of this work, the authors of [SII96] developed a pet robot that is able to localize a speaker from visual and auditory cues by using a neural network. In the same period, [Iri95] proposed to develop auditory capabilities of humanoid robots with a sound localization framework based on neural networks, through the robot *COG*. This latter work was followed by contributions on social robots such as *Hadaly* of [HNK<sup>+</sup>97] that can localize the speaker as well as recognize speech with a microphone array. Similarly the robot *Jijo-2* introduced in [AHH<sup>+</sup>97] could understand a voice command. The auditory capabilities were also extended to navigation tasks in [HSTO97], where the sound localization coupled with sonar sensors allowed a mobile robot to approach a sound source while avoiding obstacles. The so-called *cocktail party effect* (see Section 1.1.3), which characterizes the phenomenon of being able to focus the auditory attention on a particular aural stimulus was primarily addressed in [HOS97]. These fundamental researches inspired the birth of the “robot audition” concept.

Lately, with the recent development of autonomous humanoid robots and companion robots, new challenges (see Section 1.2) related to robot hearing emerged. These challenges induced by the autonomy of robots (*i.e.*, flexibility to unknown environments) renewed the interest of robot hearing through the concept of robot audition, as acknowledged nowadays. In [NOK00b] the authors defined this concept as robots’ ability to localize, separate, recognize sounds in realistic environments.

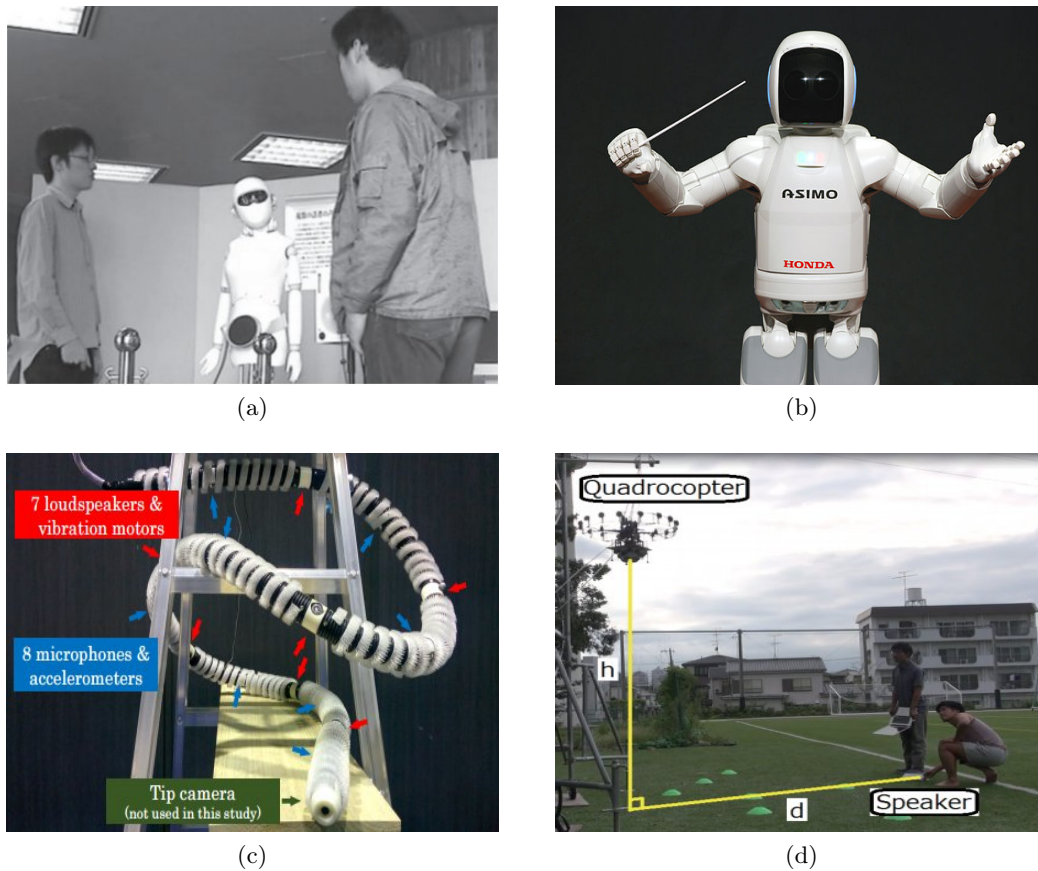


Figure 1.3: A wide variety of nowadays applications related to robot audition: (a) A human-robot dialogue proposed in [TOO14], (b) Asimo leading an orchestra based on tempo and beat detection<sup>2</sup>, (c) a hose-shaped robot developed for search-and-rescue mission in narrow and cluttered environment [BIK<sup>+</sup>15] and (d) sound localization endowed on an aerial robot [ONM<sup>+</sup>14].

This definition gave birth to different research topics oriented towards the design of artificial auditory systems. These topics focused on the scattering effect of the head [NMOK03, NOK01] or on the design of artificial auditory systems, such as pinnae. For instance, the artificial pinna proposed in [KN11] can move and deform its shape actively in order to improve the sound perception. In the same vein, the authors of [MAB11], [TR13] or [SD11] studied the best microphones topology to optimize the auditory capabilities. More specifically in [MAB11], the authors proposed a reconfigurable microphone array. Different paradigms of localization were also investigated such as binaural approaches based on two microphones, array-based approaches or more recently *active audition* [NOK00b] that uses motion. These methods will be detailed in Chapter 2. Furthermore, the diversity of robots widened the field of

<sup>2</sup>By Vanillase -CC BY-SA 3.0, commons.wikimedia.org/w/index.php?curid=17300496



application of robot audition. Robot audition is also a topic related to entertaining robots like dancing robots or music playing robots, as a legacy of the first *WABOT* robots. In this context, the authors of [ANK<sup>+</sup>99] proposed a Theremin-playing robot that uses the audio-feedback related to the sound pitch to control its arm. In [OBI<sup>+</sup>15], the authors presented a robot that can dance synchronously with music and human dancers from music beats detection. Applications in scenes of disaster is also a recent field of research of robot audition technologies. In [BSLF12] the authors introduced an Unmanned Aerial Vehicle (UAV) system that is able to localize emergency whistles.

This wide variety of applications as illustrated in Figure 1.3 let us foresee that robot audition becomes increasingly important in the field of robotics. However robot audition is still in a growth and maturation phase. Despite a growing interest reflected by the increasing number of projects (see Section 1.3), robot audition community remains modest in size compared to fields like vision. But on the other hand robot audition offers an exciting context that allows ample space for improvements and that promotes original contributions. Indeed most of the applications discussed above are for now limited by the current level of maturity of robot audition: there is still a long way to go before reaching substantially autonomous robots with auditory capabilities.

### 1.1.2 A typical scenario involving robot audition capabilities

To illustrate the auditory capabilities required for robot audition, let us start from the scene pictured in Figure 1.4. This scene depicts a common everyday life living room in which interactions between humans and a robot are occurring. In this environment the robot should be able to interact and apprehend the surrounding environment with high flexibility. In particular several sensory stimuli may trigger an action of the robot. From an acoustic point of view, several sound sources may be active independently to each other. Domestic noises, such as television, radio, vacuum cleaner, or simply people are all potential sound sources or disturbances. More specifically, in the illustrated scene, the robot action is triggered by the boy asking the robot to retrieve the ball to him. This trivial task would be performed efficiently and effortlessly by a human. For robots, this task underlines the need of auditory capabilities. First, an essential question arises: from all the sensory stimuli, how to focus the attention on relevant signals and discard spurious signals? In our example, the robot must be able to segregate sound sources from a signal mixture, and then classify them with respect to their relevance. Furthermore the robot must localize these sources and identify each of them uniquely: the robot is acoustically monitoring the scene. From this monitoring stage, the robot can determine the location and identity of the boy. The signal content should also be deciphered through speech recognition process. The output from the speech recognition module is then interpreted as an action: the robot heads towards the ball. During robot motion the hearing modules are not disabled. Any new "auditory command" should be interpreted immediately, and the sources of interest are likely to move. A continuous and real-time sound processing module is running on the robot in order to update the

sounds location/presence or the position of the robot with respect to these sources. Furthermore, the motion of the robot generates noise, hence each relevant sound signal monitored is enhanced by denoising and/or dereverberation modules. The ball is finally reached and taken by the robot. Thereafter, the robot heads towards the boy using localization and identification information. This complex sequence of

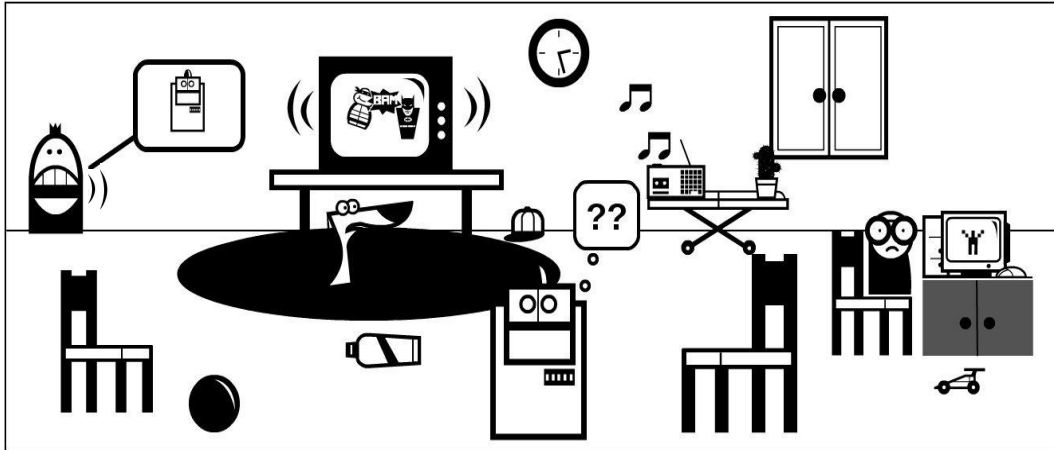


Figure 1.4: A typical acoustic scene to be apprehended autonomously by a robot. Drawing made from ([www.stripgenerator.com](http://www.stripgenerator.com))

tasks emphasizes complementary capabilities that are based on hearing skills. Endowing a robot with each of these capabilities is a subject of research of its own. The interest in robot audition being quite recent as stated in the previous section, it goes without saying that endowing robots with this kind of hearing sense as imagined in the above scenario, is still an intensive subject of research, that is not achieved yet. In this perspective, more formally, the robot is required to solve questions such as: *Where are the source(s)? What are the sound source(s) of interest? What are they saying? Who are the speakers? When are they speaking?* These questions could also be extended to advanced skills that require to solve: *Where am I? What is the shape of the environment? What are the objects?* The capabilities of solving these problems build up the field of robot audition. These sets of questions let us foresee two types of auditory-based interaction: the first one concerns the analysis of the sound sources already present in the environment and requires sound localization, sound separation and classification, speech recognition and can potentially involve navigation tasks. The second category consists in analyzing the environment from its acoustic properties and includes methods such as self localization or scene mapping. These interactions are detailed in the next section.

### 1.1.3 Hearing skills for robot audition

#### 1.1.3.1 Perception of the sound sources

*Where are the source(s)? What are the sound source(s) of interest? What are they saying? Who are the speakers? When?*

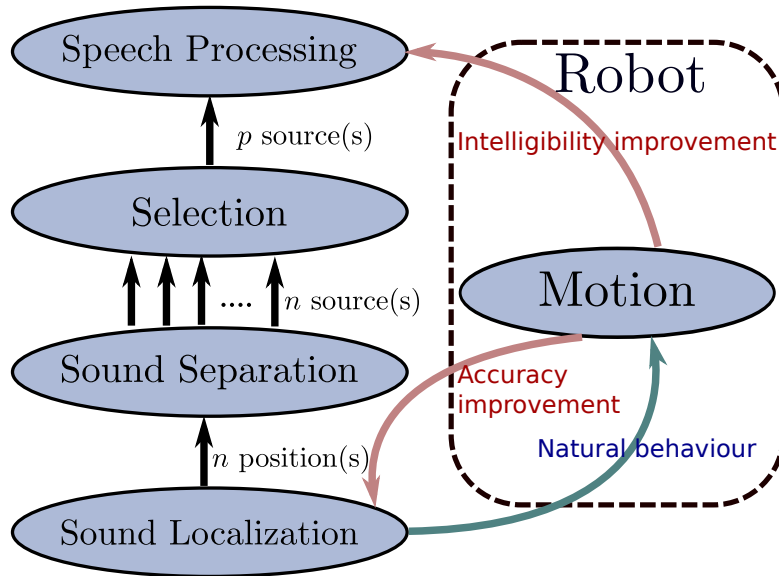


Figure 1.5: Computation auditory scene analysis: the robotic context adds a supplementary element, which is motion

These issues have been addressed under the theme of computational auditory scene analysis (CASA) in [BC94] or [RO98]. By essence, CASA systems are not specifically designed for robots, but rather consist in a "machine" understanding of an arbitrary sound mixture similarly to human listeners. Fundamentally, CASA is inspired by the psychoacoustic knowledge on the human auditory capabilities. Physiologically, human abilities for solving these initial questions have been described in 1953 [Che53] as the *cocktail party effect*. The *cocktail party effect* is an imaged reference to the human ability to selectively focus and recognize a given sound source in a noisy environment, as it could happen in a cocktail party. The authors of [HC05] modeled this process with three different neural steps that are

- **Analysis.** The first step principally consists in segregating the sound sources with respect to their location. The sound mixture is split so that sounds emitted from the same location are grouped together.
- **Recognition.** The second step consists in analyzing the statistical structure and patterns in each segregated group obtained from the analysis. These patterns allow to identify a signal of interest among the segregated groups.
- **Synthesis.** The last step consists in extracting and reconstructing the clean (deconvolved from reverberation) signal of interest from each segregated group. This step provides the capability of focusing attention on a signal of particular interest for the listener.

The *cocktail party effect* discussed above gives a quite generic overview on the requirements of CASA for a defined system. Globally the neural process involved in

solving the *cocktail party problem* does not necessarily have a precise order and the boundaries between each process is not as neat as it is exposed above. In psychoacoustics there is not yet a global consensus about the neural processing of sound: several studies brought to light experimental evidence supporting different order of processing [Bre94] or [DH97].

In robotics, CASA frameworks are especially centered on human-robot interactions. As exposed in [OOKN04] or [ON15] the interaction processes are generally expressed in a bottom-up framework, depicted in Figure 1.5, that is conceptually easier to develop. In this framework the sound localization is performed first. Then from the obtained location(s), sound separation and selection can be performed if required. The last steps consist in sound identification and processing. The CASA framework emphasizes the prominence of sound source localization that is a prerequisite to guarantee the successful completion of interactions. This framework is slightly modified by robotic context that adds a component to CASA: motion. Indeed, the localization process can control robot motions to favor the naturalness of interactions. For instance, it is expected that the robot should face, at a given distance, the speaker whom the interaction is established with. In return, motion can improve localization with the fusion of measurements from different poses, which corresponds to *active audition* principle [NOK00b]. Speech processing may also be improved by guiding the robot in a pose that optimizes the speech intelligibility. This is applied, for instance, by the authors of [KFKI10] who propose a motion planning strategy, based on occupancy grid, in order to maximize the effectiveness of speech recognition. Nonetheless the motion that can be considered, at first glance, as a valuable addition to CASA framework raises new challenges related to dynamic perception of auditory data, that would require real-time and flexible processing as well as robust tracking (see Section 1.2).

During the last decade, several contributions aimed to apply CASA on robots. These contributions especially focused on sound localization and separation. This is the case of [DH12], that learns the relationship between the auditory perception and the robot pose. CASA has also been applied to given contexts, as the work [NIYO15] illustrated by Figure 1.6. In this paper, the robot leading a quiz game, is required to perform sound localization and separation, speech processing and identification in presence of multiple speakers that are potentially competing. Similarly, a cooking robot described in [KSN15] is endowed with advanced auditory scene understanding in order to support cooking recipes. Nonetheless these applications of CASA remains far from realistic conditions. Indeed they are generally performed in controlled and static environments. For instance, Figure 1.6 is a relevant illustration of the current limitations of CASA, as floor markers are used to accurately position the robot with respect to human players.

Robot audition is not limited to human-robot interactions, and several applications may be derived from CASA frameworks. The core statement of robot audition is rather “*robots should understand all sounds picked up by their ears, not just voices*”. First, interactions with non-verbal, non-speech sounds might be useful for some particular applications. In entertainment applications, music playing or dancing robots require similar robot audition functions. Dancing robots need to track the musi-





Figure 1.6: Typical CASA application: a robot quiz-master is able to locate, separate and process sound sources in a controlled environment [NIYO15].

cal beat in order to perform a coherent choreography [YNT<sup>+</sup>07]. In the case of [ONT<sup>+</sup>11], a framework allows a robot to play music in collaboration with a human performer by analyzing music temporal fluctuation and pitch. These approaches exhibit the strong link between auditory features (*e.g.*, beat, pitch...) and robot control. Moreover the sense of hearing is also useful for mobile robotics and brings new navigation modalities based on sound localization. One of the first navigation system based on audition is exposed in [HST<sup>+</sup>99]. This system consists in a finite state automaton combining the sound localization results to a sonar system for obstacle avoidance, in order to autonomously approach a sound source with a mobile robot. This kind of approach has also recently emerged in the context of distributed multi-robot systems. For instance in [BSLF16] a leader-follower scheme is applied to drones in a navigation task based on the localization of the leader.

### 1.1.3.2 Perception of the environment

*Where am I? What is the shape of the environment? What are the objects?*

This second set of questions refers to robot capabilities to sense and analyze their surrounding environment from auditory stimuli. Conversely to source-oriented perception that aims to directly interact with the sources of interest, this section deals with a higher-level of auditory scene understanding. Indeed, sound from its properties (*e.g.*, acoustic, location, type) can provide information describing the environment structure.

This is the case for mapping and scene reconstruction that can be performed with acoustic information. Sound mapping consists in building an auditory representation of the environment, so that zones of interest can be monitored. In the case of mobile robots, sound mapping is developed as an automated tool for condition monitoring, for instance in factories that often contain noisy machines. In this type of environment, any malfunctioning in a machine will be reflected in an acoustic change of the

acoustic map. For instance, in [KMIH13] an acoustic map is created while a robot is autonomously navigating across an environment by continuously generating audio scans with a steered response power algorithm (detailed in Chapter 2.3). This kind of approach has also been extended to 3D cases in [EMK<sup>+</sup>14a]. In this paper, a map is created by a mobile platform equipped with a microphone array and laser range sensors. Similarly, the authors of [EFW<sup>+</sup>15] proposed a 3D sound mapping using RGB-D sensor to build clouds of 3D auditory points. Globally, these approaches can be considered as extensions of sound source localization and separation techniques: the map is build from the locations of the different sound sources. This kind of sound localization and separation remains complex to perform because these approaches should cope with an unknown time-varying number of sound sources that may have different levels. These constraints explain that most of these approaches concerning mapping are performed in empty scenes. In environments such as like factories, sounds produced by the machines are loud, constant, still, and do not vary in number. Contrarily, sound mapping in populated spaces remains extremely complex, notably because of the non-stationarity of speech, the dynamic nature of the environment and the wide variety of potential sound sources as illustrated in our previous example in Section 1.1.2. Nevertheless auditory scene mapping is of a high interest for interactions, since it would provide auditory awareness ability to robots, that could then focus on or change more easily of interest points. Until now, only few works are interested in this kind of mapping because of the task complexity. This problem is addressed in [KGD13] using the cepstral coefficients of the recorded sounds to classify and define several acoustic regions. This approach is however not suitable for real-time implementation and is performed offline. One can also cite the project EARS (see Section 1.3) that targets a demonstration application of a welcoming robot at the reception desk of hotel lobby in 2017.

Other approaches concern self-localization of robots using auditory perception. In this context, sound sources can be considered as acoustic landmarks that can ease robot navigation. One of the first approaches of localization was exposed in [WIA04], in which an array of 24 microphones embedded in walls was used to localize a robot. With this system, a robot was able to perform guiding tasks. In connection with search-and-rescue missions, the authors of [BIK<sup>+</sup>15] proposed a framework to estimate the pose of a hose-shaped robot carrying at the same time several microphones and speakers. In the continuity of approaches aiming to localize the robot, acoustic SLAM (Simultaneous Localization And Mapping) is a recent topic of study. The work exposed in [HCW<sup>+</sup>11] proposes an acoustic SLAM framework that computes the robot path using a particle filter and the bearing information of several sources. More recently the authors of [EMN16] developed a framework using the acoustic bearing information on the basis of a probability hypothesis density filter.

In a quite different context, acoustic sensing based on echoes is also a topic of interest. This approach relies on the acoustic responses of the environment with respect to an emitted sound. By analyzing these acoustic responses, information about the surrounding environment can be obtained. This approach can be opposed to acoustic simulator or real-time auralization techniques that estimate auditory sensations from rooms' geometrical properties. Auditory responses carry information

such as room geometry, material properties, cluttering and potential obstacles. Such an approach can be related to mammals like bats or dolphins that use sounds and more particularly echoes, to perceive their surrounding environment. In this topic, [KKK<sup>+</sup>15] proposed a hammering sound analysis for infrastructure inspection of industrial field based on legged robots. Within this project, a robot can attest of the quality of steel bars from the sound produced by a hammering step. [EMK<sup>+</sup>14b] used reflections to detect and avoid objects for instance at intersection of corridors, in order to prevent collision with unseen objects. In [SUW15] a robot is able to localize itself in a pre-explored structured environment with odometry measurements and environment acoustic responses. One can eventually cite [KDV16] proposing an echoSLAM framework that can reconstruct the shape of a given environment from wall echoes and potentially retrieve the position of the robot.

## 1.2 Challenges in robot audition

The tasks evoked in the previous section, remains challenging in unknown environments. Until now, there is still a gap between actual auditory capabilities of robots and auditory scenarios such as the one described in Section 1.1.2. For instance, most of human-robot dialogues require the attendant to wear headset microphones or at worst to be very close to the robot. In order to characterize the limitations of robot audition in realistic environments, it is important, first, to understand sound properties. Sound as a wave is defined by its propagation model and behavior at physical boundaries. From these properties that influence perception the challenges faced in robot audition field naturally arise.

### 1.2.1 Spherical sound propagation

According to the definition provided by the Acoustical Society of America (ASA) sound can be defined as "*(a) Oscillation in pressure, stress, particle displacement, particle velocity, etc., propagated in a medium with internal forces (e.g., elastic or viscous), or the superposition of such propagated oscillation. (b) Auditory sensation evoked by the oscillation described in (a)*". The definition proposed by ASA covers two distinct but complementary ways to analyze an acoustic scene. In this section, we are more interested in the first definition that considers sound as a physical phenomenon described by wave propagation. Wave propagation does not only describe sound but also light or fluids like water as illustrated in Figure 1.7. In our context, acoustic theory describes the propagation from a known sound source through air or any other medium as formulated in [Str77] and [Str96]. The fundamental results on sound propagation are derived from what is known as the wave equations, that are partial differential equations that govern all patterns of wave propagation. To find a solution to these equations, sound sources are often modeled as vibrating surfaces which have known surface velocities or pressures. One can notice that in a loose sense robot audition corresponds to an acoustic inverse problem: the sound field that is measured by sensors is known while sound sources must be calculated with respect to their location, type or contents.

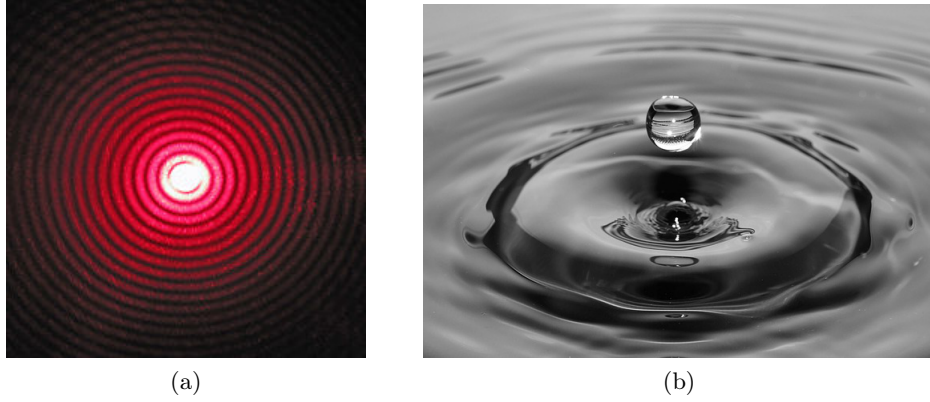


Figure 1.7: Example of planar radial wave propagation: (a) in the case of a diffracted laser beam, (b) in the case of water ripples

In an homogeneous and linear medium such as the air, sound propagation is governed by the following equations

$$\begin{cases} \nabla^2 p(\mathbf{x}, t) - \frac{1}{c^2} \frac{\partial^2}{\partial t^2} p(\mathbf{x}, t) = 0. \\ p(\mathbf{x}, t_0) = g_1(\mathbf{x}) \\ \lim_{\mathbf{x} \rightarrow \mathbf{b}_i} p(\mathbf{x}, t) = g_2(\mathbf{x}) \end{cases} \quad (1.1)$$

where  $p(\mathbf{x}, t)$  is sound pressure at time  $t$  and point  $\mathbf{x}(x, y, z)$ .  $\nabla^2$  is the Laplacian expressed in 3D coordinates, and  $c$  the constant wave speed inherent to a given homogeneous medium. The initial conditions  $p(\mathbf{x}, 0) = g_1(\mathbf{x})$  and the boundary conditions  $\lim_{\mathbf{x} \rightarrow \mathbf{b}_i} p(\mathbf{x}, t) = g_2(\mathbf{x})$  with  $\mathbf{b}_i$  as boundaries, are required to obtain a unique solution. Note that (1.1) can also be expressed in function of the velocity potential  $\phi(\mathbf{x}, t)$  or in function of particle velocity  $\mathbf{u}(\mathbf{x}, t)$ , instead of the sound pressure  $p(\mathbf{x}, t)$ . In our context we are particularly interested by the solutions implied by a spherical sound radiation caused by a point source in a free-field (*i.e.*,  $\mathbf{b}_i = \infty$ ). Point sources correspond mathematically to points in space that radiate sound equally in all directions, producing a spherical wave front as illustrated in Figure 1.8. This modeling is interesting since it eases the modeling of reverberations and reflections (see Section 1.2.2.2). Given a source signal  $s(t)$  emitted from a position  $\mathbf{x}_0$ , the wave equation (1.1) expressed in spherical coordinates [MI68] for the pressure  $p(\mathbf{x}, t)$  at the time  $t$  is

$$\nabla^2 p(\mathbf{x}, t) - \frac{1}{c^2} \frac{\partial^2}{\partial t^2} p(\mathbf{x}, t) = -f_{\mathbf{x}_0}(\ell, t). \quad (1.2)$$

$f_{\mathbf{x}_0}(\ell, t)$  is the source function and  $\ell$  the radial distance between the point source  $\mathbf{x}_0$  and the observation point  $\mathbf{x}$ . Considering point sources,  $f_{\mathbf{x}_0}(\ell, t)$  characterizes that the sound source occupies a limited and compact region in the space. At each instant  $t$ ,  $f_{\mathbf{x}_0}(\ell, t)$  corresponds to a Dirac impulse so that

$$f(\ell, t) = \delta(\mathbf{x} - \mathbf{x}_0)s(t). \quad (1.3)$$

Considering the sound pressure  $p(\ell, t)$  at a distance  $\ell$  of  $\mathbf{x}_0$  the wave equation becomes

$$\nabla^2 p(\ell, t) - \frac{1}{c^2} \frac{\partial^2}{\partial t^2} p(\ell, t) = -f_{\mathbf{x}_0}(\ell, t). \quad (1.4)$$

Furthermore, the Laplacian operator is therefore given by

$$\nabla^2 = \frac{1}{\ell^2} \frac{\partial}{\partial \ell} \left( \ell^2 \frac{\partial}{\partial \ell} \right), \quad (1.5)$$

assuming spherical coordinates and isotropic propagation. In the frequency domain, the Fourier transform of (1.4) corresponds to the Helmholtz equation [MI68] that is

$$\nabla^2 p(\ell, \omega) + \frac{\omega^2}{c^2} p(\ell, \omega) = -f(\ell, \omega), \quad (1.6)$$

where  $\omega$  is the angular frequency. The solution of such an equation can be obtained from Green method [Gre28] that transforms (1.6) to

$$\nabla^2 p(\ell, \omega) + \frac{\omega^2}{c^2} g(\ell, \omega) = -\delta(\ell), \quad (1.7)$$

where  $g(\ell, \omega)$  is the solution of the equation also called Green's function when (1.3) is satisfied. The Green's function is given by

$$g(\ell, \omega) = K_1 \frac{e^{-i\frac{\omega\ell}{c}}}{\ell} + K_2 \frac{e^{i\frac{\omega\ell}{c}}}{\ell}, \quad (1.8)$$

in which  $K_1$  and  $K_2$  are constants. The general solution can then be written as a convolution

$$p(\ell, \omega) = \int_{-\infty}^{\infty} f(\tau, \omega) g(\ell - \tau, \omega) d\tau = S(\omega) \left( K_1 \frac{e^{-i\frac{\omega\ell}{c}}}{\ell} + K_2 \frac{e^{i\frac{\omega\ell}{c}}}{\ell} \right). \quad (1.9)$$

From (1.9), it can be observed that the Green's function is a filter that transforms the input signal  $s(t)$  into a radiated spherical wave. However as (1.9) is expressed, two solutions emerge under free-field assumption: the first one representing a sound wave radiating from the point source to infinity, and the second one a sound wave radiating from infinity to the point source. From Sommerfeld radiation conditions [P+81] that state that all sources lie within a bounded region, thus cannot arrive from infinity, it can be deduced the initial conditions  $K_1 = \frac{1}{4\pi}$  and  $K_2 = 0$ . Thus a unique solution can be obtained from 1.9 that becomes

$$p(\ell, \omega) = S(\omega) \frac{e^{-i\frac{\omega\ell}{c}}}{4\pi\ell}. \quad (1.10)$$

Returning in the temporal domain, the latter equation gives the sound pressure as

$$p(\ell, t) = \frac{s(t - \frac{\ell}{c})}{4\pi\ell}. \quad (1.11)$$

Equation (1.11) states that from a point  $\mathbf{x}$  located at a distance  $\ell$  of the source, the sound is delayed by  $\frac{\ell}{c}$  and attenuated by a factor proportional to  $\ell$ . In Cartesian coordinates the solution of the spherical wave propagation in (1.2) is then

$$p(\mathbf{x}, t) = \frac{s(t - \frac{\|\mathbf{x} - \mathbf{x}_0\|}{c})}{4\pi\|\mathbf{x} - \mathbf{x}_0\|}. \quad (1.12)$$

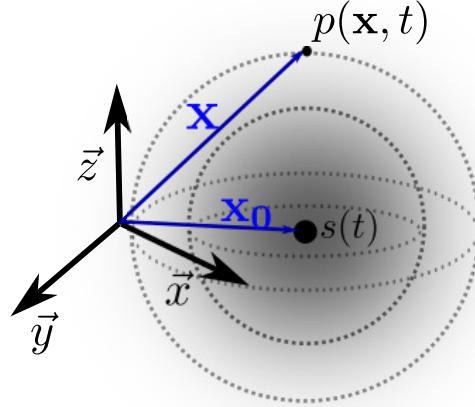


Figure 1.8: The spherical wave propagation in a free-field of a point source  $\mathbf{x}_0$  generating a signal  $s(t)$  leads to observe a pressure  $p(\mathbf{x}, t)$  from a point  $\mathbf{x}$  that is function of the delayed and attenuated version of the signal  $s(t)$ .

## 1.2.2 Environmental constraints

### 1.2.2.1 Observation distance and plane wave approximation

After describing the spherical sound wave propagation, we can focus on the properties implied by such a model. First it is interesting to analyze how sound behaves at different distances of observation. This is characterized by the near-field and far-field sound radiation. If sound pressure is observed at a great distance, the curvature of the wave fronts look flat while at a close distance, the curvature of wave fronts is more pronounced as illustrated in Figure 1.9. The far-field or near-field regions depend on the observation distance from the point source, the microphone array aperture and the sound frequency. Sound is observed from the far-field when in (1.12)  $\|\mathbf{x}_0\| \gg \|\mathbf{x}\|$ . Conversely the near-field corresponds to observation points where  $\|\mathbf{x} - \mathbf{x}_0\| < \lambda$ , with  $\lambda = \frac{c}{f}$  is the wavelength and  $f$  the frequency. Note that audible sound has a wide range of wavelengths from about 20 mm to 17 m, that corresponds to frequencies of 20 KHz and 20 Hz, respectively. In the field of antennas, the boundary between the far-field (Fraunhofer region) and the near-field (Fresnel region)  $d_f$  defined as

$$d_f = \frac{2d^2}{\lambda} \quad (1.13)$$

where  $d$  is the inter-microphone distance. This formula is derived from the Rayleigh distance  $d_r = \frac{d^2}{2\lambda}$  that characterizes the transition zone from the near-field and the far-field.

In robot audition, observing the source from a far-field region (*i.e.*, with flat front

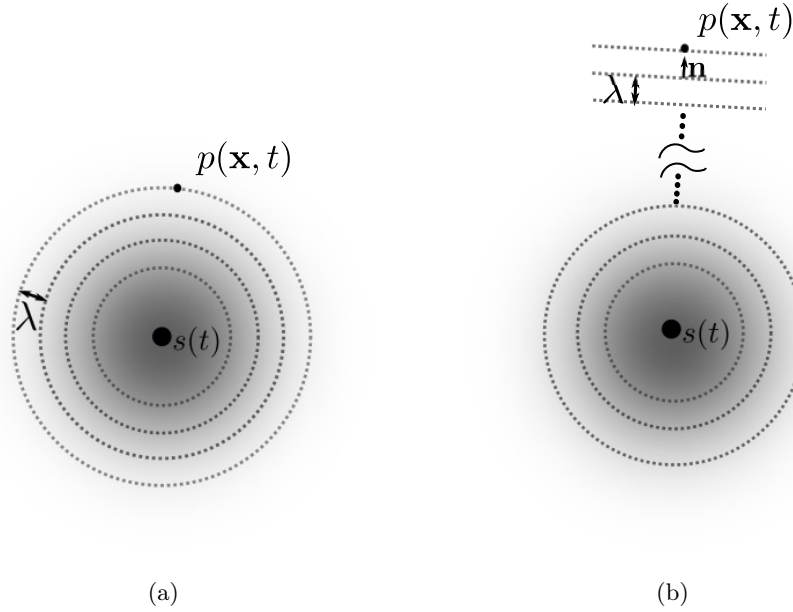


Figure 1.9: 2D view of the propagation wave with a wavelength  $\lambda$  in (a) the near-field where waves are curved and in (b) in the far-field where waves are flat

wave) is particularly convenient. Under far-field conditions one can express

$$\|\mathbf{x} - \mathbf{x}_0\| = \|\mathbf{x}_0\| - \epsilon(\mathbf{x}) \quad (1.14)$$

where  $\epsilon(\mathbf{x})$  is a residual term. When injecting (1.14) in (1.12), the solution to the wave equation becomes

$$p(\mathbf{x}, t) = \frac{s(t - \frac{\|\mathbf{x}_0\|}{c} + \phi(\mathbf{x}))}{4\pi\|\mathbf{x}_0\| - 4\pi\epsilon(\mathbf{x})} \quad (1.15)$$

where  $\phi(\mathbf{x}) = \frac{\epsilon(\mathbf{x})}{c}$  is the residual phase depending on the observation point. Knowing that  $\epsilon(\mathbf{x})$  is negligible with respect to  $\|\mathbf{x}_0\|$  the latter equation can be simplified as

$$p(\mathbf{x}, t) \propto s(T + \phi(\mathbf{x})). \quad (1.16)$$

where  $T = t - \frac{\|\mathbf{x}_0\|}{c}$ . Without loss of generality, the proportional gain can be set to 1. From far-field observation properties planar wave fronts have unique directions of propagation driven by a unit vector  $\mathbf{n}$ . As a consequence  $\mathbf{n} \cdot \mathbf{x} = 0$  defines a constant plane where the phase (or time) shift of the signal is the same. Hence (1.16) can be rewritten as

$$p(\mathbf{x}, t) = s(T + \frac{\mathbf{n} \cdot \mathbf{x}}{c}). \quad (1.17)$$

This result states that sound pressure depends only on the direction of propagation  $\mathbf{n}$  of wave fronts. This property in robot audition is generally assumed and widely used for sound localization process [NOK01]. In this case the sound source



direction can directly be obtained from the time shift from two different observations of the signal as illustrated in Figure 1.10 (see also epipolar geometry in Chapter 2). However as robots or sources locations are not supposed to be known, far-field conditions cannot be always ensured. As shown in [VMRL03] and in Figure 1.10, the accuracy of sound localization may drastically decrease as the source get closer to microphones, since the assumption of planar wave fronts does not hold anymore. Hence, some signal processing methods are specially designed to cope with near-field conditions such as localization techniques based on spherical harmonic for MUSIC or beamformer localization for near-field conditions respectively proposed in [KH16] and [ADS06], however with a higher computational cost.

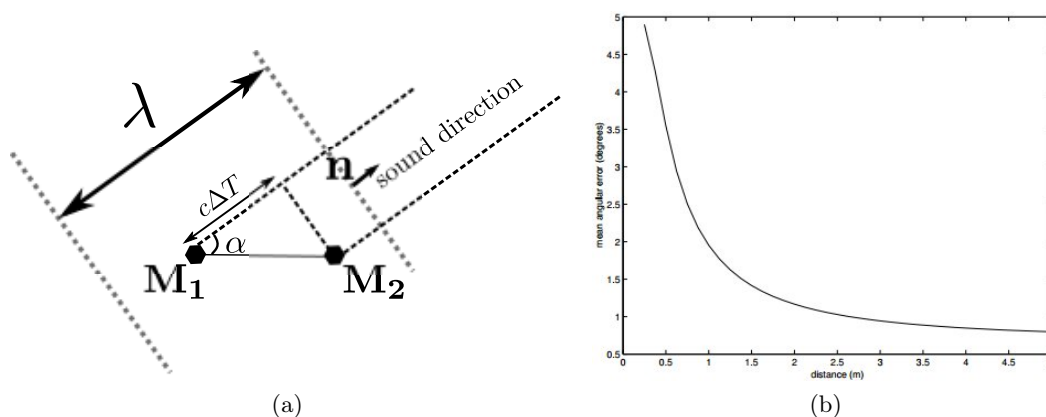


Figure 1.10: Far-field assumption for localization: In (a) the direction  $\alpha$  of the source is directly linked to the time difference of arrival  $\Delta T$  of sound between the microphones  $M_1$  and  $M_2$ . Note that the distance between the microphones should be lower than half of the wavelength to avoid  $2\pi$  ambiguities. However the results in (b) from [VMRL03] show that the localization accuracy decreases when the source get closer to the microphones and the far-field assumption is not ensured anymore.

Thus having a flexible approach that does not rely on any assumption on the observation field [AD07] and a limited computational cost is a crucial point for robot audition. Most robot audition solutions consider that the sources are located at more than 1.5 m (which approximates (1.13) for a 8-kHz sound and inter-microphones distance of 15 cm) from the microphones so that the far-field condition is ensured.

Nonetheless, the far-field condition is not always satisfactory for robot audition. Indeed, because of the sound decay characterized by the attenuation in (1.11) sound localization performance as well as speech recognition performance decreases when the source is too far from the microphones [Rod10]. This is particularly the case when considering indoor environments that are subject to reverberation.

### 1.2.2.2 Sound reflections

Until now, sound propagation has been modeled in a free space. For a more realistic modeling in indoor environment, propagation models should be formulated in



bounded areas. For this purpose, it is required to account for reflective spaces. Rigorously speaking, the following results are dedicated to reflection properties of sound at interfaces between two fluid media. These results to some extent are applicable to room acoustics, and describe quite well reflections of sound wave by walls. Reflections of sound on solids is a rather complex study, because of the variety of surfaces and the existence of several modes of wave propagation in solids. For simplicity, let us consider a planar scene implying a plane wave propagating in a medium 1 with a given density  $\rho_1$  and where the sound propagates at a speed  $c_1$ . The sound wave reaches a boundary that separates the medium 1 from the medium 2 that has a density  $\rho_2$ . The sound in the medium 2 has a speed  $c_2$ . The angles  $\theta_i$ ,  $\theta_r$  and  $\theta_t$  respectively correspond to the incident angles, the reflected and transmitted angles of the sound wave as depicted in Figure 1.11.

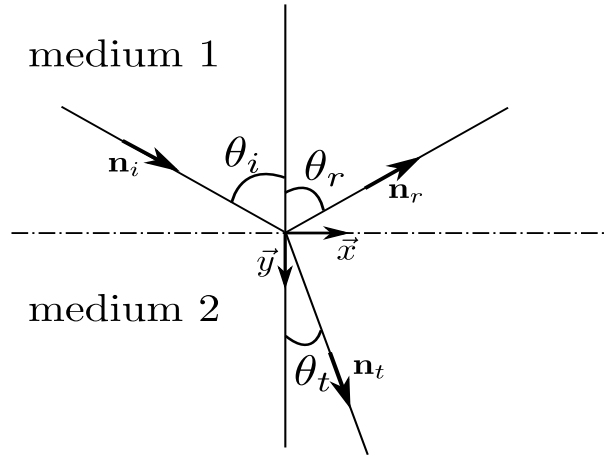


Figure 1.11: A sound reflection at a boundary: the incident sound wave propagating in a medium with  $\mathbf{n}_i$  direction and an angle  $\theta_i$  is reflected at the boundary with an angle  $\theta_r$ , while a part of the wave is transmitted to the second medium with an angle  $\theta_2$

Sound reflections are governed by the following equations

$$\frac{\sin \theta_i}{c_1} = \frac{\sin \theta_r}{c_1} = \frac{\sin \theta_t}{c_2}. \quad (1.18)$$

From this result it can be inferred that  $\theta_i = \theta_r$ , which corresponds to the reflection law derived from Fermat's principle in optics, stating that the incident angle of a wave is equal to the reflected angle. Similarly, from (1.18), the Snell-Descartes law can be obtained with  $\frac{\sin \theta_i}{c_1} = \frac{\sin \theta_t}{c_2}$ . The Snell-Descartes equation explains the well-known refraction property of sound that is characterized by a change of speed and orientation at the boundary of two media. Moreover the level of sound reflection depends on the boundary material. Indeed the coefficients of reflection  $R_p$  and transmission  $T_p$  are given by the following equations

$$T_p = \frac{2\rho_2 c_2 \cos \theta_i}{\rho_1 c_1 \theta_t + \rho_2 c_2 \theta_i} \quad (1.19)$$

and

$$R_p = \frac{\rho_2 c_2 \cos \theta_i - \rho_1 c_1 \theta_t}{\rho_1 c_1 \theta_t + \rho_2 c_2 \theta_i}. \quad (1.20)$$

These parameters characterize the level of pressure transmitted or reflected at the boundary of two media. The more similar the two media, the less reflection and the more transmission. This property is particularly used to adjust the acoustic of rooms, that is to say control the level of reverberation depending on the materials of walls. A hard material such as concrete is very dissimilar to the air through which the sound moves, subsequently, most of the sound wave is reflected by concrete walls, while fiberglass and acoustic tiles are more similar to air than concrete and thus have a greater ability to absorb sound. A standard measure of reverberation level in a room is the  $RT_{60}$  which refers to the time that takes the sound pressure level to decrease by 60 dB. Typical measurements of  $RT_{60}$  in indoor environments are given in Table 1.1

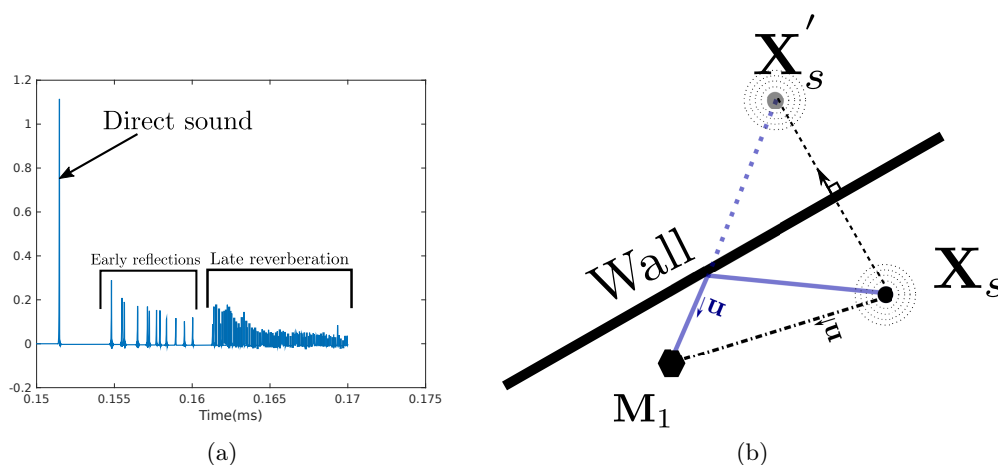


Figure 1.12: Sound reflections in indoor environment: in (a), the room impulse response characterizing the three modes of sound perception. In (b) the image source model [AB79], expressing that a reflection of a source  $X_s$  recorded by a microphone  $M_1$  can be represented by a symmetric virtual source  $X'_s$  emitting.

In robot audition as in many other tasks in audio processing, sound reflections represent a major issue. The issue lies in the fact that incident waves get reflected and add up to the actual sound wave. A geometrical way to represent reflections is to use image-source methods [AB79] by modeling the spherical waves generated by virtual point sources mirrored by an actual source, over each wall (or surface). Hence, as direct and reflected waves start hitting walls, reflections start adding up until they fade out due to sound pressure attenuation (see (1.11)) and walls transmission properties. This phenomenon can be observed in the room impulse response (RIR), that is, an acoustic observation of a generated impulse sound. RIR characterizes acoustically a given environment through the direct sound path, early reflections, and the late reverberation as illustrated in Figure 1.12. Thus after several reflections,

the sound pressure at the position  $\mathbf{x}$  is given by

$$p(\mathbf{x}, t) = \frac{s(t - \frac{\|\mathbf{x} - \mathbf{x}_0\|}{c})}{4\pi\|\mathbf{x} - \mathbf{x}_0\|} + \sum_{r=1}^{\infty} \frac{s(t - \frac{\|\mathbf{x} - \mathbf{x}_r\|}{c})}{4\pi\|\mathbf{x} - \mathbf{x}_r\|} \quad (1.21)$$

where  $\mathbf{x}_r$  denotes the virtual point sources. The superposition of all these waves has repercussions on sound perception. Reverberation corresponds to reminiscences of the actual signal attenuated with a phase shift. The sound waves generated by early echoes are highly correlated and interfere with the direct path signal. Sound waves interference greatly modifies the perceived sound with respect to the actual emitted sound. The observed sound pressure in  $\mathbf{x}$  might be greater, lower, or with the same pressure depending on interference phase shifts that condition the way the waves add up (i.e., constructively or destructively). Late reverberation, on the other hand, particularly degrades speech intelligibility. Indeed syllables that are crucial for sound recognition can be completely buried by a long and persistent reverberant field. For instance in Figure 1.13, it becomes computationally challenging to detect the different syllables because of the addition of a persistent sound field slowly decaying, smearing the envelope of the signal.

Location	Volume (m <sup>3</sup> )	RT <sub>60</sub> (s)
Recording studio	< 50	0.3
Classroom	< 200	0.4 - 0.6
Private Office	< 1000	0.6 - 0.8
Open Plan Office	< 1000	0.8 - 1.2
Lecture Hall	< 5000	1.0 - 1.5
Concert Hall, Opera	< 20000	1.4 - 2.0
Church	-	2 - 10

Table 1.1: Typical **RT**<sub>60</sub> in indoor environments from low (green) to high (red) reverberation time. This classification is based on human sensing. In robot audition 0.3 s is already considered as a moderate reverberation time while 0.8 s is high.

Furthermore localizing and separating sound sources become more challenging. The strong early reflections illustrated in Figure 1.12 are perceived as potential sources by the localization system. These reflections might even be more perceptible than the direct sound path, especially, when the robot is far from actual sources. Moreover interference might also decrease the pressure of actual sources. Similarly, it becomes more difficult to segregate the sound sources with respect of their location because of interference and the high correlation of early echoes. This motivated the topic of dereverberation that is widely used in speech recognition and by extension could be used in robot audition. The goal of dereverberation is to automatically remove the effect of reverberation from a recorded signal. Reverberation depends on the shape of the room but also on the positions of the actual sources, their directivity and eventually microphone(s) position(s). The solutions considered in signal processing literature are mainly based on this information and generally consist in an inverse

filtering of room acoustics (see [YSD<sup>+</sup>12]). For instance [GESS08] and [GNN13] use RIR in order to reduce late reverberation and enhance speech processing stage. A

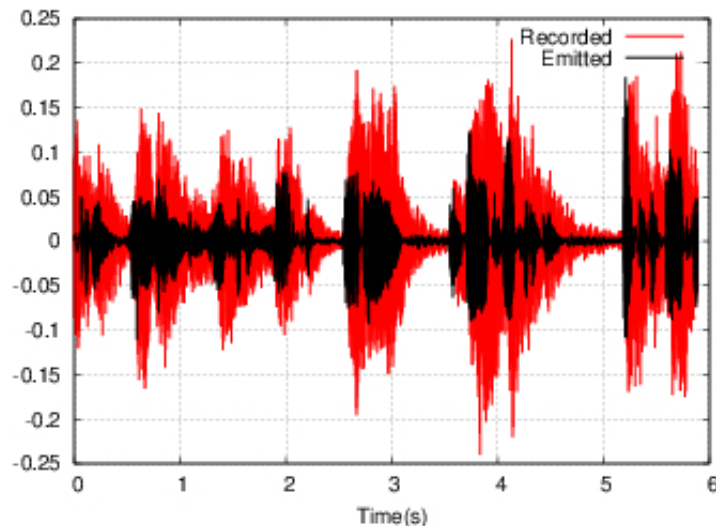


Figure 1.13: Reverberation effect on speech: At 3 m from the speaker, in a  $4m^2$  room with moderate reverberation time,  $RT60 \approx 400$  ms, the speech signal is completely smeared by reverberation which makes it computationally challenging to process.

more intuitive solution consists in positioning the robot close to the sound source of interest. Conceptually, it consists in solving the reverberation problem at the sound localization stage. However when facing real reverberant environments the accuracy of localization techniques remains mitigated. Localization under conditions of high reverberation still remains a challenging task and most of the techniques are very sensitive to variations of acoustic environments. For example, the authors of [VMR07] proposed to subtract reverberation spectrally similarly to noise spectral suppression techniques (see Section 1.2.3.2). This approach requires some tuning parameters that depends on room acoustics. Otherwise, the main path of improvements relies on enhancing the GCC angular spectrum with spectral weighting function such as the so-called PHase Transform (PHAT) [KMOO11, LLY11] that is known (heuristically) to be more robust to reverberation. Other methods try to benefit from the redundancy of measurements by increasing the number of microphones “physically” or virtually through the *active audition* framework [NOK00b].

## 1.2.3 Challenges raised by robotic context

### 1.2.3.1 Technical constraints

The first challenges raised by robotic context concern the technical and physical limitations inherent to this type of machines. These limitations are unique compared to other hearing machine systems. First, robots are machines that meet physical and geometrical constraints due to their structure. An auditory system should be embeddable on a robot considering space restriction. Space limitation and embeddability of

auditory systems are conflicting with the purpose of robustness and accuracy. Indeed hearing performance can be improved either by using a maximal number of microphones in order to increase the redundancy of the measurements or by maximizing the overall array aperture, namely the distance between microphones as stated in [Tre02]. A maximal distance between the microphones improves the accuracy of the localization since the estimated source direction resolution gets smaller as the spacing between the microphones increases. But on the other hand, as the spacing between the microphones increases, spectral aliasing is more likely to happen for high frequencies and the far-field zone moves away (see (1.13)). The far-field assumption usually admitted in sound localization becomes more difficult to ensure. From the latter results, finding the optimum positioning and topology of microphones builds up a thorough subject of research studied in antenna and microphone array literature. Some contributions focus on optimizing array configurations in order to increase the number of detectable source such as the solutions proposed in [PPL90]. In [ZB10], the authors introduced an optimal array configuration that improves the localization accuracy by maximizing the mutual information between the actual source location and the estimated one. These techniques are unfortunately not applicable on real robots due to the embeddability constraint among others. For instance, on humanoid robots the space between microphones does not exceed a few tens of centimeters in general. A tradeoff between performance, accuracy (*i.e.* far-field assumption) and embeddability of auditory systems is required: commonly, on robots the number of microphones is limited to eight, for microphone array systems, while binaural systems are limited to two microphones. Hence, in this case, robust algorithms should be developed with very few and close microphones.

Another aspect of the embeddability constraint concerns real-time requirements that imply restrictions on available computational resources and the maximum payload of sound processing algorithms. As a consequence auditory approaches for robots should be low demanding on resources, which limits their complexity and their computational cost. For instance, sound localization estimations must be available within a short time interval. Typically, up to 150 ms are acceptable for reflex tasks such as sensor-based control. Inverse methods developed in signal processing such as inverse boundary-element models [AT00], basically localization techniques relying on acoustic wave equations described in the previous section, are not suitable for robot audition because of their computational cost. Furthermore real-time processing implies to limit the signal duration from which the localization is performed. The signal duration is generally less than 100 ms for real-time localization. A short signal duration decreases localization performance as stated in [VVO04] for humans and in [MMWB15] for machines. Similar real-time constraints hold for speech processing algorithms in order to guarantee an appropriate social behavior during interactions. Fortunately, in this case, this constraint is generally fulfilled thanks to the recent progress and wide application of automatic speech recognition for machines (*e.g.*, mobile phones) that also requires real-time computations. The main challenges related to automatic speech recognition in robotics are about the perturbations caused by the environment, as discussed previously, and ego-noises, that are discussed in the next section.

At last, a fundamental difference between robot audition and classic machine hearing systems, is obviously motion: robots are moving and evolving autonomously in dynamic environments. Auditory techniques should then be able to perform dynamic perception and to cope with acoustic changes in the environment. These constraints are challenging to solve and until now only few works have tried to tackle this problem. The main obstacles come from auditory cues that are continuously changing during data acquisition. Consequently inconsistent cues are likely to be extracted and may degrade the hearing capabilities of the robot. Generally most approaches perceive sound in a static way as previously mentioned with CASA applications in robotics [NIYO15]. When robots are moving, the path taken by many approaches consists in the “stop-and-perceive” principle that suggests to stop the robot while acquiring data [NOK00b][TTLJ15]. This kind of approach adds latency and may prevent robots from receiving new commands. Recently in [TR15], the authors proposed a method to improve localization cues by compensating the motion of the robot. However this method assumes anechoic conditions. Some other approaches address this situation in an opposite way: a fixed auditory system with a moving sound source, which can be considered conceptually as the same problem. For localization, [PDA12] introduced a stochastic approach to cope with a moving sound source, that has been validated in anechoic conditions. Tracking systems based on particle filters as proposed in [VMR07] or probability hypothesis density trackers [EMN<sup>+</sup>15] are also alternative solutions to cope with moving sources. Similarly in [NNIH10] or [NNHT09] the authors proposed a sound source separation and recognition technique for moving sources based on an adaptive window size. Nevertheless, most of the aforementioned works are performed either in simulation or in ideal acoustic conditions (*i.e.*, anechoic or prepared room) and with an array of microphones. The degree of complexity raises to another scale when in addition to the motion of the robot, the source of interest is moving, as it may occur in realistic scenarios.

### 1.2.3.2 Ego-noise

Another challenge consists in dealing with ego-noises, that is to say noises caused internally by robots. Noises can be caused by several type of perturbations. One special type of ego-noise, which is observed while robots are performing actions using their motors, is called ego-motion noise. Indeed joint mechanical frictions, quite present in humanoid robots, or motors that control robot motions increase the level of noise in the recorded sound signal. A different type of noise is caused by the embedded electronic systems, that generally use a cooling system with loud fans generating noisy perturbation. Eventually noises can be caused by speech uttered by robots during barge-in situations: humans could freely speak during robots utterance requiring to "clean" recorded signal.

One can deal with ego-noises at different levels. First, they can be limited at a mechanical and topological level, which refers to robots design stage. Microphones can be positioned in such a way that they are far from potential sources of noise such as joints and cooling systems. Nonetheless, this solution is quite restrictive

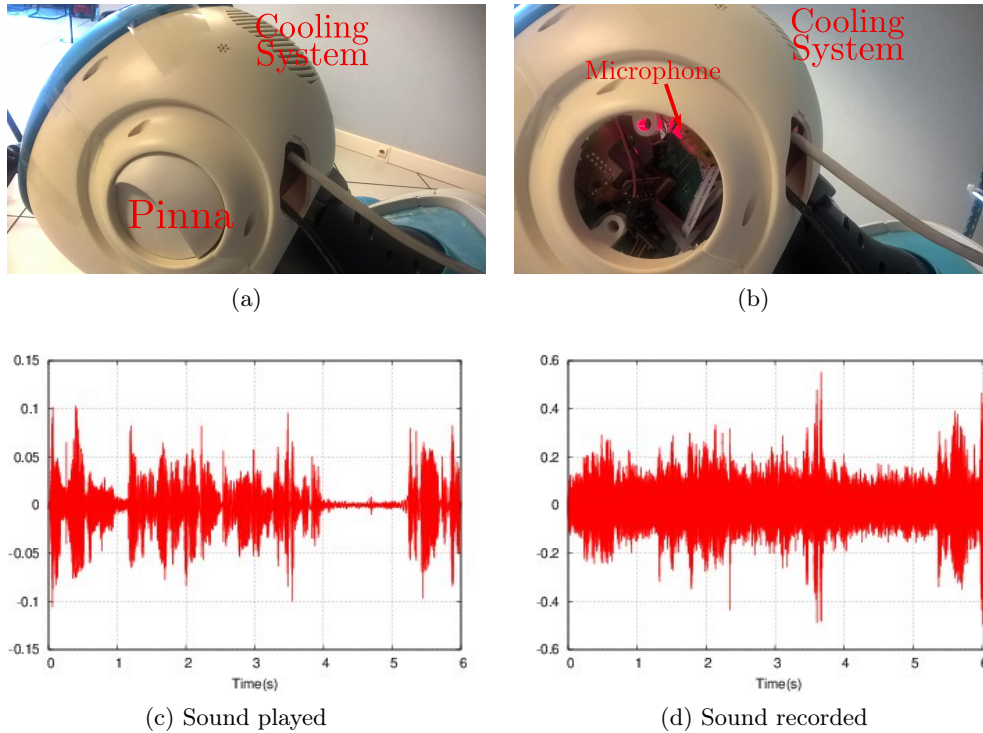


Figure 1.14: An example of ego-noise caused by an unsuitable microphone configuration: the sound played in (c) from a loudspeaker is severely altered by the cooling system noise as illustrated in (d).

and is limited by the type of robots considered. In this context robot manufacturers play a major role in robot audition development. Indeed as the interest for audition appeared quite lately in the timeline of robotics, the mechanical design of robots regarding the auditory system efficiency was not, until recently, a priority. Thus it is not surprising that many commercial robots do not embed auditory systems or that they possess auditory systems located at unsuitable positions (*e.g.*, near fans) that severely degrade sound quality as illustrated in Figure 1.14. Such noisy signals may seriously deteriorate localization and speech processing capabilities of robots. Fortunately, with the higher demand of social robots coupled with the increase of robot audition projects (see Section 1.3), specific robots designed for or supporting robot audition are arising. One can refer to the robotized head *POPeYE* [CBL<sup>+</sup>09] or the dummy head *KEMAR* [BDFP16] that are specially designed for robot audition research. Moreover, recent projects such as EARS, involving robot manufacturers (*i.e.*, Softbank Robotics) participate in raising the awareness about the importance of designing consistent microphones topology.

A second solution to deal with ego-noises is to use spectral subtraction algorithms that are particularly suitable for stationary noise. The denoising filters proposed in signal processing literature can be integrated in robots hearing process. This is the case of the well-known Ephraim-Malah filter [EM85] that consists in filtering



spectrally a noisy signal with respect to an estimate of the signal-to-noise ratio (SNR). When considering the Short Time Fourier Transform (STFT)  $X(k, f)$  of a noisy input signal  $x(t)$ , where  $f$  and  $k$  respectively stand for the frequency and the short-time frame, the Ephraim-Malah spectral gain  $G(k, f)$  is then defined as:

$$G(k, f) = \frac{\sqrt{\pi}}{2} \sqrt{\frac{1}{1 + SNR_{post}(k, f)} \frac{SNR_{prio}(k, f)}{1 + SNR_{prio}(k, f)}} \times \mathbf{B} \left[ (1 + SNR_{post}(k, f)) \frac{SNR_{prio}(k, f)}{1 + SNR_{prio}(k, f)} \right] \quad (1.22)$$

where the function  $\mathbf{B}$  is the function

$$\mathbf{B}(x) = e(-x/2) [(1 + x)I_0(x/2) + xI_1(x/2)] \quad (1.23)$$

in which  $I_0$  and  $I_1$  stand for the Bessel functions of zero and first order. Equation (1.22) depends on  $SNR_{prio}(k, f)$  and  $SNR_{post}(k, f)$  that are respectively the *a-posteriori* and *a-priori* SNR.  $SNR_{post}(k, f)$  is given by

$$SNR_{post}(k, f) = \frac{|X(k, f)|^2}{|N(f)|^2} - 1 \quad (1.24)$$

where  $|N(f)|^2$  denotes the noise spectrum.  $|N(f)|$  can be easily obtained from a sample of signal that contains only noise or in a more accurate way from the Minima-Controlled Recursive Average (MCRA) algorithm [CB02]. MCRA algorithm estimates the noise spectrum by averaging past spectral power values with respect to the speech presence probability. Thence,  $SNR_{post}(k, f)$  is an estimate of the SNR in the current short-time frame. On the other hand,  $SNR_{prio}(k, f)$  is a prediction of the ratio of clean signal power over noise power, that is recursively estimated from decision-directed approach [EM85] as

$$SNR_{prio}(k, f) = (1 - \alpha_a) \mathbf{P}[SNR_{post}(k, f)] + \alpha_a \frac{|G(k-1, f)X(k-1, f)|^2}{|N(f)|^2} \quad (1.25)$$

where  $\mathbf{P}(x) = x$  if  $x > 0$  and 0 otherwise, and  $0 < \alpha_a < 1$  is a smoothing parameter. Some other denoising filters exist in literature such as Wiener filters that have been used in robot audition in [VMR07] or [MP10]. In these papers, the spectral weighting  $G(k, f)$  function is simply defined as

$$G(k, f) = \frac{SNR_{prio}(k, f)}{1 + SNR_{prio}(k, f)}. \quad (1.26)$$

More recently in [GM15], the authors proposed to improve  $G(k, f)$  by considering (1.26) as a hard mask that is set to 0 or 1 depending on the value of  $SNR_{post}(k, f)$  with respect to a given threshold. In [GM16], they enhance the noise filter  $G(k, f)$  with a transition function that can cope with abrupt changes in broadband noise level by relying on the non-stationarity of speech signal compared to noise. Globally such techniques are particularly efficient to decrease noise caused by fans for instance.



Still the problem remains open for internal noises caused by robots motors or joint frictions. Unlike sounds emanating from cooling systems, the noise level emanating from motors is unpredictable, highly fluctuating and in some cases directional. For instance in UAVs, the sound produced by rotors and propellers is particularly loud and vary with the motion and speed of the vehicle. This problem has been particularly addressed in the context of sound source localization which performance is strongly degraded by noise level. In robot audition literature, a first type of approach proposes to perform sound separation by splitting internal sounds from external sounds. In [NOK00a] ego-motion noises are measured by an internal pair of microphones in order to assess the relevance of external auditory measurements. In [ESS+09] the authors use blind sources separation processes to split internal sounds from the external ones thanks to additional sensors that only measures internal noises. Nevertheless this kind of approach requires additional internal sensors, which is not always possible. And more importantly the efficiency of these methods depends on the strong assumption that the internal sensors measure only internal noises.

The second type of approach considers the mapping between robot motions and the generated noise. In [INR+09] and [INA+11], the method of mapping is based on template matching techniques with the idea that motor noises can be predicted, if the motion has a pattern of a prior known duration. Predicted noises are then subtracted from the measured signal. Similarly, [IKSM05] introduced an approach based on neural networks that predicts noise spectrum from angular velocities of robot joints, while in [NNN+06] ego-motion noises are subtracted using pre-recorded templates of more or less complex motions. Recently, the authors of [DK15] proposed to create a dictionary by combining a K-SVD algorithm [AEB06] to sources phase in order to extract a learned signal from a new input mixture. This approach has been applied to ego-noise reduction. The noise related to each robot motion could be learned and extracted afterwards from microphones recordings. These approaches are limited by the potentially high number of robot joints: for anthropomorphic robots, the prediction of ego-noises becomes particularly challenging. Generally ego-motion noise predictions are limited to few motions such as in [NNA+09], that focuses on noise caused by robot head rotation, or for UAVs as introduced in [FON+13]. The latter paper uses an adaptive estimation method of noise spectrum from motor control values.

## 1.3 Robot audition research

### 1.3.1 Research projects

Globally, the previous sections attested of the main difficulties of robot audition when evolving in real-world environments. To face these challenges, various projects were issued from collaborations of several research teams during the last decade. These fruitful collaborations impuled the path taken by robot audition community. In this section, we review these projects through their achievements.

### 1.3.1.1 Software and hardware solutions

To begin with, a description of software platforms available in robot audition is important. Currently there are two major pieces of software for robot audition that are detailed below.

**HARK**<sup>3</sup>: Based on their early contributions on robot audition for humanoid platform like *SIG*, the authors of [NMOT04, NOK00b] released an open-sourced robot audition software HARK (Honda Research Institute Japan Audition for Robots with Kyoto University). HARK is the result of a 5-year research project **Robot Audition from Computational Auditory Scene Analysis** started in 2007. As outlined by HARK structure, this project focused mainly on the functions to build up a CASA framework, that is to say, binaural [KMOO11] or array-based [ONOO11] localization, source separation [HYT<sup>+</sup>12] and speech processing [SKHO08]. HARK consists in a set of modules corresponding to the latter functionalities to construct a robot audition framework independently of microphones configuration or the type of robot used. Currently the project **Multiple Development of Robot Audition** aims to extend HARK features to cope with moving sound sources and/or robots [NNA<sup>+</sup>09] and with the development of robot audition on aerial vehicles [OYNN12]. More recently the project **Impact**, aims to extend applications of robot audition to the context of search-and-rescue missions. Advanced functionalities are expected to be implemented in the software. The details of the techniques and performances of sound source localization, separation or speech recognition are given in [NTO<sup>+</sup>10] or more recently in [ON15]. Globally, the aim of HARK is to provide a common platform for the robot audition community with high usability and real-time performance. The software is regularly updated and supports several libraries and tools such as ROS (Robot operating System), OpenCV, or Kinect.

**ManyEars**<sup>4</sup>: The second software, ManyEars project is an open GPL software principally based on the work published in [VMR07]. ManyEars [GLF<sup>+</sup>13] provides real-time sound source localization, separation and tracking using an array of eight microphones. ManyEars is implemented in C as a modular library, with no dependence on external libraries. Attached to this project, the 8SoundUSB hardware [AC<sup>+</sup>] provides an interesting sound acquisition system that consists in a dedicated sound card and eight microphones. This solution can be especially useful for mobile robots, since only little hardware dedicated to robot audition exists. Commercially available sound cards and microphones are not designed to be embedded on robots and may integrate functionalities that are not useful for robot audition (*e.g.*, mixing, sound effects...). Eventually, similarly to HARK, the software is also compatible with ROS.

### 1.3.1.2 Collaborative projects

Research projects concerning robot audition have significantly increased in the last few years, showing the growing importance of this topic in the scientific community.

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<sup>3</sup><http://www.hark.jp>

<sup>4</sup><https://sourceforge.net/projects/manyears>

Some of them are described in [ON15] and [APB<sup>+</sup>13]. We review below the main projects related to robot audition that we are aware of.

**POP:** POP (Perception on Purpose) was a 3-year scientific project that lasted from 2006 to 2009. The project concentrated onto the understanding and modeling of the interactions between an agent and its physical environment. These interactions were modeled from auditory and visual cues, with a biological and computational perspective. The main topic concerned the models of cognitive mechanism of attention, the extraction of auditory and visual cues as well as their integration and the study of coordination between sensor observations and motor activities. Although the studies mentioned in the project were not especially designed for a robotic context, a binaural robotized test platform, *POPeye* [CBL<sup>+</sup>09], has been developed in order to validate the theoretical findings of the aforementioned research. From the robot audition viewpoint, this project developed novel algorithms for real-time localization of sound sources [CMWB09] based on pitch and interaural cues. This project also thoroughly studied active perception from which one can improve sound localization performance [CLLH07] by turning robots towards source direction, or recover the distance information [LC08]. Furthermore the database CAVA<sup>5</sup> (Computational Audio-Visual Analysis) was released in order to develop and evaluate computational methods of audio-visual scene analysis. Eventually, POP involved partners that are Inria Grenoble, more particularly *Perception* team, University of Sheffield, through the *Speech and hearing* group, University of Osnabrück with *Institute of Cognitive Science* group, University Hospital Hamburg-Eppendorf *Institute of Neurophysiology and Pathophysiology* and the *Institute for Systems and Robotics* of Coimbra University.

**HUMAVIPS:** In the continuity of POP, HUMAVIPS (Humanoids with Auditory and Visual Abilities In Populated Spaces) was also a 3-year project started in 2010. HUMAVIPS was particularly centered on audio-visual perception for Human-robot interaction and humanoid behavior. The objective of HUMAVIPS was to endow humanoid robots with audiovisual abilities that is to say exploring populated spaces, localizing people and possibly engaging interaction with one or two people with an appropriate behavior. The main idea of this project was to build-up tools in order to solve the *cocktail party problem* with the help of the visual sense. This project led to contributions such as sound event recognition [JAPGH12] or categorization [APSRH13], source localization and separation through learning methods [DH12] or methods of sound localization based on geometrical time-delay [APH12]. Similarly to POP, a database of binaural recordings, CAMIL<sup>6</sup> (Computational Audio-Motor Integration through Learning), provides real-word recordings of various type of sources (speech, music, noise...) and configurations (still or moving recording system). This project involved *Perception* and *Mitis* team of INRIA Grenoble, Aldebaran Robotics (currently known as Softbank Robotics), the *Research Institute for Cognition and Robotics* at Bielefeld University, IDIAP Research Institute and the *Center for Machine Perception* at Czech Technical University.

<sup>5</sup>[http://perception.inrialpes.fr/CAVA\\_Dataset/Site/](http://perception.inrialpes.fr/CAVA_Dataset/Site/)

<sup>6</sup>[http://perception.inrialpes.fr/~Deleforge/CAMIL\\_Dataset](http://perception.inrialpes.fr/~Deleforge/CAMIL_Dataset)

**BINAHR**: BINAHR (Binaural Active Audition for Humanoid Robots) was established from 2009 to 2013 as a French–Japanese collaboration focused on active binaural audition. The project aimed to endow humanoid robots with auditory capabilities inspired from human skills. This project investigated the relationship between motion and auditory perception through *active audition* with contribution such as [PDA12] or sensorimotor theories for auditory perception in [BPDCG12]. Multi-party communication was also a topic of interest of the project with high level functions such as speaker identification [YBA<sup>+</sup>11], speech recognition with audiovisual activity detection [YAZ12a], attention mechanisms [YTK<sup>+</sup>11] [HTOO11], source separation [NTOO12], ego-noise and reverberation cancellation [TNT<sup>+</sup>12]. BINAHR also included software developments with several of the aforementioned contributions which were incorporated in HARK, and hardware developments with a design of an artificial binaural sensor based on EAR sensor, a fully programmable eight-channel data-acquisition board [LMD12]. Finally this collaboration involved LAAS, ISIR and LPP from French side, while Kyoto University, Tokyo Institute of Technology (TITECH) and Kumamoto University were involved in Japan.

**TWO!EARS**<sup>7</sup>: This project started in 2013 and is ending by 2016. The aim of this research project is to develop a computational framework for modeling active human listening that enables the analysis of auditory scenes. The main difference of this project with respect to usual auditory frameworks lies in the two-way process of the modeling. A bottom-up framework as described in Figure 1.5 is used for acoustic scene understanding as well as a top-down process where hypotheses from the acoustic scene are integrated in order to improve the elementary functions of sound localization or separation [SWK<sup>+</sup>14]. Although this project is devoted to the wide field of machine hearing, several works have been validated on robotic platforms. This approach is used for instance in [PBDM14] that uses prior independent learning of interaural transfer functions and environment noise statistics to develop a sound source localization method. *Active audition* is also a major topic of this project with contributions that use motion to improve sound source localization [MMB15] or use feedback loop [BDFP16] to minimize the localization uncertainty. Learning methods are also explored with deep neural networks for localization [MBM15] or speech recognition [MMBB15]. As a result of this project, an auditory framework was released as a GNU software that includes several stages for modeling active human listening (localization, separation, speech processing...) coupled with a blackboard architecture that allows the inclusion of world knowledge, as well as an acoustical simulation framework for creating binaural signals. TWO!EARS includes several partners such as LAAS, ISIR, the University of Sheffield, through the *Speech and hearing* group, the Institute of Communication Acoustics, Ruhr University Bochum, the Department of Electrical Engineering–Hearing Systems from Technical University of Denmark and others.

**EARS**<sup>8</sup>: EARS (Embodied Audition for RobotS) that will last until 2017 is centered on human-robot interaction in adverse environments. A technical solution should be embedded on a commercial robot (NAO from Softbank robotics). This

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<sup>7</sup><http://twoears.aipa.tu-berlin.de>

<sup>8</sup><https://robot-ears.eu>

solution will be evaluated through a final demonstration in a scenario of a welcoming robot at a reception desk. For this perspective, EARS focus on designing an efficient and embeddable microphone array [TR14] and sound localization techniques devoted to this configuration like [EMN14] or [MEN<sup>+</sup>15]. A part of the project is also devoted to scene analysis through audio and visual modalities via [GAPFH16] or [GBEH15]. For robustness and high applicability, EARS project is based on array-based processing with a particular focus on dynamic scenes that cover localization of moving sound sources [EMN<sup>+</sup>15] and sound separation of moving sources [KBGAP<sup>+</sup>15] or localization while the robot is moving [TR15]. The project involves an industrial partner (SoftBank Robotics), and scientific partners including *Perception* from INRIA Grenoble, the *Speech and Audio Processing Group* of Imperial College London, the *Multimedia Communications and Signal Processing* group from Friedrich-Alexander-University, the *Acoustics Lab* from Ben-Gurion University and *Adaptive Systems* of Humboldt University of Berlin.

### 1.3.2 Conferences and organized sessions

Since 2000s, the growing interest of the robotic community in robot audition has also been reflected in the increasing space given to audition in international conferences. These conferences are opportunities to raise awareness of robot audition issues and by the same mean to increase the interest towards such a challenging topic.

First of all, among robotics conferences, IROS (International Conference on Intelligent Robots and Systems) and ICRA (International Conference on Robotics and Automation) are probably the biggest roboticists meetings and give a relevant overview of the current state-of-the-art of robotics in general. The first sessions about robot audition appeared in IROS 2003. Until 2013, sessions about Robot Audition were organized at the instigation of K. Nakadai and H. Okuno. Nowadays, since 2014, "Robot Audition" is registered in the keyword list of research topics of RAS (Robotic and Automation Society), and regular sessions on the topic are organized at IROS each year. IROS is probably the conference that gathers the most contributions related to robot audition each year. For instance in IROS 2015, 26 papers with the keyword "Robot Audition" were submitted. In comparison with computer vision field which account for 78 papers, this number is relatively low but is nevertheless in the head of list of frequent keywords. Among these papers, 14 papers were accepted and presented at the conference. Moreover special sessions are regularly organized at IROS in order to synthesize the main progress in the field and set roadmaps and direction of future research. For IROS 2016, the special session "New Horizon for Robot Audition Application" organized by K. Nakadai, H. Okuno and others is centered on applications based on robot audition technologies, and innovative technologies towards such applications. Regarding ICRA, 2015 edition saw the inclusion of the keyword "Robot Audition", while this keyword was surprisingly missing for ICRA 2016 edition. Until now, there is no regular session devoted to robot audition in this conference. As a result, generally few papers concerning robot audition are submitted to ICRA. For instance in ICRA 2016, only 4 papers related to robot audition were presented, dispatched in diverse sessions. Robot audition contri-

Contributions can also be found in different robotic conferences. One can cite RO-MAN (International Symposium on Robot and Human Interactive Communication), RO-BIO (International Conference on Robotics and Biomimetics) or HRI (International Human-Robot Interaction Conference), where robot audition applications are used in the context of the latter conferences, that is to say for interactions or as cognitive processes.

In signal processing, ICASSP (International Conference on Acoustics Speech and Signal Processing) one of the biggest conference in signal processing sometimes holds special sessions about robot audition. In the last decade, a handful of robot audition sessions were organized, such as "Signal Processing Techniques and Algorithms on Robot Audition" organized by A. Sugiyama and H. Okuno in 2009 or "Audio for Robots – Robots for Audio" in 2015 organized by E. Vincent and J. Le Roux. However most of the contributions in ICASSP are generally oriented towards theoretical aspects of signal processing unlike robotic conferences that encourage applications. Targeting this conference is also a good way to encourage signal processing community to consider robotic context for their developments or at least real-world applications. In parallel other conferences related to signal processing involve the participation of the robot audition community. For instance EUSIPCO 2015 (European Signal Processing Conference) held a tutorial about EARS project, organized by H. Löllmann, C. Evers and R. Horaud. One can also cite INTERSPEECH (Conference of the International Speech Communication Association) or ICA (International Congress on Acoustics) that are however more centered on computational machine understanding of auditory perception.

## 1.4 Conclusion

To sum up this introductory chapter, robot audition represents a promising path of improvement of robots interaction capabilities. Hearing capabilities complement vision and touch senses for robots evolving in populated spaces. Thus, endowing robots with a reliable hearing sense is crucial for scene analysis and would allow robots to apprehend more comfortably realistic environments. The main needs in robot audition concern sound localization, sound separation and speech processing. These functionalities build up most of the applications in robot audition. Robot audition can be used in a wide variety of applications such as human-robot interaction, navigation, search-and-rescue missions or mapping without restrictions on the type of robot (humanoids, UAV, mobile robots...). Despite this tremendous potential, there is little research in robot audition in comparison to the huge amount of research in robot vision, for instance.

The explanation lies in the fact that robotic audition is extremely complex and limited, especially in real-world environments. First some limitations come from the inherent properties of sound waves, that induce interference from echoes or modeling constraints on the field of observation of sound. Moreover, environments may be dynamic and include several types of sound sources and noise, which raise the degree of complexity of sound processing. Eventually the requirements induced by

the robotic context, such as real-time constraints end up making robot audition a very challenging topic. Modeling accurately all potential perturbations and environment variability turns out to be complicated, and many realistic situations remain unsolved until now. As a result, most of the models and techniques proposed in robot audition literature consider static scenes, that are either validated in simulation or experimentally in controlled conditions such as anechoic rooms.

To tackle robot audition limits, sound source localization, among other auditory functions, is certainly the subject that receives most attention. The ability to localize sound sources is a crucial aspect of auditory perception: it creates a sense of space for listeners, and provides information about the spatial location of objects in the environment. And more importantly, from a computational perspective, sound localization is the first stage of auditory perception and undeniably represents one of the pillar of robot audition. The stage of sound localization generally conditions the good progress of auditory interactions, since sound separation and speech processing rely or can be improved by the initial localization step. Thus, by strengthening and ensuring accurate sound localization abilities, the interaction process is greatly simplified, which increases the robustness of robot audition applications, whether it be for interaction, navigation or scene mapping. Consequently, solving robot audition limitations, comes back in a loose sense to solve the problem of sound source localization in realistic environments.

## Chapter 2

# Sound source localization for robotics

This chapter introduces the issues of sound localization in the robotic context. As underlined by the previous chapter, one of the toughest challenge for machine hearing consists in finding from where the sound source originates. This task is particularly complex in realistic environments, inducing interference from echoes and fluctuating acoustic conditions. The focus on this particular topic is easily explained by the fact that most of the interactions and studies of human auditory sense, as illustrated by the *cocktail party effect*, largely inspired robot audition.

The current chapter reviews, first, human hearing ability in Section 2.1, through the anatomy of auditory systems and the physiological hypotheses that explain hearing process. More specifically, we focus here on the mechanisms involved in the sound localization process through computational models and the related auditory cues.

In a second phase, we focus on how the knowledge of sound localization mechanisms is transferred to the robotic context. In detail, Section 2.2 introduces the low-level auditory cues and their processing for machine hearing and by extension for robot audition.

The last segment of this chapter expounded in Section 2.3 reviews current progress and limitations of sound source localization in robotics. More precisely, we focus on the binaural paradigms that are in general inspired by physiological knowledge about mammals auditory systems.

## 2.1 Human auditory system as a source of inspiration

### 2.1.1 Human auditory system

Similarly to many aspects of human sensing skills, human auditory system is a relevant illustration of ability for sound localization and separation. For the past centuries, the human auditory system has been a major topic of research and inspiration for psycho-acousticians. An initial stage for building an artificial auditory system generally requires a thorough understanding of the biological structure of hearing organs and the neural activities associated to auditory sensation.



As described in Chapter 1.2.1, the sound is a wave propagating in a medium. This wave when reaching the ears should be interpreted by the auditory system into exploitable and understandable information. The sensation of sound is caused by varying pressures in ears. Due to the airborne nature of sound, auditory perception is a range sensing that played a crucial role during species evolution. Hearing organs for airborne sounds probably appear around 250 millions years ago [SC09], during the Triassic era that corresponds to a period of transition from water to land. On land, hearing organs evolved separately among species and mammals were endowed with tympanic ears that provided the more “recent” sensing skill. The sense of hearing stimulated afterwards the rise of communication and languages. Until now, for many mammal species, sound localization skill is essential for detecting preys or conversely predators when vision is limited by a restrained field of view, obstacles or obscurity. Then, it is not surprising that our ancestors developed sensitive tympanic ears and highly specialized auditory organs capable of processing airborne sounds.

#### 2.1.1.1 Mechanical structure

The auditory system transforms sound waves into neural activity. For this purpose, the auditory system is physically composed of three parts that are respectively the outer, the middle and the inner ear illustrated in Figure 2.1. The outer ear, composed of the pinna and the ear canal, principally amplifies sound waves before transmitting it to the middle ear. The elastic cartilage that builds up pinnae is involved in the sound localization process with complex spectral filtering patterns governed by pinnae shape as stated in [Str07] (see Section 2.1.2). The ear canal that has a resonating conic shape transmits a filtered sound to the tympanic cavity through the tympanic membrane. Due to resonance in the auditory canal the sound wave pressure is increased before reaching the tympanic cavity [Wie47]. In the middle ear, the sound pressure is therefore transformed into vibrations by the eardrum (*i.e.*, tympanic membrane) and transmitted to the auditory ossicles. The auditory ossicles consist in three small bones that are the malleus, the incus and the stapes. As sound waves vibrate the eardrum, it in turn moves the nearest ossicle, the malleus, to which it is attached. The malleus then transmits these vibrations, via the incus, to the stapes, and ultimately to the membrane of the oval window, on the outside of the cochlea.

The cochlea that builds up the inner ear is filled with a salty watery liquid, the perilymph, which moves in response to vibrations coming from the middle ear via the oval window. The cochlea is sealed by the round window membrane, which vibrates with opposite phase to vibrations entering to the inner ear. Mechanical amplification of sound waves through the ossicles is essential since it allows to improve the transmission of sound waves from air to liquid medium. It is said that without ossicles, only 3% of sound waves would be transmitted to the cochlea [HWA08]. It is not surprising that damaged or absent ossicles lead to a moderate-to-severe hearing loss [AGF+99]. Thus, the presence of the ossicles to concentrate vibration intensity improves the sensitivity to sound. Cochlea exact functioning is still a discussed subject. A first mechanical model of cochlea was proposed in 1863 by von Helmholtz,

while the current mechanical model of cochlea is defined in [VBW60] one century later. Cochlea consists in a coiled tube enclosed in a hard bony shell split into two canals by the basilar membrane as illustrated in Figure 2.2. These two canals, respectively named vestibular and tympanic canals, compress and dictate the motion of perilymph thanks to alternate vibrations of the oval and round windows. Basilar membrane model is equivalent to a set of harmonic oscillators tuned to different frequencies from 20 Hz at the apex to 20 kHz [MOD89]. Basilar membrane resonates accordingly to the signal frequency thanks to different stiffness from its base to its apex that is more flexible. By combining this property to the inertial effects of perilymph, that transmits a phase delay of a few cycles to the basilar membrane oscillation, the cochlea acts as a filterbank which transforms vibrations into a neural spike train. The spike train has the same phase as sound vibration [Wev49]. Inside the basilar membrane, the organ of Corti filters sound. Indeed the basilar membrane motion stimulates hair cells contained in the organ of Corti. The 15000 hair cells that are contained in the organ of Corti are the hearing receptor cells that are sometimes compared to photo-receptor cells of eye retina. These hair cells are from two types: outer cells play a role of amplification, while inner cells transform vibration into electric signals that are transmitted to auditory nerves. The information passed to the brain is represented as a cochleagram that is equivalent to a spectrogram (see Section 2.2.2). Yet, the crucial point of the auditory system remains the interpreta-

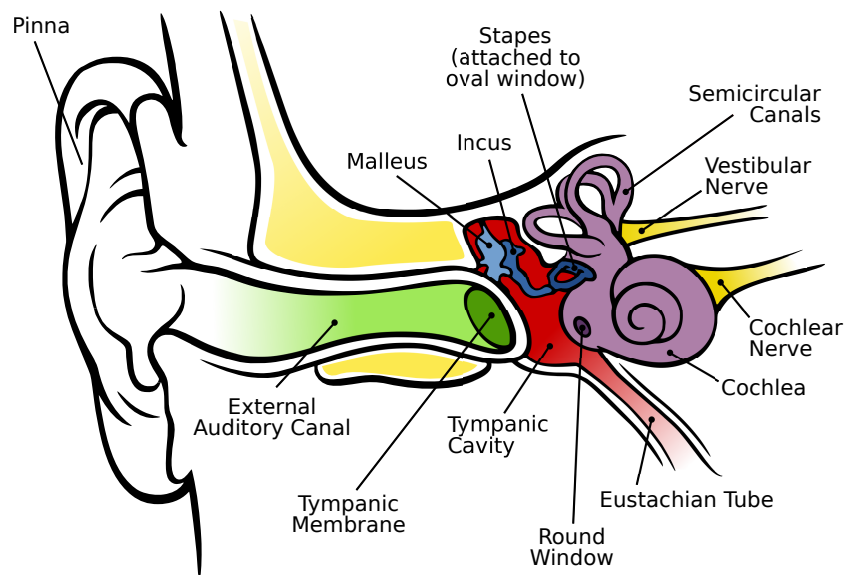


Figure 2.1: Anatomy of the human auditory system as represented in [CB05]

tion of this neural activity. Actually, the neural activity travels from cochlea to the auditory cortex where signals are decoded and interpreted into an auditory sensation (localization, sound identification...).

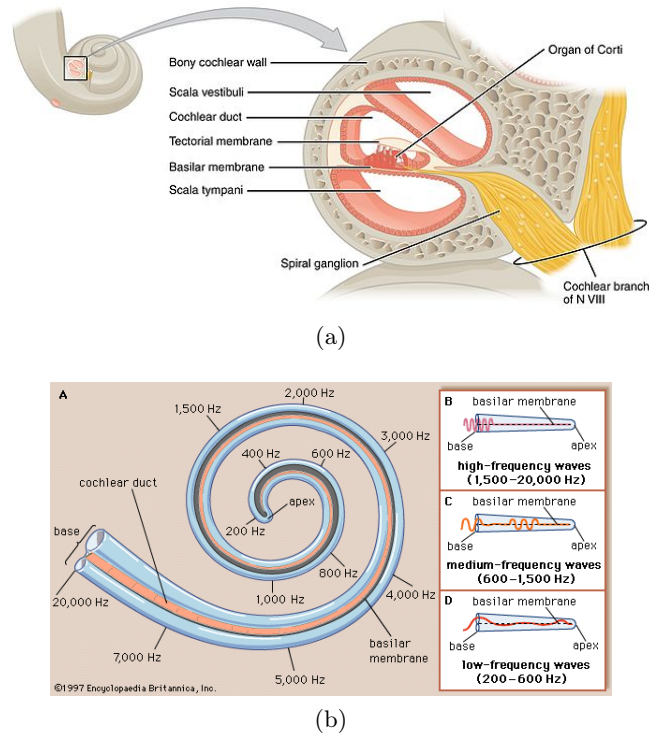


Figure 2.2: The cochlea: In (a) Cochlea transverse view [B<sup>+</sup>16] and (b) basilar membrane functioning (www.britannica.com)

### 2.1.1.2 Neuronal path

The brain has to solve the difficult task of transforming acoustic wave inputs into meaningful auditory sensation. As already depicted in Figure 2.1 the sound signal travels from the ear to the auditory cortex. The auditory information is then processed in the midbrain, that is divided into several nuclei. The superior olivary complex (SOC) is the where station that receives input from both ears through connections with cochleas. The SOC is the very early stage of signal interpretation since interaural comparisons are processed [TY02], which is an important source of information to determine the sound location. The SOC is divided into two nuclei that are respectively the lateral superior olive (LSO) and the medial superior olive (MSO). The MSO seems dedicated to process the time difference between sound waves as exposed in [JSY98], while the LSO is more devoted to the difference of level observed between sound waves [JY95]. As detailed in the next section, these two types of information are used for localization in the azimuth plane. This information coupled with the elevation information induced by pinna is afterwards integrated to the Inferior colliculus (IC) where a 3D spatial representation of source location can be inferred [CFC02]. The IC is also connected to the superior colliculus and transmits localization information to an attention mechanism [ZVVO04].

Eventually this information is transferred to the medial geniculate nucleus, before reaching the corresponding auditory cortex. In the auditory cortex, the spatial layout

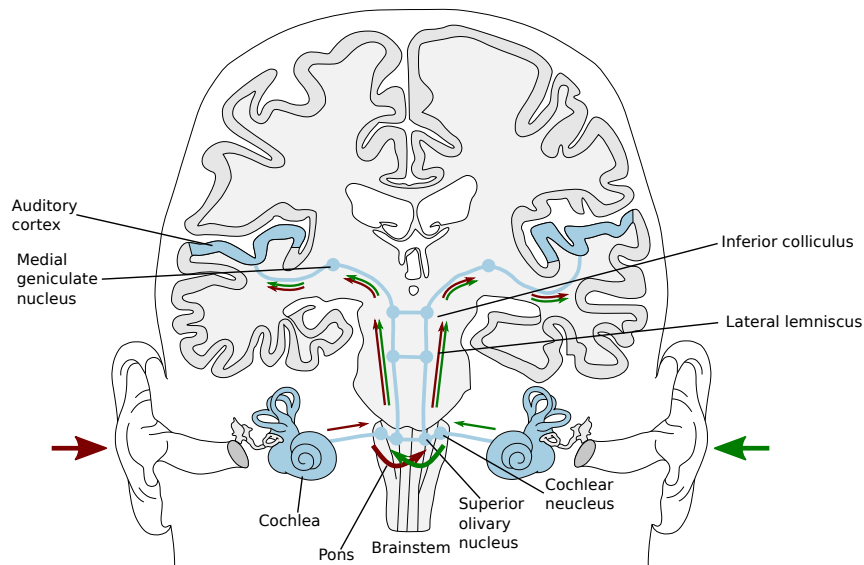


Figure 2.3: Neuronal path of sound from ears to the auditory cortex. Figure adapted from <http://twoears.aipa.tu-berlin.de>

of frequencies in the cochlea along the basilar membrane is repeated through what is called the tonotopic organization [TSM<sup>+</sup>04].

Globally this neural scheme stresses a central process of sound waves where the localization step seems to be performed by a spectral interaural comparison. This interaural comparison coupled with the time-frequency decomposition are the core ideas behind sound localization in robotics.

## 2.1.2 Physiological mechanisms of sound localization

### 2.1.2.1 Binaural cues

Now, one can focus on the localization process that is performed in the SOC. In the case of sound source localization, interaural cues have been identified as carriers of spatial information. These assumptions have been formalized through the *Duplex Theory* in [Str07] under the assumption of a perfectly spherical head without any external ears (pinnae). This theory explains many properties of sound localization process for humans. The identified interaural cues are respectively the interaural time difference (ITD) and the interaural level difference (ILD) that are respectively generated by the delay and attenuation of a given sound wave between the two ears (see Figure 2.4). Both ILDs and ITDs are related to the sound direction of arrival in the azimuth plane. ITDs and ILDs are complementary cues that are relevant in different ranges of frequencies, as illustrated in Figure 2.5. Actually, ILDs are significant for frequencies above approximately 1.5 kHz, that is to say when the head is large compared to the wavelength  $\lambda$  of an incoming sound. In this case, the head shadows most of the incoming waves, by the reflection mechanisms discussed in the previous chapter. Consequently the perceived sound level is drastically attenuated

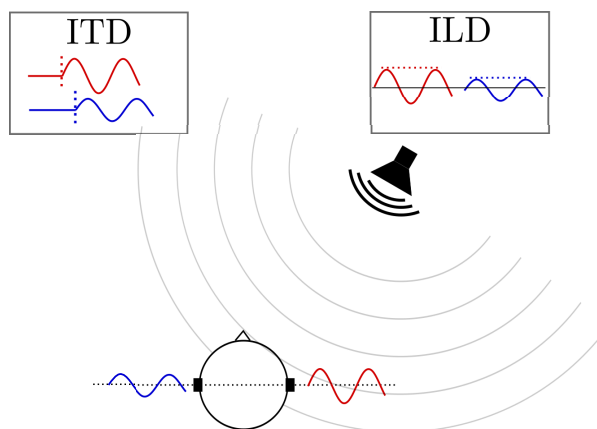


Figure 2.4: The ILD is caused by the difference of sound pressure between the two ears while the ITD reflects the phase difference between the signal perceived by these ears

for the contralateral microphone compared to the ipsilateral one. Conversely ITDs are meaningful for low frequencies. ITDs can be interpreted unambiguously only for frequencies for which the maximum physically-possible ITD is less than half of the wavelength  $\lambda$ , which corresponds somehow to the Nyquist frequency. Otherwise ITDs are subject to spatial aliasing related to the phase ambiguity. For a typical human head size (*i.e.*  $\approx 22$  cm between each ear) the maximum ITD value is about  $660\mu\text{s}$ , which sets the frequency limit of meaningful ITDs to around 1.5 kHz. However, the frequency ranges for which an auditory system is sensitive to ITDs and ILDs significantly overlap each other, and most useful sound signals are wideband including low and high frequency components. Auditory systems in most cases combine information from both ITDs and ILDs to extract the location of a sound source [Bla97].

Several neuro-computational models have been proposed in psychophysics literature to explain how the brain extracts interaural cues. The time-delay neural network of the Jeffress architecture [Jef48] is still at the core of current models about how humans perform binaural localization using ITD cues. This theory characterizes the MSO neurons functioning as a coincidence mechanism. Specifically, a coincidence mechanism consists of a number of central neural units that record coincidences in neural firings from two peripheral auditory-nerve fibers, one from each ear, with the same frequency. Furthermore, neural signals from the two ears are delayed by given amounts fixed for each pair of fiber, similarly to the architecture depicted in Figure 2.6. Because of the frequency synchronization, a given binaural coincidence-output unit at a particular frequency is maximized when the ITD at that frequency is exactly compensated by the internal delay of the fiber pair. As a result, the ITD of a sound stimulus can be determined from the delay that has the greatest output over a range of frequencies. The short-term average of a set of coincidence outputs at a given frequency with respect to their internal delay corresponds to some extent to a short-time cross-correlation function of the neural signals arriving from the left and

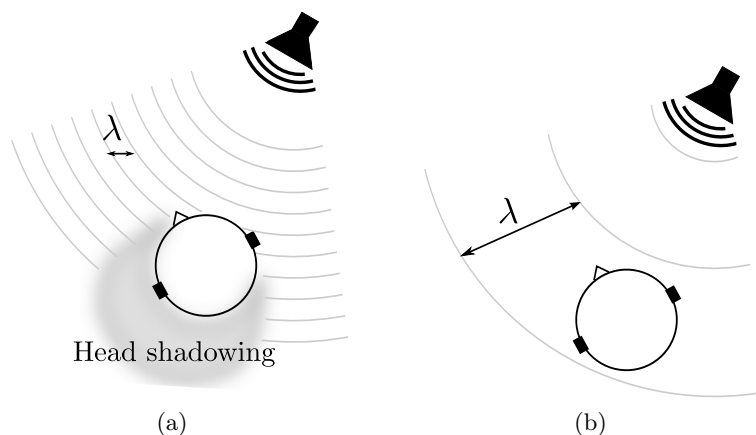


Figure 2.5: The ILD cue is consistent at high frequencies (a) since the head shadowing effect induces significant sound level difference. On the other hand, ITDs suffer from aliasing since small wavelengths  $\lambda$  induce phase ambiguity. At low frequencies (b), the aliasing is suppressed but the head shadowing effect is not significant.

right ears. The utilization of such a mechanism has been validated for owls auditory system in [CK88]. It has been found that MSO neurons of owls are connected in a very similar way to Jeffress diagram. The neurons communicating signals from the two ears converge in an orderly way, by increasing the delay in the transmission length. Furthermore, to account for the influence of ILDs in the latter model, Jeffress proposed to complete the neural mechanism with the latency hypothesis which is based on observations that neural responses to more intense sounds tend to start more rapidly than responses to less intense sounds. This implies that ILDs are somehow converted to ITDs. Nevertheless, it has been suggested recently in [MG03] that mammals auditory systems do not fully comply with the Jeffress model. The authors of the latter paper exhibit evidence of neural broad sensitivity to just two ranges of ITDs, that could subsequently describe a continuous space of possible ITD values. Likewise, a number of researchers have also reported cells that appear to respond to ILDs at several levels of the brainstem [BT68, CK83]. These researches suggest that ILDs could be detected by a unit which has an excitatory input from one ear and an inhibitory input from the other. Despite these new results, the Jeffress model still remains acknowledged, and several auditory models are inspired from this pioneering work. This model stresses cross-correlation mechanisms between two neural inputs that is underlaid by the coincidence principle. In [Che61], the authors proposed a cross-correlation method directly based on the sound signal input rather than the neural input, in order to relate the interaural cross-correlation to a sound location. Since the cross-correlation function is mainly designed to extract ITDs, they considered an additional mechanism to account for ILDs, through the ratio of energy of the signals measured from the left and right ears. Some other auditory models are oriented towards the neural response to the sound signal. This is the case of [Col73, Col77] in which the author characterizes a binaural localization through a model comparing auditory-nerve responses. The latter model has also been extended

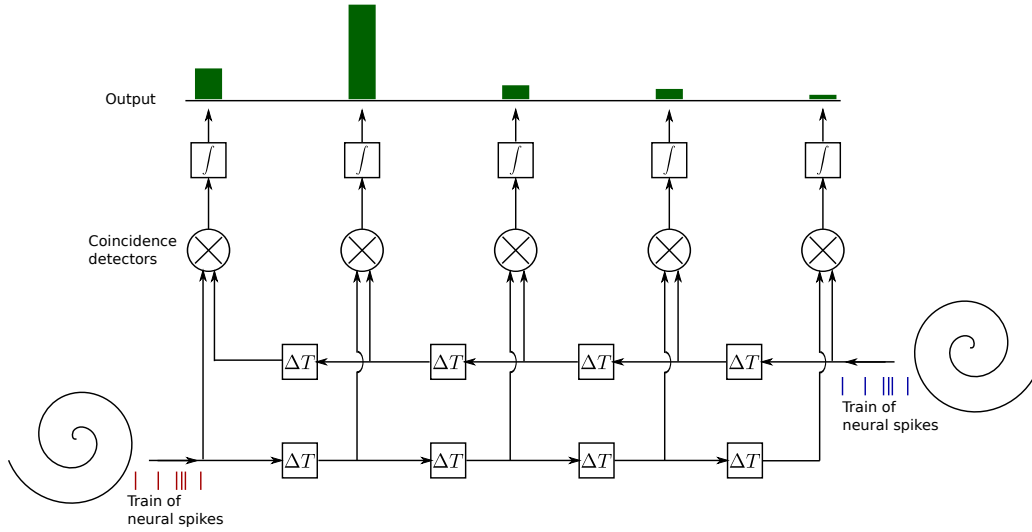


Figure 2.6: The Jeffress neuronal architecture

in [SJC78] to include explicit assumptions concerning time-intensity interactions, a mechanism for extracting subjective lateral position. In the continuity of these works, the authors of [Bla80, Bla97, Bod93] proposed more complex binaural hearing models able to cope with real environments (*e.g.*, *cocktail party effect*).

Unfortunately, despite this variety of hearing models, even the most complex neuro-computational one cannot fully explain the performance and accuracy of the human auditory system. This fact strengthens the idea of interaction between several cues and mechanisms for sophisticated localization processes. This idea is confirmed by various experimental studies [WK89, SN36, WAKW93] that stress significant errors of localization (in static configuration) when a listener is asked to indicate if a source stimulus is produced from the front or the back side. This is due the so-called *front-back ambiguity* that is inherent to binaural cues. *Front-back ambiguity* may be considered as the main limitation of interaural cues. This ambiguity implicitly confirms that there exists multiple source locations that share the same ITD and ILD cues.

As expounded in [vHW20, Bla97], in their simplest forms, iso-ITD locations tailor a hyperbolic surface symmetric to the interaural axis. In the far-field (*i.e.*, distances greater than 1 m for a typical head), the hyperbolic surface approximates a cone centered on the interaural axis, that shapes the well-known *cone of confusion*. Empirical measurements of the ITD with respect to source direction also confirm the cone-shaped iso-ITD surface [Mil72]. On the other hand, in the near-field low frequency ILDs are constant on a sphere centered on the interaural axis [SCSK00], while these ILDs become perceptually insignificant as the source gets farther from the head. Consequently, positions that give rise to the same binaural cues (*i.e.*, constant ITD and ILD contours) in the near-field form a torus that has been described in [SCSK00] as the *torus of confusion*. The *torus of confusion* degenerates into the *cone of confusion* for sources in the far-field as the influence of low-frequency ILDs





Figure 2.7: In the far field, localization based on interaural cues lead to a cone of confusion. In the near field the cone of confusion degenerates into a torus of confusion (black donut) principally because of the action of ILDs (*i.e.* *intersection of the cone and a sphere*).

is diminished. This confusion can be solved by the motion of the head [Wal40] (see Section 2.1.2.3).

### 2.1.2.2 Monaural cues

As a consequence of the previous results, it can be inferred that interaural cues do not fully describe the localization process. Some complementary cues are necessary to disambiguate source positions on the cone of confusion. Actually the localization process is also described by other dimensions that are the elevation and the distance. Several circumstantial results tend to relate sound localization to an action of pinnae. As an illustration, experimental results exhibited in [OP86] showed that localization performance is substantially degraded when performed from a single ear. More specifically, most of the azimuth angle were erroneous while elevation angles remained consistent, showing that some localization ability is somewhat preserved in the monaural case. After some training azimuth angles could be recovered ( however without the same accuracy as with a binaural auditory system). This typical experiment emphasizes mechanisms of localization implied by monaural cues that concerns elevation.

The following attempt [Bat67] aiming to characterize the effect of pinna filtering action, revealed that sounds reaching the ears interact with the physical characteristics of the listener, that is to say, head, shoulders and upper torso. The modifications induced by these interactions can be subsequently used to estimate the directions and distance to the sound source. All together, these interactions characterize a complex response function known as the head-related transfer function (HRTF). Analogously, in the temporal domain, HRTFs corresponds to the head-related impulse response



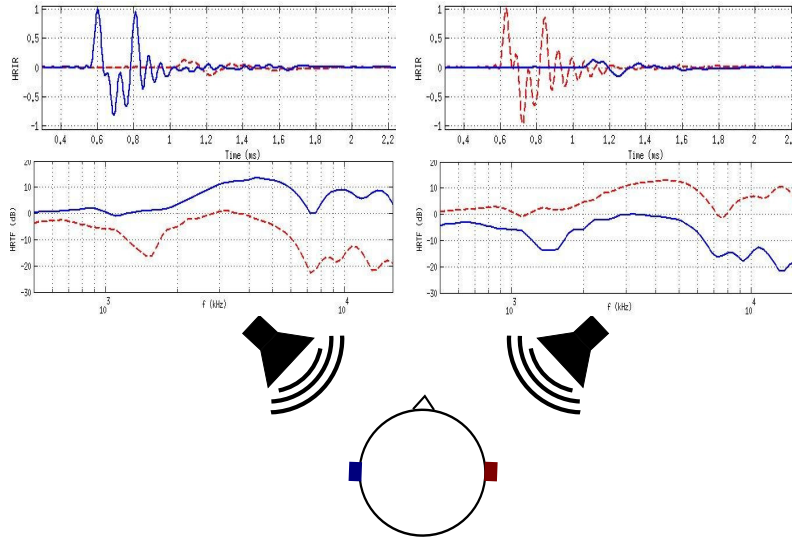


Figure 2.8: The HRIR (top), as well as the HRTF (bottom) captures the relationships between a signal measurement in a free space and the signal perceived by one of the ear. HRTFs (respectively HRIRs) depend on the source location and frequency. Data plotted from the CIPIC database [ADTA01].

(HRIRs) (see Figure 2.8). HRTFs have been proven to subsume sound localization cues (*i.e.*, ITDs, ILDs) and the sound spectral shape reaching a listener [Har99]. Indeed sound needs to propagate around the head (diffraction) in order to reach the ears. HRTFs establish the relationships between a sound measurement from a point in the free-field area and a sound measurement at the ears [WK89, Bla97]. For instance in [WK89] the authors used probe microphones in ears to measure and describe the transfer function from sound source to eardrums in anechoic environments. Hence HRTFs vary significantly between different individuals due to anatomy differences. Such influence characterizes the individual variability of sound perception as stated in [Bla97]. An HRTF itself can be decomposed into two separate components: the common transfer function (CTF), which is common to all sound source locations [MG90] (*i.e.* depending on the anatomy of the listener) and the directional transfer function (DTF), which is related to sound source direction. Explicitly, when a sound source is situated in the near-field, HRTFs are simultaneously depending on the direction and the distance to this source across all frequencies [BR99]. In the far-field the dependence of HRTFs on distance vanishes. As a result, the HRTFs to each ear are then a function of four variables: the sound frequency, the sound source azimuth and elevation angles and the distance to the source as stated in [ZDD04].

With the use of HRTFs, many of the localization limitations (*e.g.*, *cones of confusion*) inherent to models based on ITDs and ILDs alone can be overcome. Although elevation mechanisms have received less attention than localization in the horizontal plane (*i.e.*, through interaural cues), the hypothesis that elevation is recovered through spectral filtering actions of pinnae is now acknowledged [MOD89, PB93]. This filtering is especially useful to determine the source location in the vertical plane,

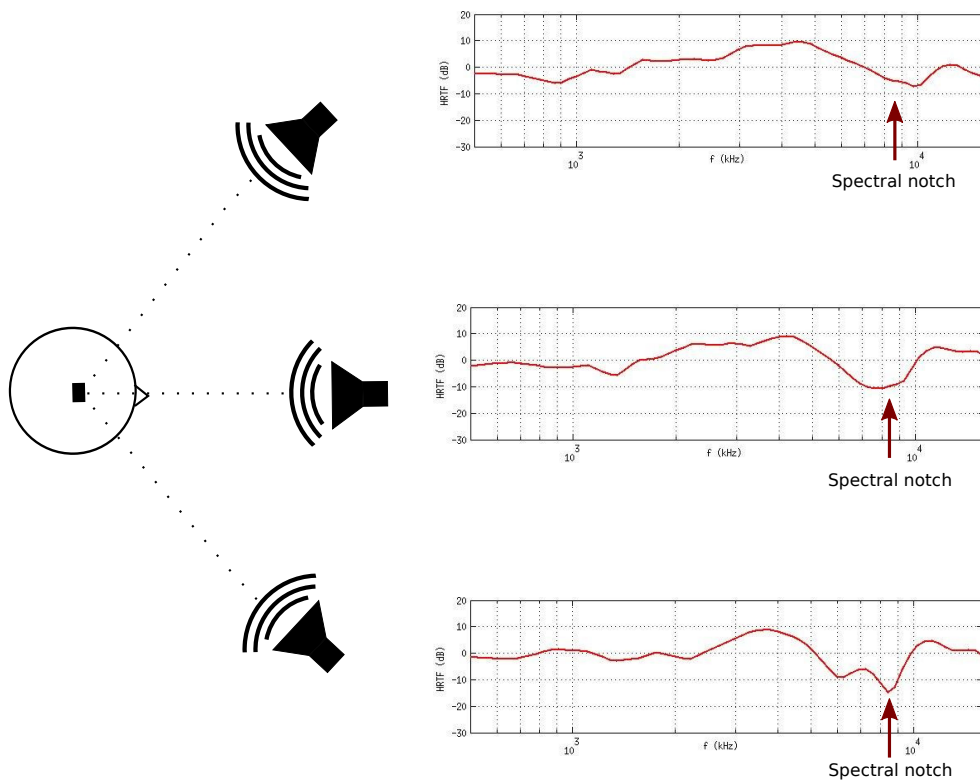


Figure 2.9: The spectral notches in an HRTF, are related to sound source elevation. Data plotted from the Cipic database [ADTA01]

and the resolution of *front-back ambiguities*. In the first case, there are substantial psychoacoustical results to support the hypothesis that pinna spectral notches are important cues for determining source elevation [MOD89, WHW74, TY03, RDY05]. In [CP94] it has been shown that the locations of spectral notches near 7 kHz change as a function of elevation, as illustrated in Figure 2.9. Moreover, the asymmetry of pinnae could help to distinguish sources from the back or front side. On the basis of the reflections of pinnae various ridge and different filtering effects are observed depending on which side the sound source is located [MG91]. This selective sound shadow in resolving the *front-back ambiguity* problem was confirmed in [FF68, PN97b]. However, because of the high variability of size and shape of pinnae between listeners, the task of relating the anatomy of pinnae to localization cues they create remains elusive. As a consequence, it is not yet possible to infer the sound location (azimuth and elevation angles) from unknown HRTFs.

Eventually distance estimation is processed from monaural cues. Naturally, sound intensity is used for estimating distances, particularly when the sound stimulus originates from a well-known source. Indeed as already mentioned in the previous chapter, the typical sound decay certainly provides the most intuitive cue to estimate distances. Several studies [War68, SG62] showed experimentally the correlation between the perceived distance and source loudness, that is perceptually related to

sound intensity. In enclosed rooms, the distance to a sound source can also be estimated from reverberation. The ratio of the direct sound over the first reflections, also known as Direct-to-reverberant ratio (DRR) or the time gap between the direct sound and the first reflections are assumed to be cues for distance estimation [She82]. The effect of reverberation has also been denoted in [Nie92] where distance judgments were much more veracious under reverberant conditions than anechoic conditions. We refer the interested reader to a more exhaustive review provided in [ZBB05] addressing distance cues.

### 2.1.2.3 Influence of the motion in the localization process

Chapter 1 emphasized the constraints related to robot motion, that requires to be able to process sound dynamically and in real-time. However a question arises to know if the motion is an actual constraint or conversely an advantage in the localization process. Thus in the following, we study evidence related to the influence of motion on human localization. The individual variability of auditory cues [Bla97] reveals that they do not explain solely the performance of the human auditory system. More importantly, it suggests that other mechanisms are involved in the perception of sound source. Thus some other modalities must be taken into consideration. Among them, self-motion plays an important role for the correct sensing of sound events. Indeed, in daily sound listening, our heads are constantly and unconsciously moving, which tends to emphasize the potential influence of motion.

From this perspective, almost all studies agreed on the fact that localization through motion performs better than localization in a static way. In [Wal40], it has been shown that the sound localization accuracy is improved when listeners are allowed to move their heads during the sound stimuli. A prior work [Wal39] also demonstrated that for longer sound duration, head motion cues dominate other cues. Some other studies like [TMR67] suggest that the sound localization accuracy is improved by head motions. Compared to static cases, they report an improvement of 10 to 15%, especially when the sound stimulus is long enough or when the source is located in eccentric positions.

A first thought could assume that motion multiplies the auditory cues measurements and builds up a spatial representation of the source location by fusing these measurements. As a consequence localization should be more robust thanks to the redundant auditory measurements. However it appears from the different experimental studies exposed below that the variation of auditory cues, more particularly the variation of interaural cues, plays an important role for the localization performance.

Actually, motion contributes to the disambiguation of the *front-back confusion* [WK99] inherent to the interaural cues. Indeed the variation of ILDs or ITDs with respect to the head motion, characterizes the fact that the sound source is in front or in the back of the listener. Moreover, the variation of interaural cues also plays a role for estimating sound source elevation. When the head is rotated, ITDs and ILDs change differently depending on the sound elevation [PN97a]. More specifically, the latter study showed that localization within *cones of confusion* remained possible despite distorted HRTFs, thanks to head movements. Likewise, head mo-

tions (translations) induce change in azimuth that is range dependent. For sources that are very close, a small shift causes a large change in azimuth, while for sources that are far, the azimuth variation is minimal [ADN95]. In the same way, [LHC90] suggested that distances could be perceived dynamically through motion. Actually, when the sound level (loudness) varies with varying distance, the apparent distance judgments of a listener change accordingly. In anechoic environments, this relationship is characterized by the inverse-square law: sound level falls by more or less 6 dB for each doubling of source distance [Col63]. There are also many other ways in which motion can help in audition: moving towards a source can increase the DRR (*i.e.*, increase the speech intelligibility as stated in Chapter 1.2.2.2) while head rotations can position the target source in the frontal plane where spatial resolution is at the greatest [CLH97] compared to peripheral locations. The latter property is often related to the *auditory fovea*, by reference to *visual fovea* at the retina center that is capable of higher resolution than in the periphery.

These studies emphasize the strong link between motion and auditory cues. Similarly, a study of [VVGVO04] involving a moving sound source, showed that acoustic inputs are continuously combined with feedback signals about changes in the head orientation. The motion is not pre-programmed as it could be assumed but it is rather dynamically adjusted by the variation of auditory cues. These results support the theory of a real-time feedback loop between motion and auditory cues. Moreover, interaural cues variation seems to be less sensitive to listener's HRTFs as exposed by [KUKH03]. The authors suggest that the information induced by head motions is useful for overcoming changes in listener's HRTFs. Globally, auditory scene analysis appears to be a dynamic process that is influenced by an active sensing of the environment as stated by [KPTK12]. The auditory process can then be considered as a sensorimotor loop where motion and auditory perception are interwoven.

## 2.2 Localization cues for binaural robot audition

### 2.2.1 Modeling the spectral information

The main cues used for machine audition are derived from the knowledge of mammals auditory system. It is not surprising that most contributions in binaural robot audition are based on interaural cues. A first step before extracting interaural cues deals with the representation of sound signals. The signal can be naturally exploited in the time domain, but for more efficiency and flexibility, its spectral decomposition is preferred. This way to process the sound draws closer to auditory mechanisms in the ear and cochlea. The simplest way to transform the temporal signal spectrally is by a Fourier transform. However, since sound signals used in robot audition are generally broadband and non-stationary, the short-time fourier transform (STFT) provides more relevant information. Simply, the function to be transformed is multiplied by a window function which is nonzero for only a short period of time. The Fourier transform of the resulting signal is taken as the window slid along the time axis, resulting in a two-dimensional representation of the signal. Considering two signals  $x_1(t)$  and  $x_2(t)$  respectively sensed by two microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$ , the

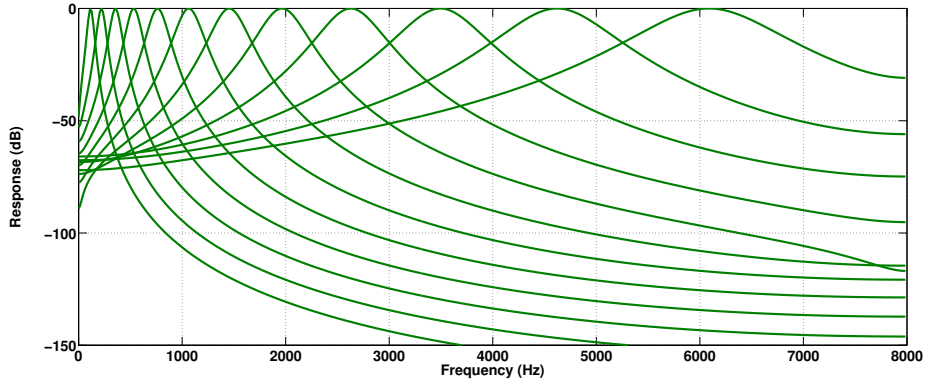


Figure 2.10: Gammatone filters frequency response based on Lyon's cochlear model [Eli09]

STFT  $X_i(f, T)$  at a given frequency  $f$  of the signal  $x_i(t)$  is simply defined as

$$X_i(f, T) = \int_{-\infty}^{+\infty} x_i(t)w(t - T)e^{-2j\pi ft} dt, \quad (2.1)$$

in which  $w(t - T)$  stands for a window function (*e.g.*, Hamming, Hann, sinusoidal...) centered on  $T$  sliding along the time axis. For discrete-time signals, as generally processed in machines, we assume that the following indices  $t = 1, \dots, T$  and  $f = 1, \dots, F$  are respectively the time frame and frequency bin indices. With these notations, (2.1) becomes

$$X_i(f, l) = \sum_{t=1}^T x_i(t)w(l - t)e^{-2j\pi ft/N}, \quad (2.2)$$

where  $N$  is the window length, and  $l = 1, \dots, L$  characterizes the frame that is processed<sup>1</sup>. The STFT transforms the sound signal into a spectrogram, that represents all spectral components of the measured sound (see Figure 2.11b). Nevertheless, methods based on STFTs do not perform a frequency decomposition similar to human inner ears. Actually as mentioned in the previous section, the signal is broken up into different frequencies that are naturally selected by the cochlea and hair cells. A frequency decomposition trying to mimic the effect of the basilar membrane inside the cochlea can be obtained from a gammatone filterbank [Joh72]. In this case, each filter is associated to a specific area of the basilar membrane. Gammatone filters are nowadays widely used to model cochlea filtering action as illustrated in Figure 2.11. A gammatone filter impulse response comes as the product of a gamma function and a tone. The output from this filter can be regarded as a measurement of the basilar membrane displacement. Gammatone filters constitute a good approximation of human spectral analysis at moderate sound levels [PAG95]. Their bandwidth increases with frequency (see Figure 2.10), mimicking the human auditory ability to

<sup>1</sup>For simplicity we use the notation  $x(t)$  instead of the  $x[t]$  for discrete signals

discriminate low-frequencies signals [SVN37]. It has also been shown in [PRH<sup>+</sup>92] that the impulse responses of gammatone filters fit to human auditory filter shapes, derived in [PM86].

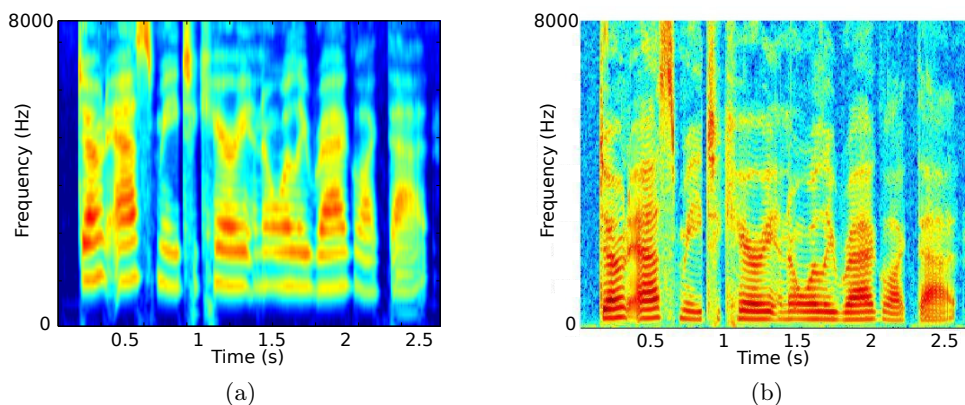


Figure 2.11: Two different representations of the spectral components of a sound: in (a) the cochleagram based on gammatone filterbanks and in (b) the spectrogram obtained by the STFT.

Mathematically, a gammatone filter is defined by its impulse response given by

$$g(t) = At^{n-1}e^{-2\pi Bt} \cos 2\pi ft + \phi \quad (2.3)$$

where  $A$  is the amplitude,  $n$  the order of the filter and  $B$  is the bandwidth of the filter that determines the duration of the impulse response. The filter bandwidth  $B$  is usually set according to an equivalent rectangular bandwidth (ERB) given by:

$$ERB(f) = \left[ \left( \frac{f}{Q} \right)^p + (B_{min})^p \right]^{\frac{1}{p}}, \quad (2.4)$$

in which  $Q$  is the asymptotic filter quality at high frequencies and  $B_{min}$  is the minimum bandwidth for low frequency channels. The parameters of this ERB are generally adjusted from human auditory data. Hence in literature, there exists several models such as the Glasberg and Moore cochlea model summarized in [GM90] ( $Q = 9.26$ ,  $B_{min} = 24.7$ ,  $p = 1$ ), the Lyon's cochlea model given in [Sla88] ( $Q = 8$ ,  $B_{min} = 125$ ,  $p = 2$ ) or the Greenwood model [Gre90] ( $Q = 7.23$ ,  $B_{min} = 22.85$ ,  $p = 1$ ). Eventually the filter bandwidth in (2.3) is approximated as  $B = 1.019ERB(f)$ . Hence a discrete-time cochleagram is obtained as

$$C_i(f, l) = \sum_{t=1}^{t=N} x_{ci}(t, f)w(l - t) \quad (2.5)$$

with  $x_{ci}(t, f)$  defined as the gammatone-filtered signal. With this model, cochleagrams can provide better results when extracting auditory cues as demonstrated in [SM15, YAZ12b].



### 2.2.2 HRTFs models

Similarly to human listeners, HRTFs also characterize robotic platforms, that can shadow, reflect or refract sounds. Modeling HRTFs is a big concern in robot audition since the relationship between the sound direction and the interaural cues is modified in a non linear way. The induced scattering and sound spectral modifications particularly affect the localization performance. Indeed the filtering effect of the robot structure is reflected in the sound level attenuation and delays. This effect is particularly stressed by anthropomorphic robots that are endowed with a head, a torso and pinnae mimicking human morphology.



Figure 2.12: Typical material required for HRTF measurements [ZX14]. In this solution the measurements are performed from two rings of speakers that rotate according to azimuth and elevation angles.

Hence, the signals  $x_1(t)$  and  $x_2(t)$  measured by the microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$  are defined spectrally by

$$\begin{cases} X_1(f) = H_1(f)S(f) \\ X_2(f) = H_2(f)S(f), \end{cases} \quad (2.6)$$

where  $X_i(f)$  corresponds to the Fourier transform of the measured signal,  $H_i(f)$  the corresponding HRTF and  $S(f)$  the Fourier transform at a reference point of the signal  $s(t)$ . However, it should be specified that HRTFs only account for the action of the robot structure (or body) on the sound in free-field. Hence, HRTFs do not model acoustic effect such as reverberation. In reverberant conditions, (2.6) becomes more exactly

$$\begin{cases} X_1(f) = H_1(f)H_{1_{room}}(f)S(f) \\ X_2(f) = H_2(f)H_{2_{room}}(f)S(f), \end{cases} \quad (2.7)$$

where  $H_{i_{room}}(f)$  characterizes the room acoustic effect such as reverberation. In this case  $H_i(f)H_{i_{room}}(f)$  is the acoustic transfer function (TF) characterizing the difference between the sound sensed at a reference point and the sound at each microphone.

There exists several ways to estimate or measure HRTFs. The straightforward way consists in obtaining HRTFs from direct acoustic measurements of a given sound signal at different locations and frequencies. The typical HRTF identification procedure consists in placing an artificial humanoid head and/or torso on a motorized turntable which can be rotated accurately to any required azimuth. A speaker mounted on a boom stand enables an accurate positioning of the speaker to any elevation with respect to the humanoid. This solution is, however, not so convenient, since it requires anechoic conditions. Furthermore this specific solution generally requires expensive equipment to cover all the possible location of the sound source (see Figure 2.12) and a long acquisition time. Fortunately, in literature, several HRTF databases are available such as the CIPIC [ADTA01] and [WGS11] that are related to the dummy head *KEMAR*, or [TAM+02]. This solution provides accurate HRTFs but requires the same robot, with the same structure (*e.g.* presence of torso, pinna...) as the one used for creating the database.

Nonetheless the exact HRTFs of robots are not always accessible, due to their complex structure or the tedious procedure of HRTFs measurements. As an alternative to the measurement process, a second solution consists in accurately modeling body and head effects. When the robot has a simple shape, HRTFs can be expressed through simple models such as the spherical model proposed in [Bla97]. For more realistic models, acoustic simulation software might be used to estimate HRTFs through signal processing methods [OI06]. In robot audition, simplified models of the head have been proposed in the literature. The auditory epipolar geometry (AEG) model [NOK00b] is the simplest one and is adapted to free-field microphones in a planar scene. In this model interaural cues are directly related to the sound azimuth direction since HRTFs are given by

$$\begin{cases} H_1(f, \alpha) = 1 \\ H_2(f, \alpha) = e^{-2j\pi f \frac{d}{c} \cos \alpha} \end{cases} \quad (2.8)$$

in which  $d$  is the inter-microphones distance,  $c$  the sound celerity and  $\alpha$  the sound azimuth with respect to the interaural axis, in the far-field. With  $\mathbf{M}_1$  as an arbitrary reference point, the auditory epipolar geometry states that signal measured by  $\mathbf{M}_2$ ,  $x_2(t + \tau)$  is delayed by  $\tau = \frac{d}{c} \cos \alpha$ , with respect to the reference signal  $x_1(t)$ . This model is by far the most used in sound source localization because of its simplicity, and the low computational load that it requires for inferring the sound location. However this model assumes free-field installations and is generally used for array-based localization (see Section 2.3).

The revised auditory epipolar (RAEG) [NOK01] is an extension of the AEG by considering, this time, a disk between the microphones. The disk models the diffraction induced by the robot head before reaching the microphones. In this case, HRTFs are characterized by the following set of equations derived from the



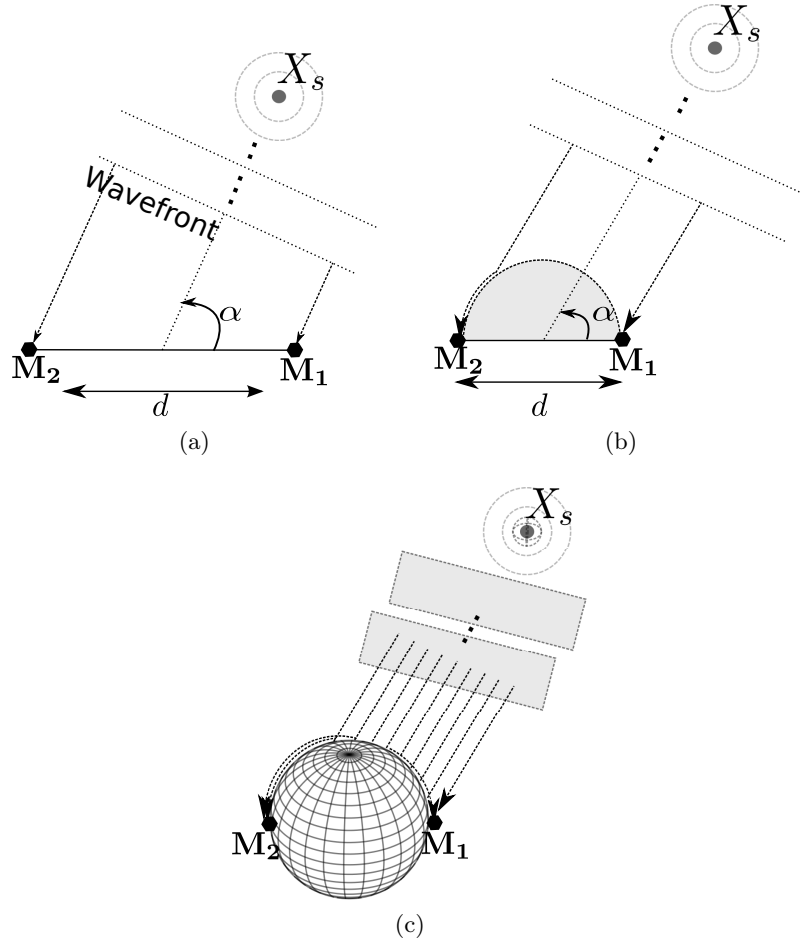


Figure 2.13: Head models characterizing HRTFs: In (a) the auditory epipolar geometry considering free-field microphones, in (b) the revised auditory epipolar geometry considering a disk between the microphones and in (c) the scattering theory modeling a sphere between the microphones.

Woodworth-Schlosberg approximation:

$$\begin{cases} H_1(f, \alpha) = 1 \\ H_2(f, \alpha) = e^{-2j\pi f \frac{F(\alpha)}{c}} \end{cases} \quad (2.9)$$

in which the head shadowing effect is modeled by the function:

$$F(\alpha) = \begin{cases} (\alpha - \frac{\pi}{2})\frac{d}{2} + \frac{d}{2} \cos \alpha & \text{when } 0 \leq \alpha \leq \frac{\pi}{2} \\ (\alpha - \frac{\pi}{2})\frac{d}{2} - \frac{d}{2} \cos \alpha & \text{when } \frac{\pi}{2} \leq \alpha \leq \pi \end{cases} \quad (2.10)$$

Through (2.10),  $F()$  characterizes the sound path that follows the circular shape of the head (*i.e.*, diffraction) in order to reach the microphones as illustrated in Figure 2.13.

Eventually, the scattering theory (ST) proposed in [NMOK03], models HRTFs through a sphere diffracting the sound. Hence, the elevation angle is not null unlike the two aforementioned HRTF models. The scattering theory is based on the expression of diffraction of sound when considering an ideal rigid spherical head introduced in [DM98] as

$$\begin{cases} H_1(\ell, \beta, f) = \frac{\ell c e^{-2j\pi f/c}}{2j\pi f (\frac{d}{2})^2} \sum_{m=0}^{\infty} (2m+1) P_m(\cos(-\frac{\pi}{2} - \alpha)) \frac{h_m(2\pi\ell f/c)}{h'_m(d\pi f/c)} \\ H_2(\ell, \beta, f) = \frac{\ell c e^{-2j\pi f/c}}{2j\pi f (\frac{d}{2})^2} \sum_{m=0}^{\infty} (2m+1) P_m(\cos(\frac{\pi}{2} - \alpha)) \frac{h_m(2\pi\ell f/c)}{h'_m(d\pi f/c)} \end{cases} \quad (2.11)$$

in which  $\ell$  characterizes the distance between the sound source and the center of the sphere that is the reference point.  $\alpha$  is then the source incident angle on this point with respect to the interaural axis. Furthermore the function  $P_m()$  and  $h_m()$  are respectively the Legendre polynomial of degree  $m$ , and the spherical Hankel functions of order  $m$ .  $h'_m()$  simply denotes the derivative of  $h_m()$ .

These different models underline the additional complexity induced by HRTFs, that should be accurately modeled as well as the environment acoustics, to obtain exploitable auditory cues for humanoid robots.

### 2.2.3 Auditory cues

The ratio of HRTFs encodes binaural information through the interaural transfer function. Indeed HRTFs depend on the direction of the sound that is implicitly related to the interaural cues. This relationship is not trivial when considering the actual HRTFs of robots. A typical solution consists in building a look-up table relating the ratio of HRTFs and direction of the sound source to interaural cues as exploited in [HLSVL06]. The relationship between the interaural cues and the sound location can also be obtained through the simplified HRTF models depicted above. In both AEG and RAEG, the ITD appears as a function of the azimuth angle of the source. The phase difference between the HRTFs is directly related to the ITD  $\tau$  that can be extracted as follows in the AEG configuration

$$\tau_{AEG} = \frac{d}{c} \cos \alpha \quad (2.12)$$

when in the RAEG, the ITD is defined as

$$\tau_{AEG} = \frac{d}{2c} \left( \left( \alpha - \frac{\pi}{2} \right) + |\cos \alpha| \right) \quad (2.13)$$

However, these two models are centered on modeling the direct path in AEG and delayed path in RAEG of the sound source, that is to say the effect of the head model on the ITD. Consequently they do not provide any meaningful ILD. On the other hand, the ST strategy is more realistic by accounting for the delay and attenuation induced by a spherical head. Hence from this model both ITD  $\tau$  and ILD  $\rho$  can be

inferred from (2.9) as

$$\begin{cases} \tau_{ST} = \frac{\arg(H_1(\ell, \alpha, f)) - \arg(H_2(\ell, \alpha, f))}{2\pi f} \\ \rho_{ST}(dB) = 20 \log_{10} \frac{|(H_1(\ell, \beta, f))|}{|(H_2(\ell, \alpha, f))|}. \end{cases} \quad (2.14)$$

From the ST approach, more reliable azimuth directions can be extracted from ITD and ILD values. This reliability will strongly depend on the capacity to cope with environment acoustics (*e.g.* reverberation, noise..), that alters the latter auditory cues.

Prior to inferring the sound location from HRTF models, ITDs and ILDs should be estimated from the recorded sound signal. The coincidence mechanism of the Jeffress auditory model inspired the techniques of interaural cues extractions, commonly based on the cross-correlation of the two recorded signals. In the basic form, the interaural cues can be obtained directly from the temporal signals. Considering a discrete sampling, the ITD obtained from the standard cross-correlation between the signals  $x_1(t)$  and  $x_2(t)$  is defined as

$$\mathbf{r}_{1,2}(\tau) = \sum_{t=1}^T x_1(t)x_2(t - \tau) \quad (2.15)$$

Hence the estimated ITD  $\hat{\tau}$  is obtained from the peak of the cross-correlation function:

$$\hat{\tau} = \operatorname{argmax} \mathbf{r}_{1,2}(\tau) \quad (2.16)$$

As ILD is concerned, it is simply defined as

$$\rho = \frac{\sum_{t=1}^T x_1^2(t)}{\sum_{t=1}^T x_2^2(t)} \quad (2.17)$$

that can be also expressed in dB as  $\rho_{dB} = 10 \log_{10}(\rho)$ .

As outlined by the localization mechanism in mammals auditory system, ITDs and ILDs are inferred from the spectral representation of the signal. In this context, ITDs are generally estimated in the spectral domain with the Generalized cross-correlation (GCC) [KC76]. The Cross-Correlation function exposed in (2.15), becomes

$$\mathbf{R}_{1,2}(\tau) = \sum_f^F \psi(f) \phi_{x_1, x_2}(f) e^{j\varphi(\tau)}. \quad (2.18)$$

where  $\varphi$  corresponds to the phase shift for a defined ITD  $\tau$ . The cross-spectral power density  $\phi_{x_1, x_2}$  is usually defined by

$$\phi_{x_1, x_2}(f) = \frac{1}{L} \sum_{l=1}^L X_1(f, l) X_2^*(f, l) \quad (2.19)$$

$X_1(f, l)$  and  $X_2^*(f, l)$  are respectively the STFT of  $x_1(t)$  and the conjugate of the STFT of  $x_2(t)$ . There exists also other alternative that uses a "max" pooling function so that

$$\mathbf{R}_{1,2}(\tau) = \max_l \sum_f^F \psi(f) X_1(f, l) X_2^*(f, l) e^{j\varphi(\tau)}. \quad (2.20)$$

This solution can perform better for intermittent sound sources since it does not integrate irrelevant information when the source is inactive [BOV12], as it would be the case with (2.19). Eventually  $\psi(f)$  is a weighting function, that generally aims to enhance the cross-correlation function. It should also be noted that when  $\psi(f) = 1$ , (2.18) simply corresponds to the standard cross-correlation. In (2.18), the maximum peak of the cross-correlation function gives an estimation of the actual ITD and can therefore be written as:

$$\hat{\tau} = \underset{\tau}{\operatorname{argmax}} \mathbf{R}_{1,2}(\tau) \quad (2.21)$$

When  $k$  sound sources are active, the latter function returns several peaks wherein the  $k$  first peaks are assumed to correspond to the ITD of each sound source.

Depending on the signal of interest, there exists in the literature various weighting functions  $\psi(f)$  that aim to enhance the sharpness of the peaks given by the cross-correlation function. The most common weighting function used in robot audition is certainly the PHASE Transform (PHAT). The PHAT filter is a normalization factor that is also known to be robust towards reverberation as demonstrated in several studies [Her86, DHD07] by only preserving the information related to the phase. Mathematically, the filter is thus defined as follows

$$\psi_{PHAT}(f) = \frac{1}{|\phi_{x_1, x_2}(f)|}. \quad (2.22)$$

The PHAT weighting is uniformly applied to all frequencies, without differentiating signal frequencies to noise frequencies. As a result GCC techniques using PHAT weighting functions are particularly sensitive to noise. Recently, extensions of the method have been proposed in order to control the degree of whitening and to limit the amount of degradation from the independent noise with a parametric modification. One can cite  $PHAT_\beta = |\phi_{x_1, x_2}|^{-\beta}$  where  $0 < \beta < 1$  is the tuning parameter that control the level of weighting [MP10]. Similarly the Reliability-Weighted Phase Transform (RWPHAT) integrates a noise weighting function, a Wiener filter gain, based on the SNR of the signal [VMR06] (see Chapter 1.2.3.2).

Another classic weighting function is the ROTH processor [Rot71]. The weighting function is expressed as follows.

$$\psi_{ROTH}(f) = \frac{1}{\phi_{x_1, x_1}(f)} \quad (2.23)$$

The Roth processor has the property of suppressing the frequency regions where noise is large [KC76].

There is also the Smoothed Coherent Transform (SCoT) filter given by

$$\psi_{SCoT}(f) = \frac{1}{\sqrt{\phi_{x_1, x_1}(f) \phi_{x_2, x_2}(f)}} \quad (2.24)$$

The SCoT method can be considered as a pre-whitening filter followed by a cross-correlation. It should also be noted that when  $\phi_{x_1,x_1}(f) = \phi_{x_2,x_2}(f)$ , the SCoT filter becomes a ROTH filter.

Eventually, the maximum likelihood (ML) weighting functions derived from [Saa96] and [BS97] are oriented to increase the accuracy of the estimated ITD by means of attenuating the signals fed into the cross-correlation function in spectral regions where the SNR is very low. Such a weighting function is defined as

$$\psi_{ML}(f) = \frac{\phi_{x_1,x_2}(f)}{|\phi_{n_1,n_1}(f)\phi_{x_1,x_1}(f)| + |\phi_{n_2,n_2}(f)\phi_{x_2,x_2}(f)|} \quad (2.25)$$

in which  $\phi_{n_i,n_i}(f)$  corresponds to the power spectral density of the noise  $N_i(f)$ . This method is known to be robust to noise, but its performance in real-world applications is relatively poor, because reverberation is not modeled in its derivation. The approach proposed in [RF04] integrates the reverberation in the noise calculation so that the noise spectrum is defined as

$$\phi_{n_i,n_i}^{rev}(f) = \gamma\phi_{x_i,x_i}(f) + (1 - \gamma)\phi_{n_1,n_1}(f) \quad (2.26)$$

where  $\gamma$  is a gain to tune depending on the reverberation level.

All these approaches can also be applied to gammatone filtered signals, that better approximate the filtering action of the cochlea. Hence ITDs and the ILDs are computed from each frequency decomposition given by the gammatone filters.

Eventually one can notice the wide variety of approaches centered on ITDs, whether for estimating its value or for establishing its relationship with source azimuth. Actually ILDs are quite straightforward to estimate but on the other hand, are challenging to exploit. There is no trivial modeling directly linking the value of an ILD to an azimuth direction. Hence, the main path of improvement for sound source localization approaches consists in enhancing ITD estimations from which it is easier to infer the source azimuth. In the same vein, distance cues and elevation cues until have rarely been exploited until now since they cannot be related by a simple model to a spatial location. Most approaches using elevation cues are based on the spectral notches, and thus the learning of HRTFs based on a specific pinna [KSK<sup>+</sup>05, RIJG08, HLSVL06]. As for distance estimation, a review proposed in [Rod10] based on the signal amplitude, spectral amplitude and the binaural cues, showed that a coarse approximation of the distance ( $< 1$  m) could be obtained. This study was, however, performed under controlled and static acoustic conditions. Some other approaches obtained a coarse estimation of the distance by using the DRR. In [LC10], the authors proposed a distance estimation from a DRR obtained through an equalization-cancellation method [Dur63], while the author of [Ves09] proposed a learning method based on the magnitude-squared coherence of binaural signals.

## 2.3 Binaural sound source localization

### 2.3.1 Typical approaches

For practical reasons (embeddability, space, resource consumption...), as well as for biomimetic sensing that gives more naturalness to interactions, binaural localization

remains a major topic of interest in robot audition [NN15]. In this context, localization solutions are based on the short-time cross-correlation of the signals between two microphones, derived from the Jeffress architecture. The interaural cues, more specifically ITDs, are at the core of binaural techniques. Nevertheless, the task is highly challenging due to the low number of sensors. Moreover, in realistic environments, interaural cues are particularly influenced by the robot morphology through HRTFs and by the acoustic properties of the robot location (*e.g.*, reverberation, noise) as mentioned in Section 2.1.2.2 and in Chapter 1.2. Regarding the complexity of acoustical environments, most early approaches tackled the localization problem in static scenes through simulations or fully controlled environments such as anechoic chambers.

Typically, these studies are based on the knowledge of HRTFs, to determine the azimuth and the elevation of the sound source, in controlled conditions. In [HLSVL06] or [KND06], localization is performed through the analysis of given HRTFs in order to link the azimuth and elevation to auditory cues. These latter works are principally based on ITDs and the spectral notches of the pinna. But these approaches are limited since they are fully based on HRTFs knowledge: any change in the acoustic conditions, or in the binaural setup may severely degrade the localization performance. The simplified head HRTF models provides a flexible alternative, so that the localization does not depend on accurate measurements of the HRTFs. For instance, in [HAGK03], experiments of planar localization are successfully performed using the ST models in environments with moderate reverberation. The RAEG has also been used in [NHOK02] or [NOK02] where complete solutions of localization, source separation and recognition are implemented. Yet, as expected the accuracy of such methods are decreased compared to HRTF-based localization techniques. Alternative approaches have also been proposed in [PDAP13] and [PBDM14] where the authors propose a maximum likelihood estimate based on the interaural transfer function to infer a sound direction.

With the growing interest for robot audition, a wide variety of approaches have been proposed in order to cope with realistic conditions. The major problems tackled in literature are: the flexibility of the localization performance, the robustness to acoustic conditions, robot/source motion, multi-source scenes. Such adverse conditions motivate the robot audition community to consider learning-based approaches instead of modeling all these conditions. In this context, because of the frequency-dependent patterns of ITDs and ILDs that vary considerably across robot platforms, azimuth-computation models are generally based on prior training or calibration with a binaural setup. For more flexibility, the authors of [RVE10] show that the azimuth can also be estimated by jointly taking into account ITDs and ILDs in a parametric model relating the azimuth angle to the interaural cues from pre-measured HRTFs. Unsurprisingly, this approach is less accurate than HRTF-based methods, but is able to cope with different binaural setups (*i.e.*, unknown HRTFs). However, the latter approach has been validated only in ideal simulated conditions. In more realistic conditions, the authors of [MvdPK11] introduce a probabilistic model (Gaussian mixture model) characterizing the complex interaction of ITDs and ILDs by training a model under various acoustic conditions to gain robustness. This method can cope with

multiple sources and reverberation, but has also been validated in simulated environments. Likewise, in [WW12], from the knowledge of HRTFs, the authors propose a supervised learning method integrating noise and reverberation to HRTFs to cope with adverse conditions. For a more practical purpose, in [DHSG15], the auditory cues are directly mapped to the spatial location of the source through a learning process in real world conditions. This learning process characterizes the related transfer function (RTF), the ratio of the TFs from the left and right microphones, that is a mixture of HRTFs and acoustic perturbations (see (2.7)). Hence, the method is limited to static configurations and to dedicated positions in the environment: any change in the position of the robot or in the experimental conditions may degrade the localization process. For more flexibility towards changing acoustic conditions, [YAZ13] proposes a learning approach based on a neural network. However this method requires a training stage in several and various conditions to obtain accurate auditory cues in changing acoustic conditions.

Globally, modeling all variability of acoustic environments and/or perception, that is unique for each robotic setup, remains a complex problem to solve. Hence, most of the aforementioned methods generally consider static scenes, where the source and the robot are still and at a favorable distance to each other. Integrating the robot motion or the source motion in the localization process increases the degree of complexity of such approaches as already outlined in Chapter 1.2.

### 2.3.2 Active audition

With the recent progress in understanding of the human auditory system, a renewed interest in binaural techniques has raised in the robot audition community through *active audition* [NOK00b], that exploits robot motion. The idea behind this approach resides in exploiting robot motion in order to gather additional information for sound localization. The “naive” approach to *active audition* simply consists in a triangulation of the measurements from several poses, as performed in [KB08, HST+99]. Obviously, in this context, the source must be static in order to obtain consistent results. Plethora of studies [LC11, MMWB15, TA06] show that localization is performed more accurately with a motion than in a static way. In [LC11], the authors compared several motion strategies for localization and concluded that the more beneficial strategy was obtained with purposeful motions, that consists in moving towards the given sound source in order to reduce the distance. In [MMWB15], the comparison of different head motions shows that the best performing head movement is to rotate the head towards the most likely source direction. These results were quite expected, since in the first case, reducing the distance to the sound source increases the DRR level. Consequently, interaural cues accuracy is increased, since the contribution of reflections is reduced in the recorded signal. In the second case, turning the head towards the probable sound source increases the spatial resolution of the localization process compared to eccentric positions.

In realistic environments, spurious measurements are likely to happen and should be resolved. To cope with measurements uncertainties, the framework proposed in [KNM15], formulates the *active audition* as a problem of bearing measurement



tracking. In this work, the data association problem is solved by coupling a particle filter to an interaural coherency parameter to detect the direct sound in the recorded signal spectrum. Subsequently, the source location is inferred from multiple poses of the robot. In [PDA12], the authors propose to solve these issues by combining a stochastic probabilistic data association, in order to discard false measurements, to an unscented Kalman filter. A potentially intermittent moving source can then be tracked during a motion of the robot. This method has been applied so far in a pseudo-anechoic environment. A partial experimental validation is proposed in [PBD<sup>+</sup>14]. Recently, [NCVC16] extended this approach by integrating an estimation of the source activity from bearing measurements and a voice activity detection algorithm. This solution provides promising results able to cope with moderate reverberation, however the validation is still performed in simulation and consider an array of microphones.

The aforementioned approaches consider a random motion of robots that may not be favorable to sound localization. *Active audition* also covers purposeful motion strategies aiming to position a robot with respect to the source location. In this context, the robot motion is set in order to take advantages of the redundancy of the auditory cues measurements. In general, motion strategies adopted by this kind of approaches are determined beforehand by the operator through heuristic approaches or fixed patrolling path. This is illustrated by [MEW09] where an incremental control strategy coupled with a neural network is used to position a robot closer to one static sound source. Similarly [TTLJ15] proposes a heuristic approach combining ITD cues and range cues to home towards a sound source. However this kind of approaches face some limitations. First defining a strategy of motion beforehand can lead to unnatural behaviour of the robot for an interaction task. Furthermore, the heuristic approach requires to model and predict all the possible sensory states of the robot. In realistic environments that are essentially unpredictable, such approaches would naturally be limited, while requiring the design of complex state machines. In a more sophisticated framework, the authors of [BPDCG12] develop a supervised learning method based on sensorimotor theories, where an agent learns the relationship between the auditory cues and the state of its motors and move accordingly to the recorded map. The experimental studies associated to this work were performed in anechoic conditions [BNP<sup>+</sup>10]. Purposeful motion strategies can also be used to improve the localization performance. [VSC15] proposes a strategy to find the minimal motion that could lead to the localization through occupancy grids, while recently, [BDFP16] developed an information-based feedback-loop in order to minimize the uncertainty of the localization process. Prior to this work, feedback-loops in robot audition context were introduced in [NMS02] and [KSM<sup>+</sup>03, KSK<sup>+</sup>05]. In the first work, a robot learns the acoustic map (*i.e.*, relation between ILDs, ITDs and azimuth and elevation) and uses a feedback-error learning scheme to orient the robot gaze towards a sound source. In the second set of work, the authors present a control scheme derived from a cost function characterizing the relationship between the position of a source and the orientation of the robot head. Although the cost function is empirically computed, a robot was able to turn towards the sound source without HRTF measurements, in anechoic conditions.

*Active audition* compared to static localization, widens the perspective for sound source localization. From a personal point of view, *active audition* is a necessary step towards dynamic, thus natural, sensing in fluctuating environments. This approach better reflects the auditory sensing of living beings, that is a dynamic process as expounded in Section 2.1.2.3. However when facing realistic environments, most *active audition* approaches still lack of robustness as underlined by the numerous works performed in simulation or controlled environments. Indeed, the auditory cues are measured in a static way and still requires a good modeling of the acoustic conditions and the robot structure. Furthermore, such approaches raise new challenges such as the fusion of uncertain measurements, the simultaneous motion of the sound source and the robot, and the motion strategy to adopt with respect to robot purpose.

## 2.4 Conclusion

Endowing robots with a reliable sense of hearing, and more particularly a sound localization framework, is a key feature for functional auditory scene analysis. The challenges raised by the robotic context, such as real time constraints or dynamic localization (motion of the source and/or the robot), lead the robot audition community to take inspiration from the human auditory system. Still, the human auditory system remains complex, and all the mechanisms involved in the auditory sensations are not fully understood. Among auditory cues, interaural cues, more particularly the interaural time difference, have been identified at a very early stage as related to the spatial location of the sound, and focused most of the studies in psychoacoustics. However, it is clear that the localization process involves the interaction of several cues in complex ways, that for now cannot be fully explained.

In addition to interaural cues, frequency-dependent filtering action by the body also known as HRTF plays a role for localization in the vertical plane and for resolving *front-back ambiguities*. The distance is suspected to be inferred from the ratio of direct sound over reflected paths and from the sound level. Unfortunately, until now, no simple auditory model can relate these cues to a spatial location of the source even though their contributions to localization have been acknowledged.

The prominent contributions of sound localization in robotics are based on the interaural cues in a binaural context. The use of such cues is, however, limited by the variability of the sound perception, characterized by the HRTFs and acoustic conditions. Binaural localization requires an accurate modeling of these parameters, which is a significant challenge. As a result, most of the models and techniques proposed in the robot audition literature consider static scenes, and are either validated in simulation or experimentally in anechoic conditions. Learning methods have also been explored in order to cope with realistic environments. But these methods generally require large training sets and complex learning procedures. Furthermore the estimated auditory cues generally correspond to a particular environment, with a given microphones configuration or robot structure. These methods are hardly adaptable to unknown environments since estimating all possible combinations of acoustic configuration is not affordable.

More recently, the *active audition* paradigm has been designed to overcome such limitations by using motion in order to benefit from measurements from different robot poses. The motion strategy to develop remains a central point of *active audition*. From different studies it appears that purposeful motions (getting closer or turning in the direction of the source) increase localization performance. These results are also corroborated by recent psychoacoustics studies that characterize the perception as a dynamic process where self-initiated motions are connected to the varying acoustic cues. However the approaches developed so far do not provide enough flexibility. Indeed *active localization* still requires a modeling of perturbations, that may be challenging to obtain in dynamic and realistic environments.

Hence to control robot motions, a different path, proposed in this thesis, consists in considering a low-level connection between motion and auditory perception, in order to avoid explicit localization and modeling of the perturbations. The work depicted here considers a generic approach that couples robot motion to auditory perception in a closed-loop control. The control loop moves a robot directly from an auditory feedback, instead of focusing on the preliminary explicit source localization that is the common path of improvement taken by the methods reviewed above. In these methods, the task generally consists in locating the sound source first and move the robot afterwards in an adequate position: the auditory system is completely independent from the robot control. By contrast, our approach is centered on the connection between auditory perception and robot control expressed through a sensor-based control framework detailed in the next chapter.



## Chapter 3

# Basics of sensor-based control framework

This chapter introduces the principles of sensor-based control that builds up the theoretical aspects of this thesis. The purpose here is to provide an alternative to sound source localization for robot positioning tasks in realistic environments.

In a first part, we introduce our formulation of robot motion control based on acoustic cues, that is an approach centered on robot control rather than sound localization. Thence, before a thorough analysis of the theoretical properties of sensor-based control, the benefits of using such a feedback loop are exposed in Section 3.1.1 through a scenario of a positioning task based on auditory information. In Section 3.1.2, a short review is given about the main usages and applications of sensor-based control in robotics.

The second segment of this chapter focuses on theoretical aspects of sensor-based control. The concept of task function leading to the principle of sensor-based control is first introduced in Section 3.2.1. Following this study, the actual theoretical framework that builds up *aural servo* (AS) paradigm is presented in Section 3.2.2, while accounting for the stability of the controller in 3.2.3 and for the positioning task to be performed by the robot, as exposed in Section 3.2.4.

### 3.1 Introduction to sensor-based control

#### 3.1.1 On the interest of sensor-based control for robot audition

*What would bring sensor-based control to robot audition?*

In this section we try to give a synthetic answer to this question by emphasizing the properties induced by this paradigm. Those properties greatly motivated the realization of this thesis. The previous chapter pointed out the limitations of sound source localization that prevents robot audition applications from evolving in realistic environments. Until now, most of the contributions tackled the issues of robot auditory perception from acoustic, signal processing or physiological perspectives. Typically, robot audition is addressed in the same way as machine hearing. It is not uncommon to apply robot audition techniques to other types of machines. This is

the case of the framework reported in [NSN15] that uses *active audition* to localize a sound source in 3D from a tablet device. Likewise, the human-machine interface proposed in [NMN15] uses the developments of robot audition for a car navigation system. In this paradigm commonly adopted by robot audition community, robots are considered as application platforms, since most of the solutions developed are not exclusive to robots. Only the context of robot audition differs from machine hearing, since the latter solutions should account for additional constraints such as ego-noises, real-time processing or limited resources as already mentioned in Chapter 1.2.

Nonetheless, unlike other machines, robots are endowed with motion. Robots can move in accordance with a given purpose, such as interacting with their surrounding environment. This simple characteristic can make a tremendous difference in auditory perception. Yet, the previous chapter stressed the benefits of motion for auditory perception through the *active audition* framework that can increase the robustness of sound source localization. However, *active audition* as well as classic localization techniques primarily focus on the quality of auditory perception from an acoustic and signal processing perspective. These techniques are attached to solve the question “*Where is the sound source?*” that corresponds to a machine hearing problem in a broad sense. The control of robots and the concept of giving a purpose to robots are generally neglected.

By contrast in this thesis, we aim to develop a dedicated control framework for auditory perception, that takes advantage of the characteristics of robots over machines. Indeed control, as an essential component of robotics, is introduced through a dedicated control framework for robot audition. The previous questioning about the source location, that builds up the sound source localization paradigm, becomes “*How to position the robot with respect to the sound source?*” The latter way to consider the auditory interaction can only be solved from a robotic perspective, since it requires a voluntary motion of the robot. In this paradigm, we thus model an auditory perception as a task to be completed by a robotic system. More explicitly, our approach consists in defining robot motion with respect to auditory feedback measured from several microphones. This is the principle of the sensor-based control approach.

To emphasize the interest of expressing auditory perception in this paradigm, let us detail a simple example of a positioning task. This task consists in orienting a robot head towards a given sound source. In this scenario, a robot is endowed with two microphones, while sound waves emitted by a source reach the microphones with an angle  $\alpha$  as illustrated in Figure 3.1.

In a classic approach, the task would consist in extracting the interaural cues (*e.g.*, ITD), computing the corresponding orientation  $\hat{\alpha}$  and eventually turning the robot head by  $\hat{\alpha}$ . The accuracy of this task depends on how accurate  $\hat{\alpha}$  is with respect to  $\alpha$ , so that the robot faces correctly the sound source. This constraint implies an accurate modeling of potential perturbations: reverberation, noises, head shadowing effect, ego-noises.... Taking into account all these parameters is challenging since most of them are uncontrollable, unpredictable and highly fluctuating.

In opposition to the latter approach, in a sensor-based control framework, the task

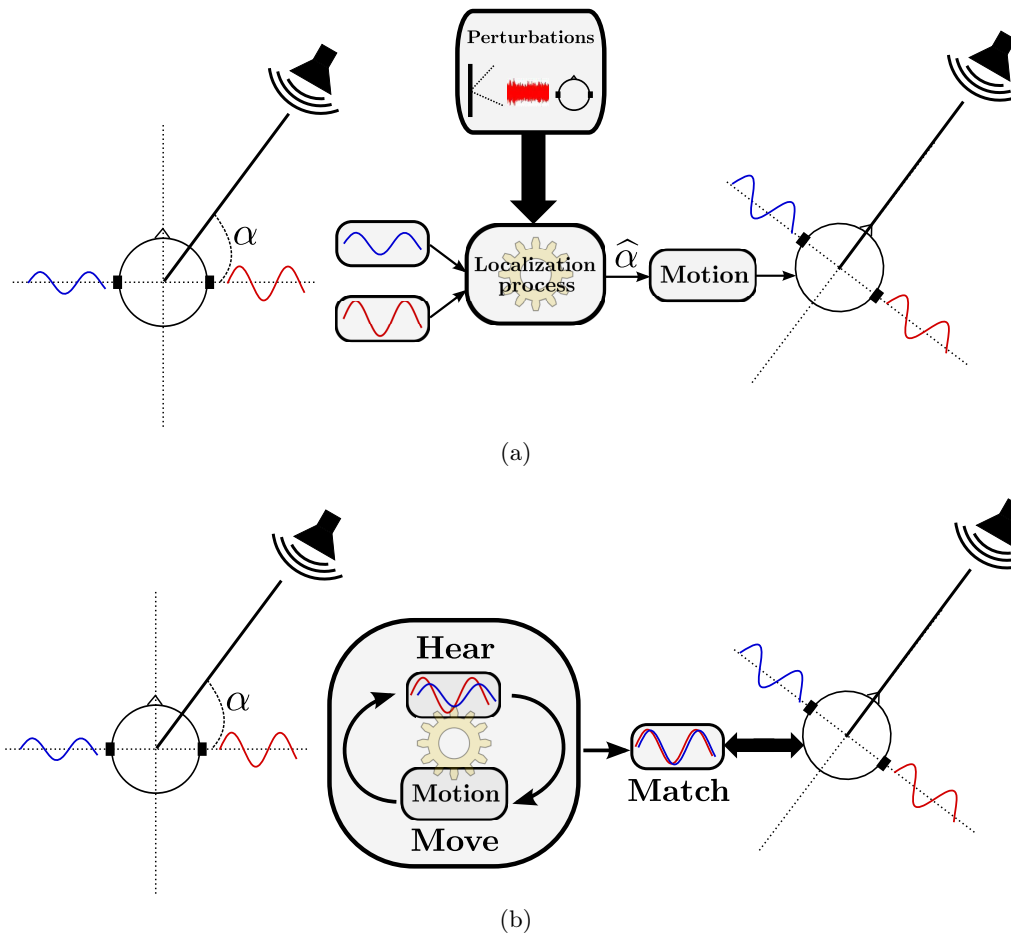


Figure 3.1: A robot perceives a sound source emitting from an angle  $\alpha$ . A head-turn task can be performed from two approaches. (b) Classic localization: After estimating the sound direction  $\hat{\alpha}$ , the head is turned accordingly; (c) Sensor-based approach: The head is turned until the two signals recorded by the left and right microphones match, which corresponds to the desired orientation of the robot.

consists in moving the robot in an appropriate direction until the signal recorded from the left microphone matches the signal recorded from the right microphone. Getting identical signals from both microphones (*e.g.* in time or level) logically implies that the robot is facing the sound source. One immediate advantage of such task modeling lies in the fact that it is not required to localize sound sources in the scene. The motion applied to the robot is not controlled by an estimation of the source location but rather steered by auditory properties implied by the desired pose (*i.e.*, similar measurements from both sensors) and the sound features variation with respect to this motion. These two characteristics are essential key points that emphasize the robustness of sensor-based approaches over sound localization methods for robot motion control.

First, sound features variation is an element that may be considered robust with



respect to usual perturbations faced in robot audition. In Figure 3.1b, without any thorough analysis, while achieving the task, it is expected that the amplitude of the right signal decreases while the left signal amplitude increases. When the robot faces the sound source, these two signals should have the same amplitude. A similar reasoning holds for the signal delay of perception between two microphones. Analogously, the sound level increases when the sensors moves closer to the given sound source while it decreases with an opposite motion. An accurate achievement of a task depends on the consistence of the sound features variation and not on accurate intrinsic values as it is for sound localization approach.

The second characteristic exploits measured auditory features with respect to desired auditory features. These desired features build up a reference that should be reached. They are either experimentally measured or estimated. The system is thus defined with respect to local measurements rather than global measurements as it is usually assumed for sound localization. Intrinsically, these desired features could be erroneous with respect to a global ground truth obtained from a localization perspective. However, as long as these values can be reached, the control task can be correctly achieved. When considering more specifically the context of facing a sound source, signal measurements around this pose are known to be the most sensitive: the spatial resolution and accuracy of auditory cues are at their finest. This property has already been discussed for machine hearing in [CLLH07] reporting that the performance of localization and tracking algorithms can be improved by turning the head to ensure that the tracked source remains in the region directly in front of the head. The same evidence holds for human localization performance as exposed in [CLH97] or [MM90]. Consequently, in this context, the control framework can reach with a higher accuracy the desired auditory features.

Hence the combination of these two characteristics constitutes the key strength of formulating the motion control in a sensor-based control approach. Our formulation of the problem simply relaxes accuracy requirements, which is a critical point for sound localization approaches. As a consequence perturbations like reverberation and potential noises that alter the auditory cues estimation, have then a limited effect. A beneficial effect of this result concerns the well-known subjectivity of auditory perception characterized by HRTFs. This subjectivity highly constrains any localization systems embedded in robotic heads (as discussed in Chapter 2) and it generally requires an accurate modeling of perturbations. In the task represented in Figure 3.1b, the symmetrical property of the robotic structure implies that the shadowing effect from left and right microphones should be the same when the robot is facing the sound source. Moreover knowing that auditory cues variation is not affected by such perturbation, the scattering effect is nullified and the latter task can be completed without any modeling of HRTFs, as it will be shown in Chapter 6.

Eventually one can highlight the advantages related to the closed-loop control scheme. The closed-loop control is a highly reactive process that is especially adapted to dynamic environments, since a new control input is computed at each processing frame. Hence this method should be able to cope with moving sound sources. Likewise, measurements are updated and corrected in real-time, which limits the effect

of spurious and erroneous estimations that are not consistent with robot motions. As stated above, robots voluntary motions induce a given variation of the measured auditory cues that can be anticipated. And, similarly to *active audition* processing, several measurements bring additional information that can increase the robustness of auditory perception.

From all these observations, it appears that AS framework can bring robustness and flexibility to auditory tasks, which is a principal requisite for robots evolving in realistic environments. The potential benefits reported at this stage will be supported all along the development of this manuscript by theoretical and experimental evidence in Chapters 4, 5 and 6.

### 3.1.2 Sensor-based control in robotics

One of the most well-known sensor-based application is the so-called *visual servoing* (VS) that uses computer vision data to control robots dynamically. VS refers to the paradigm of closed loop control techniques of actuated systems with visual feedback. With the high preference of vision over other modalities, VS naturally becomes a tool for automatic control in unknown, complex and dynamics environments. The first approaches to sensor-based control using vision were introduced in the late 70s [Agi77, Agi79] followed by [WSN87, RCE90, FM89] in the 80s. These works pioneered robot control based on visual information. Indeed despite the limited performance of the computing architectures in the 80s, these works allowed to exploit visual information in real-time with reactive behaviors.

Nowadays, regardless of the diversity of vision-based robot control, VS is still considered as an efficient approach for vision-based robotic tasks such as positioning and tracking. Based on the nature of visual information, the existing VS approaches can be classified into two main categories: *image-based visual servoing* (IBVS) and *position-based visual servoing* (PBVS) [WSN87, CH06]. In IBVS techniques, visual features (*e.g.*, points, lines, moments) in the image plane are used to compute a control law. In PBVS techniques, visual information is used to extract the pose of a given target with respect to the controlled robot. A control law is then computed from the error between the current and desired poses. In AS, this approach would consist in localizing a given sound source at each iteration of the control loop. However, because of localization limitations discussed in the two previous chapters, this approach is not interesting (yet) for AS framework.

In recent decades, VS techniques have been widely applied to different platforms such as classic robot manipulator [C<sup>+</sup>96], mobile robots [Cor03], or more recently to aerial vehicles [MH07]. VS has already been applied to diverse contexts. In medical robotics the accuracy inherent to VS is used to develop medical assisting robots. This is illustrated by plethora of contributions such as [KGD<sup>+</sup>03] proposing a robotic system that automatically positions surgical instruments during robotized laparoscopic surgical operations. In [YN01], an automated embryo DNA injection system controlled within a VS framework is reported while ultrasound images are used to control an ultrasound probe held by a medical robot in [MKC08]. VS has also been applied to the field of virtual reality by formulating the problem of camera

pose estimation as a regulation task achieved in the image plane [MC02, CMPC06]. More recently visual feedback has been used to control the deformation of compliant objects by robot manipulators [NALRL14] through the *visual deformation servoing* framework. This wide range of applications classify VS approach as a general paradigm, that is shaped by different contexts and purposes. Unsurprisingly VS has been used with different types of visual sensors beside the classic perspective camera. The utilized sensors could be catadioptric cameras [BMH03], generalized cameras [CMS11], stereoscopic [HCM94] or RGB-D sensors [TM12] among others. These sensors could also be exploited while being externally fixed [HDE98].

Nonetheless sensor-based approaches are not limited to visual feedback and have been applied to a wide range and variety of sensors. In the early 90s, [EMS90] showed that most sensor-based applications may be treated within a unified control loop framework. In the latter paper, the authors develop a global approach to the problem of proximity and force-based control in robotics applications that lead to the development of *proximity servoing*. Nowadays a significant number of applications using sensor-based control are related to touch sensing feedback, whether human-robot interaction for collaborative tasks, obstacle avoidance, manipulation or grasping tasks. This can be illustrated by [KJHJ05] in which the authors introduce a framework to control a team of robots, through proximity sensors (*e.g.*, laser) feedback, that follow walls or terrains at a given distance. The application proposed in [WS08] explores the use of electric field pre-touch as a feedback signal for closed loop control for aligning and pre-shaping a robotic hand for grasping tasks. The paper proposed by [SZS94] showed that sensor-based control could also be used with tactile images produced by tactile sensors mounted on the fingertips of a robot hand for object manipulation. This paper gives rise to the well-known *tactile servoing*. Tactile feedback can be used nowadays for tactile exploration tasks [LSHR13] or for more complex tasks, like dexterous prehension as proposed by [OSC00].

Furthermore, the versatility of sensor-based control framework engendered the topic of multimodal sensor-based robot control. In this context, a task is achieved by combining data of sensors measuring different physical phenomena in an unique control framework. Combining information from several sensors in the sensor-based approach can be interesting to overcome the limitations and flaws inherent of a particular sense (*e.g.* limited field of view of cameras). This idea is derived from the concept of *hybrid task* [EMS90], in which a given task is solved by completing additional objectives (*e.g.* trajectory tracking). The paper [HIA98] illustrates the concept of *hybrid task* by achieving a contacting task in an unknown environment, while a robot is visually guided. Sensor-based control has the advantage of achieving a given task without fusing different sensing modalities or designing complex state machines. For instance, in [ENSHW14] a gripper is controlled by combining pre-touch sensors feedback and tactile sensors feedback, for a grasping task. In [DSLSV07], the authors introduce an approach centered on human-robot cooperation combining vision and force control, in order to achieve safe human-robot interaction in a given working space.

This short review of sensor-based control, through the principle of task functions and its usage in robotics, emphasizes the wide application field of this kind of ap-

proaches. Such a vast literature cannot obviously be summed up in these two latter sections. Only the most essential aspects for the understanding of this manuscript have been introduced. We refer the interested reader to dedicated textbooks addressing the sensor-based control in robotics such as [SELB91], [C+96] or [CH08]. At our end, we concentrate on applying sensor-based control in the field of robot audition, as detailed in the next section.

## 3.2 Theoretical framework

### 3.2.1 General principles of task function

The development of exteroceptive sensors, which are now mostly affordable and embeddable, has stimulated their use on robots. Such sensors include cameras, microphones, tactile sensors, lasers and force sensors for instance. In the robotic context, exteroceptive sensors give a measurement of an interaction with the surrounding environment. The typical approach consists in regulating this interaction through the information gathered from the sensors. Flexible and reactive control of the interaction is then aimed as well as precision, speed and low computation cost.

One way to control robots is by open-loop techniques. Environment sensing is performed in a first step and robot motion is deduced afterwards. As illustrated by Figure 3.2, the approach is based on an initial measurement  $\mathbf{s}(\mathbf{r}_0)$  obtained from the sensor at a position  $\mathbf{r}_0$ , and a desired pose  $\mathbf{r}^*$ . Therefore this approach is regulated by the control input  $\dot{\mathbf{q}}$  until the desired pose  $\mathbf{r}(\mathbf{q}) = \mathbf{r}^*$  is reached from the extracted initial pose  $\mathbf{r}_0(\mathbf{q})$ , where  $\mathbf{q}$  characterizes the configuration of a given robot. This is typically the approach considered in sound source localization where the azimuth and elevation angles and/or distance information can be used to position a robot. However estimating the initial pose  $\mathbf{r}_0(\mathbf{q})$  from a unique sensors measurement is not always trivial, as illustrated by sound localization limitations, and generally requires a prior model of the surrounding environment. Moreover when  $\mathbf{r}_0(\mathbf{q})$  is extracted, a motion planning strategy is then required to reach the desired pose from the current pose. This action is performed “blindly” or more exactly in a “deaf manner” in our context, since  $\mathbf{r}(\mathbf{q})$  is not related to sensor information anymore.

By contrast sensor-based control is designed by a closed-loop control that consists in a dynamic "sense and move" approach. The control input  $\dot{\mathbf{q}}$  to a given robot is steered by the dynamic feedback of sensor measurements  $\mathbf{s}(\mathbf{r})$  and generates a motion until a pose characterized by  $\mathbf{s}(\mathbf{r}) = \mathbf{s}^*(\mathbf{r}^*)$ , is reached. Formally, the goal of a sensor-based task consists in controlling the system by regulating the error between the observed and desired values  $\mathbf{s}(\mathbf{r}) - \mathbf{s}^*(\mathbf{r}^*)$  to 0. By expressing the motion task at the sensory level, the step of estimating a desired pose is skipped (unless a PBVS approach is used). Likewise, the motion planning is avoided. Hence this approach may be less conditioned by the knowledge of the environment, and may provide better solutions in unpredictable scenarios. The core principle to perform the latter regulation lies in controlling a robot from a  $C^2$ -differentiable error function  $\mathbf{e}(\mathbf{q}, t)$ , that has to be regulated during a time interval  $[0, T]$ . This is the principle of the task function introduced in [SELB91] that depends on the robot configuration  $\mathbf{q}(t)$  at a

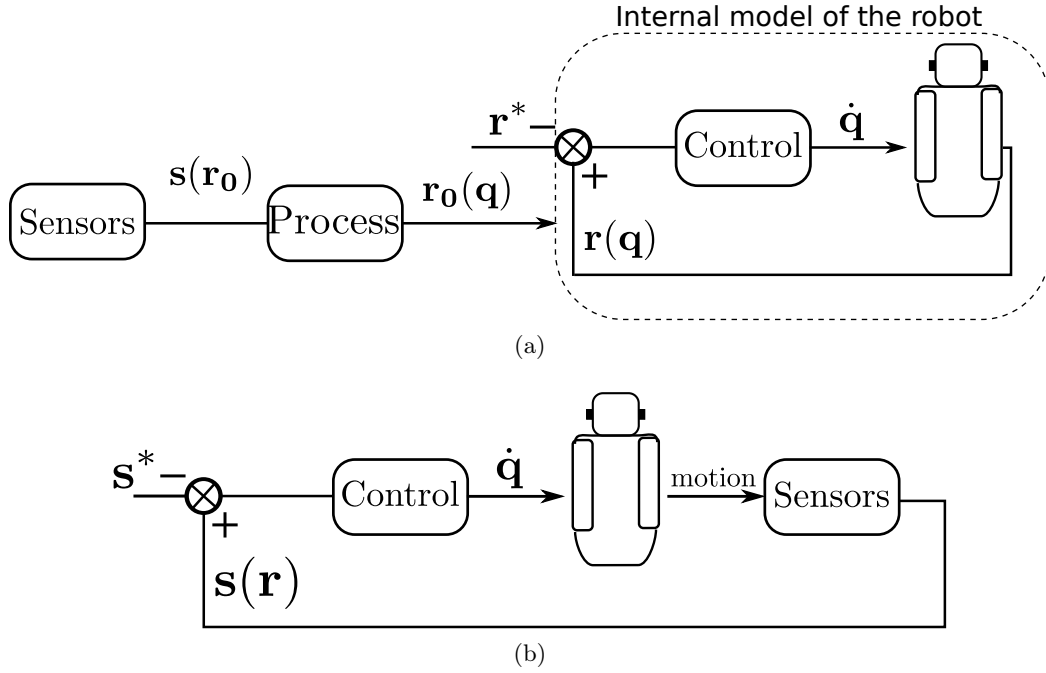


Figure 3.2: Two different approaches to robot control (a) An open-loop approach consists in deducing the pose of the robot  $\mathbf{r}_0(\mathbf{q})$  with respect to a sensed target from a unique sensor measurement  $\mathbf{s}(\mathbf{r}_0)$  and move the robot accordingly through  $\dot{\mathbf{q}}$ ; (b) A sensor-based approach links the motion of the robot to the sensor measurement  $\mathbf{s}(\mathbf{r})$  in a feedback loop until the robot reaches a desired configuration characterized by a demanded measurement  $\mathbf{s}^*$ .

time  $t$ . The task function to be regulated can be expressed in terms of a measure  $f$  of the robot state as:

$$\mathbf{e}(\mathbf{q}(t), t) = f(\mathbf{q}(t), t) - f(\mathbf{q}^*(t)) \quad (3.1)$$

Generally the function  $f$  corresponds to a feature  $\mathbf{s}$  that underlies spatial information about the robot configuration and that evolves accordingly to the motion applied to the robot. Since  $\mathbf{e}(\mathbf{q}(t), t)$  is differentiable, the derivative of (3.1) that conditions the task function regulation is defined as

$$\dot{\mathbf{e}}(\mathbf{q}(t), t) = \frac{\partial \mathbf{e}(\mathbf{q}(t), t)}{\partial \mathbf{q}(t)} \dot{\mathbf{q}}(t) + \frac{\partial \mathbf{e}(\mathbf{q}(t), t)}{\partial t}. \quad (3.2)$$

In (3.2), two terms can be distinguished. The first one, involving the task Jacobian  $\mathbf{J}_e = \frac{\partial \mathbf{e}(\mathbf{q}, t)}{\partial \mathbf{q}}$  characterizes the task function variation according to the robot motion. The second term reflects the variation of the task function  $\mathbf{e}$  with respect to time. More exactly, "disturbances" caused by the sensed target from which is inferred the robot state. From the control perspective, the task function  $\mathbf{e}(\mathbf{q}, t)$  is said to be admissible, that is to say well-conditioned if it satisfies some specific properties such as [ECR92]:

- The regularity condition:  $\mathbf{J}_e$  should not be singular (*i.e.*, it should be invertible)  $\forall t \in [0, T]$ .
- There exists a unique ideal trajectory  $\mathbf{q}_r(t)$  so that  $\mathbf{e}(\mathbf{q}_r(t), t) = 0 \forall t \in [0, T]$  starting from an initial given condition  $\mathbf{q}_r(0) = \mathbf{q}_0$ .

In our case we are particularly interested in the configuration when the sensors are mounted on robots, since we generally consider embedded microphones. Then, sensors' motions are directly governed by the robot configuration, so that the measured features  $\mathbf{s}(\mathbf{r}, t)$  can be written as

$$\mathbf{s}(\mathbf{r}, t) = \mathbf{s}(\mathbf{q}(t), t). \quad (3.3)$$

$\mathbf{s}(\mathbf{r}, t)$  depends on the configuration of the robot at the current time  $\mathbf{q}(t)$  but also on the potential evolution of the sensed target at  $t$ . Thus, the task function given in (3.1) can also be expressed as

$$\mathbf{e}(\mathbf{q}(t), t) = \mathbf{C}(\mathbf{s}(\mathbf{q}(t), t) - \mathbf{s}^*), \quad (3.4)$$

where  $\mathbf{C}$  is a combination matrix that allows to match the dimension of  $\mathbf{s}$  with the number of controlled degrees of freedom (DOF), when features are redundant. Thus,  $\mathbf{C}$  is used when the dimension of  $\mathbf{s}$  is greater than the number of controlled DOF [SAA91]. Nonetheless in the cases developed in this thesis, this configuration does not occur and  $\mathbf{C} = \mathbb{I}$ . Consider, for simplicity, the case of a regulation task in which a desired constant value  $\mathbf{s}^*$  is specified. This value can be either computed or experimentally measured by manually moving the robot to a desired configuration. Omitting the time dependence of the feature  $\mathbf{s}$  and the robot configuration  $\mathbf{q}$ , from (3.4), (3.2) is equivalent to

$$\dot{\mathbf{s}} = \frac{\partial \mathbf{s}}{\partial \mathbf{q}} \dot{\mathbf{q}} + \frac{\partial \mathbf{s}}{\partial t}, \quad (3.5)$$

which characterizes the variation of the features  $\mathbf{s}$  with respect to the applied motion. In this case the “disturbance” term exhibited in (3.2) corresponds to  $\frac{\partial \mathbf{s}}{\partial t}$  that characterizes the potential motion of the sensed target, independently from the sensors. The first term of (3.5)  $\mathbf{J}_s = \frac{\partial \mathbf{s}}{\partial \mathbf{q}} \in \mathbb{R}^{k \times m}$  is the Jacobian feature matrix that can be decomposed from the chain rule as

$$\frac{\partial \mathbf{s}}{\partial \mathbf{q}} = \frac{\partial \mathbf{s}}{\partial \mathbf{r}} \frac{\partial \mathbf{r}}{\partial \mathbf{q}}. \quad (3.6)$$

The latter equation contains the robot Jacobian  $\mathbf{J}_r = \frac{\partial \mathbf{r}}{\partial \mathbf{q}}$  that links the sensors' motion to the robot's motion. Since we consider embedded sensors, this Jacobian solely depends on the position of robot sensors and therefore is assumed to be known. The robot Jacobian is multiplied by the interaction matrix  $\mathbf{L}_s = \frac{\partial \mathbf{s}}{\partial \mathbf{r}}$  that characterizes the chosen task through  $\mathbf{s}$ , by linking the variation of the selected feature to sensors' motion [CH06]. If we assume a motionless target that does not induce any change in values to the features  $\mathbf{s}$  and if no potential disturbances are acting on the system, the second term of (3.5) becomes  $\frac{\partial \mathbf{s}}{\partial t} = 0$ . Hence, once  $\mathbf{s}$  is selected, the relation,

characterized by  $\mathbf{L}_s$ , between a set of features  $\mathbf{s}$  and the sensors velocity  $\mathbf{v}_s$  is given by

$$\dot{\mathbf{s}} = \mathbf{L}_s \mathbf{v}_s \quad (3.7)$$

in which  $\mathbf{L}_s \in \mathbb{R}^{k \times n=6}$  is a matrix sized by  $k$  the number of measurements and the dimension of  $\mathbf{v}_s$  ( $SE(3)$ ). The sensors' velocity is defined by  $\mathbf{v}_s = (\mathbf{v}_s, \boldsymbol{\omega}_s)$  where  $\mathbf{v}_s = (v_x, v_y, v_z)$  denotes the spatial linear velocity and  $\boldsymbol{\omega}_s = (\omega_x, \omega_y, \omega_z)$  the angular velocity. By extension, when using (3.4) and (3.7), we obtain the relationship between the sensors' velocity and the time variation of the error:

$$\dot{\mathbf{e}} = \mathbf{L}_e \mathbf{v}_s, \quad (3.8)$$

where  $\mathbf{L}_e = \mathbf{L}_s$ . Then, designing a task regulation mainly consists in *i*) selecting  $\mathbf{s}$ , *ii*) computing or measuring  $\mathbf{s}^*$  and *iii*) modeling the task Jacobian  $\mathbf{J}_s$  (via the interaction matrix  $\mathbf{L}_s$ ) while ensuring that the admissibility conditions of the task are not violated. Eventually, under the hypothesis of motionless target, the dynamics of the task function to be regulated is given by

$$\dot{\mathbf{e}}(\mathbf{q}, t) = \mathbf{J}_s \dot{\mathbf{q}}. \quad (3.9)$$

### 3.2.2 Control scheme

As reported previously, the sensor-based control can either be designed in order to consider sensor features or directly sensor pose in the loop. In our context, we are mainly interested in designing a control scheme directly from auditory features since sound source localization happens to be extremely complex. From then on, a simple control scheme can be designed with a purpose of exponential decoupled decrease of the error [ECR92]. In this case, the time variation of the error for the feedback system should follow

$$\dot{\mathbf{e}} = -\lambda \mathbf{e} \quad (3.10)$$

with  $\lambda > 0$  a gain that tunes the time to convergence. Then, by combining (3.7) and (3.10) we obtain

$$\mathbf{v}_s = -\lambda \mathbf{L}_s^+ \mathbf{e} \quad (3.11)$$

where  $\mathbf{L}_s^+ \in \mathbb{R}^{n \times k}$  is the Moore-Penrose pseudo-inverse of the interaction matrix. The pseudo-inverse is used when the inverse is not defined, that is, when  $n \neq k$  or  $n = k$  and  $|\mathbf{L}_s| = 0$ . When  $k < n$  and the rank of  $\mathbf{L}_s$  is  $k$ , the Moore-Penrose pseudo-inverse is defined as

$$\mathbf{L}_s^+ = \mathbf{L}_s^\top (\mathbf{L}_s \mathbf{L}_s^\top)^{-1} \quad (3.12)$$

while when  $k > n$ , and the rank of  $\mathbf{L}_s$  is  $n$

$$\mathbf{L}_s^+ = (\mathbf{L}_s^\top \mathbf{L}_s)^{-1} \mathbf{L}_s^\top. \quad (3.13)$$

Otherwise, when  $k = n$  and  $|\mathbf{L}_s| \neq 0$  we simply have  $\mathbf{L}_s^+ = \mathbf{L}_s^{-1}$ . The same reasoning can be developed by considering the feature Jacobian  $\mathbf{J}_s$ , when the sensors are embedded on a robot. Thus, the control input to the robot is given by

$$\dot{\mathbf{q}} = -\lambda \mathbf{J}_s^+ \mathbf{e}. \quad (3.14)$$



In this case since  $\mathbf{C} = \mathbb{I}$ , we have  $k \leq m$ , and only the left pseudo-inverse given by (3.12) is used. Nonetheless in real configurations, it is impossible to know perfectly in practice either  $\mathbf{L}_s$  or  $\mathbf{L}_s^+$  (and by consequence  $\mathbf{J}_s$  and  $\mathbf{J}_s^+$ ). In the first case, the interaction matrix could depend on a quantity that cannot be directly measured by using the sensors. A typical example concerns the IBVS interaction matrix that contains the depth of the image feature relative to the camera frame that cannot be obtained from the image. Thus generally an approximation  $\widehat{\mathbf{L}}_s^+$  of the inverse of the interaction matrix is used in the control loop. Consequently the control input to the system is more exactly defined as:

$$\mathbf{v}_s = -\lambda \widehat{\mathbf{L}}_s^+ \mathbf{e} \quad (3.15)$$

Several ways exist to compute  $\widehat{\mathbf{L}}_s$  and these different strategies affect the behavior of the system and possibly the stability of the control scheme (see Section 3.2.3). Some of these strategies exposed in [CH06] are thoroughly detailed and studied in a practical use case in Chapter 5.3.2.2. Such type of regulation task allows the development of control methods by directly linking perception to action. To some extent, sensor-based control can be referred to a simplified modeling of human cognitive processes, that constantly update its behavior from the sensory feedback. More precisely sensor-based control may be considered as a bio-mimicking process for robot control and may underlie several advantages, as detailed in the previous section, especially when considering the hearing sense. The works exposed in Chapter 2.1.2 reported that human auditory system is a dynamic process linking motion and perception. Until now this process has not been fully exploited in robot audition, while it is usually adopted with other sensory modalities.

### 3.2.3 Stability

In this part, we consider the fundamental issues related to controllers stability. The stability properties ensure that our controller is able to converge towards the configuration for which  $\mathbf{e} = 0$ . This analysis is based on Lyapunov theories dedicated to nonlinear control systems such as the feedback loop developed in the previous section. The Lyapunov stability is an mathematical extension of physical observation of mechanical systems: the total energy of a mechanical system is continuously dissipated before eventually reaching an equilibrium state or point [SL<sup>+</sup>91]. An equilibrium point is a point where the system can stay forever without moving. Hence, for a system  $\dot{x} = f(x)$ , a state  $x^*$  is an equilibrium state of this system if once  $x(t)$  is equal to  $x^*$ , it remains equal to  $x^*$  for all future time. The concept of Lyapunov stability consists in constructing a Lyapunov candidate function (similarly to the energy function of a mechanical system) for the system, and to examine the function's time variation that should decrease.

First, we introduce the concept of local stability that qualitatively define a system as locally stable if starting the system somewhere near its desired operating point implies that it will stay around the point ever after. By defining the close neighborhood by a sphere of radius  $R$ , an equilibrium point  $x^* = 0$  is said to be stable if, for any  $R > 0$ , there exists  $r > 0$ , such that if  $\|x(0)\| < r$ , then  $\|x(t)\| < R \forall t > 0$ .

From this definition, the asymptotic stability property can be inferred. The asymptotic stability means that the equilibrium point is stable, and any state  $x(t)$  started close to  $x^*$  actually converge to  $x^*$  as time  $t \rightarrow \infty$ . Hence, an equilibrium point  $x^* = 0$  is asymptotically stable if it is stable, and if there exists  $r > 0$  such that if  $\|x(0)\| < r$  then  $x(t) \rightarrow 0$  as  $t \rightarrow \infty$

More formally the local asymptotic stability is demonstrated from a stability criterion that makes use of a Lyapunov function candidate  $\mathcal{L}$ . This function should decrease along with the system state trajectories. For a system  $\dot{x} = f(x)$  with an equilibrium at  $x^* = 0$ , the local asymptotic stability is ensured by the set of following sufficient conditions: in a sphere of radius  $R$ , there exists a scalar function  $\mathcal{L}$  with continuous first partial derivatives such that

$$\begin{aligned} \bullet \mathcal{L}(x) \text{ is locally positive definite: } \mathcal{L}(0) = 0 \text{ and } \mathcal{L}(x) > 0 \forall x \neq 0 \\ \bullet \dot{\mathcal{L}}(x) \text{ is locally negative definite: } \dot{\mathcal{L}}(0) = 0 \text{ and } \dot{\mathcal{L}}(x) < 0 \forall x \neq 0, \end{aligned} \quad (3.16)$$

then the equilibrium point  $x^* = 0$  is locally asymptotically stable. Note that the local stability is obtained as soon as  $\dot{\mathcal{L}}(x)$  is locally negative semi-definite.

If the asymptotic stability is valid for any initial states, the equilibrium point  $x^* = 0$  is said to be globally asymptotically stable. The global asymptotic stability ensures that the system is attracted by the equilibrium point independently of the starting pose and is thus not restricted to the neighborhood of the equilibrium point. An additional condition on the function  $\mathcal{L}$  is required:  $\mathcal{L}(x)$  must be radially unbounded which is shown by

$$\|x\| \rightarrow \infty \Rightarrow \mathcal{L}(x) \rightarrow \infty. \quad (3.17)$$

Assuming that there exist a function  $\mathcal{L}(x)$  of the state  $x$  with continuous first derivatives, the global asymptotic stability is thus defined by the following set of sufficient conditions

$$\begin{aligned} \bullet \mathcal{L}(x) \text{ is globally positive definite: } \mathcal{L}(x) > 0 \forall x \neq 0. \\ \bullet \dot{\mathcal{L}}(x) \text{ is globally negative definite: } \dot{\mathcal{L}}(x) < 0 \forall x. \\ \bullet \mathcal{L}(x) \text{ is radially unbounded: } \mathcal{L}(x) \rightarrow \infty \text{ if } x \rightarrow \infty. \end{aligned} \quad (3.18)$$

For the specific sensor-based control exposed in this thesis, the system is defined by (3.8). We then consider the candidate Lyapunov function characterized by the squared error norm  $\mathcal{L}(\mathbf{e}) = \frac{1}{2}\|\mathbf{e}\|^2$ . This function is inherently globally positive definite and radially unbounded (3.18). In order to fulfill the condition related to  $\dot{\mathcal{L}}(\mathbf{e})$  being globally negative definite, we derive the candidate function as follows:

$$\dot{\mathcal{L}}(\mathbf{e}) = \mathbf{e}^\top \dot{\mathbf{e}}. \quad (3.19)$$

The latter equation can be detailed by injecting (3.8) and (3.11) in it:

$$\dot{\mathcal{L}}(\mathbf{e}) = -\lambda \mathbf{e}^\top \mathbf{L}_s \widehat{\mathbf{L}}_s^+ \mathbf{e}. \quad (3.20)$$

As a consequence the global asymptotic stability of the closed-loop control system is ensured as soon as

$$\mathbf{L}_s \widehat{\mathbf{L}}_s^+ > 0. \quad (3.21)$$

Yet, it is not always possible to ensure this condition, that consists in proving that the product in (3.21) is positive definite. A thorough analysis of the eigenvalues of the symmetric part of  $\mathbf{L}_s \widehat{\mathbf{L}}_s^+$  would probably provide boundaries for the approximation of the interaction matrix  $\widehat{\mathbf{L}}_s^+$ . Such complex demonstration are not required for the comprehension of this thesis and we refer the interested reader to [MMR10] that consists in a typical study of the stability issue. Nonetheless, some essential stability results supporting AS framework are provided for different task configurations considered in the manuscript. In general, it is assessed that if the number of features is less or equal to the number of sensor DOF (*i.e.*,  $k \leq n$ ), the global asymptotically stability condition (3.21) is ensured if the approximations involved in  $\widehat{\mathbf{L}}_s^+$  are not too coarse [CH06].

### 3.2.4 Virtual linkages

Additionally to the stability properties, an important notion used in this thesis concerns the task that is achieved by a robot. Introduced in [SELB91] and applied to VS in [CRE93], virtual linkages provide a theoretical modeling of the task. A Virtual linkage representation is a tool for analyzing the sensor-based task directly from the properties of the interaction matrix. From the interaction matrix, the behavior of a given robot can be anticipated, that is, the set of allowed poses for the robot to complete a given task. The concept is derived from the definition of mechanical linkages that characterize a mechanical structure of assembled and connected bodies or links. The connections between links are modeled as providing movement of pure rotation or translation. Mechanical linkages are usually designed to transform a given input force and movement into a desired output force and movement. A simple illustration of mechanical linkage is materialized by robotic arms, where each link building up the arm is connected by a joint with a prismatic or rotoid motions.

In our context, a virtual link is defined between a robot and the sensed object (*i.e.*, the sound source) and governs the admissible robot motion to complete a task. Let  $\mathbf{v}_{s_i}^*$  be a virtual motion of the sensor (or the robot), so that the sensor measure  $\mathbf{s}_i$  remains constant. Thus from (3.7),  $\mathbf{v}_{s_i}^*$  is a solution of the equation

$$\mathbf{L}_{s_i} \mathbf{v}_{s_i}^* = 0 \quad (3.22)$$

and a skew reciprocal to  $\mathbf{L}_{s_i}$ . The set of motions  $\mathbf{v}_{s_i}^*$  that leaves the set of features  $\mathbf{s}$  unchanged is then denoted  $\mathbf{S}^*$ .  $\mathbf{S}^*$  is a vector subspace characterizing these motions that is defined by

$$\mathbf{S}^* = \text{Ker } \mathbf{L}_s \quad (3.23)$$

where each column characterizes a motion of  $\mathbf{v}_{s_i}^*$  type. Thus it can be defined that a set of independent and compatible constraints satisfying  $\mathbf{s}(\mathbf{r}) - \mathbf{s}^* = 0$  builds up a virtual linkage between the sensor and the sensed object [SELB91]. It should be noticed that the vector subspace  $\mathbf{S}^*$  allows us to characterize virtual links since  $\mathbf{s}(\mathbf{r}) - \mathbf{s}^* = 0 \forall t$  implies that  $\dot{\mathbf{s}} = 0$ . Considering  $m$  as the rank of the interaction matrix,  $N$  the rank of  $\mathbf{S}^*$  is equal to  $n - m$ . This rank  $N$  defines the class of the virtual link. Consequently, the class of a virtual linkage corresponds to the number

of DOF, in translation or rotation, that are not constrained to complete the task since  $m$  and  $N$  complement each other.

### 3.3 Conclusion

Sensor-based control has been researched since several decades and from different perspectives, as illustrated by the vast literature supporting this field. This framework is nowadays widely developed in connection with the emergence of exteroceptive sensors and allows reactive control of a robot to sensory stimuli at a relatively low computational cost. The vision and touch senses are the common sensory feedback used in this framework. The well-known *visual servoing* paradigm, alone, illustrates the wide range of applications of sensor-based control. Robots' physical interactions (*e.g.*, grasping tasks) also benefit from the sensor-based approach through paradigms such as *proximity servoing* or *tactile servoing*, that are centered on the range and touch sensing. Such paradigms in these fields have led to impressive developments and prove their efficiency solving specific problems.

Nonetheless, the sensor-based approach remains surprisingly unused in robot audition, despite the convincing results obtained when using the touch or vision modality. However, it can be shown that expressing the problem of auditory perception through a task function can greatly simplify the interactions based on sound. The task function expresses a voluntary motion of a robot to interact with a given sound source. The auditory feedback is used to position the robot with respect to one or several sound sources. The resulting approach, AS, relies principally on the variation of auditory cues to steer the robot motion. Auditory cues variation appears to be more robust to realistic environments than their intrinsic values. AS is thus a dedicated control framework centered on hearing perception for robots, unlike classic sound localization frameworks that are adapted to machines in a broad sense.

The main tool building up AS is the closed-loop control ensued from the definition of the task function formalism. The resulting control scheme depends principally on the modeling of the task Jacobian, that characterizes the relationship between the robot motion and the variation of auditory features. The task's Jacobian models accurately the latter relationship while ensuring some given conditions. The first condition concerns the task admissibility that is generally ensured as soon as the task's Jacobian is invertible, while the second one concerns the stability of the control system that guarantees the convergence of this system towards the global solution of a given task. The development of auditory control schemes using this type of formalism is the main thread of the next two chapters.

## Chapter 4

# ILD-based aural servo

The first modeling and applications of the AS framework are introduced in the following chapter. More specifically, this chapter concerns the development of sensor-based control based on ILD. ILDs are analyzed from a binaural perspective, considering a set of two microphones in a free-field area. In this chapter and the following dealing with ITDs, the head scattering effect between the microphones is voluntarily omitted. The issues related to the scattering effect of head mounted systems are introduced later in Chapter 6. The results developed here are validated on a mobile robot endowed with two free-field microphones.

In the robot audition literature, surprisingly, when considering free-field conditions, the ILD is seldom exploited compared to the prominence of methods based on ITDs. Only few works such as [BG05] or [CCW06] try to develop localization systems using this cue. The main challenge about ILD cues comes from the fact that they cannot be explicitly associated to an exploitable source location, when using two microphones. Actually, in planar scenes, iso-ILD lies on a circle on which the source may be located. From this information, explicit source locations or directions cannot be deduced. As a result, in [BG05], the authors consider four microphones in order to infer the sound location, while in [CCW06], the authors use the intersection between ILD and ITD cues to extract the source position. Furthermore, both of these studies point out the high sensitivity of ILDs to reverberation. As a consequence, most free-field localization approaches are based on ITDs, when considering binaural sensors. On the contrary, in this chapter, we develop a control scheme, based on ILD measurements derived from acoustic properties of sound propagation in free field conditions. This contribution emphasizes the fact that AS paradigm opens perspectives on the utilization of auditory cues that are hardly exploitable in localization-based paradigms. To integrate ILDs in AS, the core idea of the method presented in this chapter consists in linking the variation of ILDs to the robot motion through an interaction matrix.

For this purpose, a geometrical analysis is performed first, in order to establish the link between an ILD measurement and a sound source location in Section 4.1.2. This analysis also draws the limits and constraints related to ILD cues.

From this preliminary analysis, in Section 4.2 the ILD variation is derived with respect to the motion of a pair of microphones, so that an interaction matrix related

to this cue can be inferred. Then this result is integrated into a control scheme, which is analyzed with respect to its stability properties, and the achievable positioning tasks. The proposed framework is eventually validated through several experiments on a mobile robots in real world environments.

In the last segment of this chapter in Section 4.3, we also integrate a distance cue to the control scheme, the absolute sound energy measured by the microphones. Following the same process as in the previous section, the interaction matrix related to this cue is computed. The feedback control is finally validated in various situations and acoustic conditions on a mobile robot.

## 4.1 ILD modeling

### 4.1.1 Scene configuration

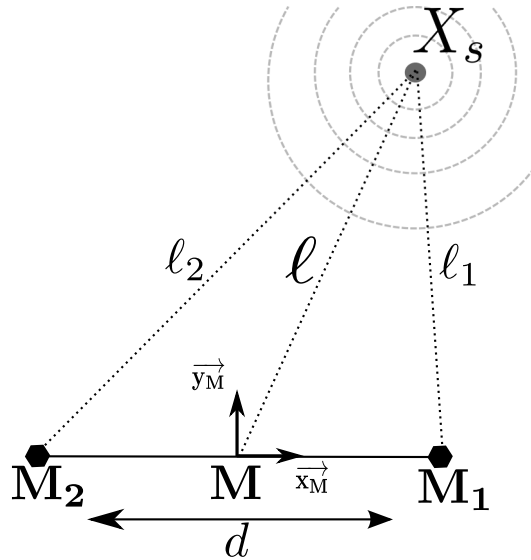


Figure 4.1: Geometric configuration of the considered problem, that includes a source  $X_s$  emitting a spherical sound wave, and a pair of microphones  $M_1$  and  $M_2$ .

In a first phase, let us consider the ideal case of a motorized pair of microphones  $M_1$  and  $M_2$  in an area free of obstacle. The free-field microphones  $M_1$  and  $M_2$  are separated by a distance  $d$  as illustrated in Figure 4.1. An omni-directional pointwise sound source  $X_s$  is continuously emitting a spherical sound wave  $a(t)$ . We also assume that the medium through which the sound travels is uniform and that linear acoustics hypotheses hold. Furthermore, we consider a frame  $\mathcal{F}_m(\mathbf{M}, \vec{x}_M, \vec{y}_M, \vec{z}_M)$  originating from the midpoint  $\mathbf{M}$  of the pair of microphones. Since we are considering a planar scene, we have  $z_M = 0$ . In this frame, the Cartesian coordinates of each microphone are respectively  $M_1(\frac{d}{2}, 0)$  and  $M_2(-\frac{d}{2}, 0)$ . The sound source  $X_s(x_s, y_s)$  is located at a distance  $l_i$  from each microphone  $M_i$ . Considering the geometric

model in Figure 4.1, the distances  $\ell_i$  are respectively given by

$$\begin{cases} \ell_1 = \sqrt{(x_s - d/2)^2 + y_s^2} \\ \ell_2 = \sqrt{(x_s + d/2)^2 + y_s^2} \end{cases} \quad (4.1)$$

Eventually, let  $\ell$  be the distance between  $\mathbf{M}$  and  $\mathbf{X}_s$ , that is then immediately related to the source position through:

$$\ell = \sqrt{x_s^2 + y_s^2} \quad (4.2)$$

In this configuration, the microphones are endowed with 3 DOF in the horizontal plane, that are the translations along  $\vec{x}_M$  and  $\vec{y}_M$  axis, and the rotation around  $\vec{z}_M$ .

### 4.1.2 Geometrical properties of the ILD

Although ILDs are mainly defined in the high frequencies of the signals measured by the microphones, one can infer ILDs directly from the acoustic properties of sound propagation in a similar process as [BG05]. Indeed, let us consider that  $a(t)$ , the signal generated by the sound source, is perceived by  $\mathbf{M}_i$ , for a frame of length  $w$ , under a high signal-to-noise ratio. In this condition, the signal at each microphone  $\mathbf{M}_i$  is defined as:

$$x_i(t) \propto \frac{a(t - \frac{\ell_i}{c})}{\ell_i}. \quad (4.3)$$

where  $\frac{\ell_i}{c}$  expresses the sound propagation delay (see spherical sound wave propagation in Chapter 1.2.1). In the following, without loss of generality of our model we consider a unitary proportional gain. By integrating (4.3) over the frame of length  $w$ , the energy received by each microphone is defined as follows:

$$E_i = \int_{t=0}^w |x_i(t)|^2 dt = \frac{1}{\ell_i^2} \int_{t=0}^w a^2 \left( t - \frac{\ell_i}{c} \right) dt \quad (4.4)$$

Equation (4.4) characterizes the inverse-square law property inherent to spherical and uniform sound propagation as supposed in Section 4.1. The ILD  $\rho$  between the two microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$  is then calculated from the ratio:

$$\rho = \frac{E_1}{E_2} = \frac{\ell_2^2 \int_{t=0}^w a^2 \left( t - \frac{\ell_1}{c} \right) dt}{\ell_1^2 \int_{t=0}^w a^2 \left( t - \frac{\ell_2}{c} \right) dt}. \quad (4.5)$$

Assuming that during  $w$ , the perceived sound signal does vary little between the two microphones, one can deduce that  $\int_{t=0}^w a^2 \left( t - \frac{\ell_1}{c} \right) dt \approx \int_{t=0}^w a^2 \left( t - \frac{\ell_2}{c} \right) dt$ . Actually, when the sound source is close to the microphones, we have  $\frac{\ell_i}{c} \ll w$ . And conversely, when the source is far,  $\ell_1 \approx \ell_2$  since  $\ell_i \gg d$ . These observations are also consistent with ITDs that are generally lesser than 1ms for realistic inter-microphones distances. In comparison,  $w$  has a length of few tenths to hundreds ms. Consequently the ILD  $\rho$  can be simplified without significant loss of accuracy as:

$$\rho = \frac{\ell_2^2}{\ell_1^2} \quad (4.6)$$



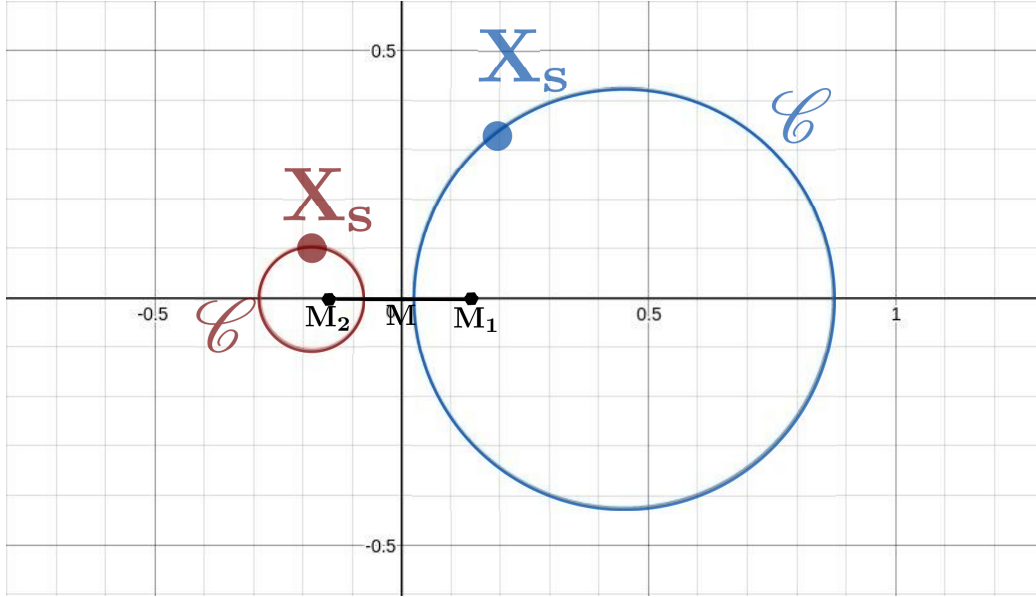


Figure 4.2: For a given sound source  $\mathbf{X}_s$ , the ILD refers geometrically to a circle  $\mathcal{C}$ , for which the *front-back ambiguity* should be solved.

Hence, from (4.6) and (4.5) the following relationship appears:

$$E_1 \ell_1^2 = E_2 \ell_2^2. \quad (4.7)$$

From these relationships, one can infer the sound location from an ILD measurement. With this aim in mind, by injecting the expression of each  $\ell_i$  with respect to  $\mathbf{X}_s$  in (4.7), we obtain

$$E_1 \left( x_s^2 - dx_s + \frac{d^2}{4} + y_s^2 \right) = E_2 \left( x_s^2 + dx_s + \frac{d^2}{4} + y_s^2 \right). \quad (4.8)$$

This relationship can also be expressed as

$$\left( x_s - \frac{d}{2} \frac{E_1 + E_2}{E_1 - E_2} \right)^2 + y_s^2 + \frac{d^2}{4} - \frac{d^2 (E_1 + E_2)^2}{4 (E_1 - E_2)^2} = 0. \quad (4.9)$$

Finally (4.9) reduces to the equation of a circle  $\mathcal{C}$  characterized by

$$(x_s - c_x)^2 + y_s^2 - \frac{E_1 E_2 d^2}{(E_1 - E_2)^2} = 0 \quad (4.10)$$

where  $c_x = \frac{d}{2} \frac{E_1 + E_2}{E_1 - E_2}$ . This result states that the sound source  $\mathbf{X}_s$  is located "somewhere" on the circle  $\mathcal{C}$ , illustrated in Figure 4.2, centered on the point  $(c_x, 0)$  with a radius of  $c_r = d \left| \frac{\sqrt{E_1 E_2}}{(E_1 - E_2)} \right|$ . In 3D scenes, the circle is changed into a sphere, generated by turning around the interaural axis, that induces the *torus of confusion* when using both interaural cues (*i.e.*, ILD and ITD), as outlined in Chapter 2.1. Moreover the latter result implies that the circle  $\mathcal{C}$  exists only if  $E_1 \neq E_2$ . Otherwise

the circle degenerates into a straight line. Actually, when  $E_1 = E_2$ , with a similar manipulation as before (4.7) reduces to:

$$2dx_s = 0, \quad (4.11)$$

that corresponds to the bisector of the microphone pair.

At last, the position of the circle with respect to the microphones and interaural axis can be analyzed. Firstly, the circle  $\mathcal{C}$  is always on the same lateral side of the microphones. Indeed  $|c_x| \geq c_r$  since  $|c_x| - c_r = \frac{d}{2|E_1 - E_2|}(\sqrt{E_1} - \sqrt{E_2})^2$  which is a degree-two polynomial identity. Thus, for all sources  $\mathbf{X}_{s_i}$  that belong to the same circle  $\mathcal{C}$ ,  $\text{sign}(x_{s_i})$  is the same. It can thus be said from ILD measurements if the source is on the left or right side of the microphones. But on the other hand, one cannot infer if the sound source is in the front or in the back side of the microphones since (4.10) imposes the circle  $\mathcal{C}$  to be centered on a point  $(c_x, c_y = 0)$ . This observation can be related to the well-known *front-back ambiguity* although no exact azimuth direction nor distance can be extracted from ILDs.

## 4.2 A typical ILD-based interaction

### 4.2.1 ILD interaction matrix

For now on, this section is about linking the robot motion to the ILD variation, so that an interaction matrix can be extracted. Indeed building the interaction matrix is one of the foundation of AS. In the configuration and analysis depicted above, one can relate the ILD variation to the microphones motion. From (4.6), the variation of an ILD can be obtained from the time variation of  $\rho$  that is given by:

$$\dot{\rho} = \frac{d}{dt} \left( \frac{\ell_2^2}{\ell_1^2} \right) = 2 \frac{\ell_2 \dot{\ell}_2 \ell_1 - \dot{\ell}_1 \ell_2^2}{\ell_1^3}, \quad (4.12)$$

that can be equally written as

$$\dot{\rho} = 2 \left( \frac{\ell_2 \dot{\ell}_2}{\ell_1^2} - \frac{\dot{\ell}_1}{\ell_1} \rho \right). \quad (4.13)$$

Consequently by replacing each  $\ell_i$  by its value given in (4.1) one obtain:

$$\dot{\rho} = \frac{\dot{x}_s(2x_s + d) + 2y_s \dot{y}_s}{\ell_1^2} - \frac{\dot{x}_s(2x_s - d) + 2y_s \dot{y}_s}{\ell_1^2} \rho. \quad (4.14)$$

Equation (4.14) can also be expressed as a matrix relationship:

$$\dot{\rho} = \mathbf{L}_\rho \mathbf{v}_M \quad (4.15)$$

where  $\mathbf{L}_\rho$  is the interaction matrix as introduced in Chapter 3.2.2, and  $\mathbf{v}_M = (v_x, v_y, v_z, \omega_x, \omega_y, \omega_z)$  the spatial velocity of  $\mathcal{F}_m$  with respect to the world expressed in  $\mathcal{F}_m$ . From the basic kinematic equation [CH06]

$$\dot{\mathbf{X}}_s = -\mathbf{v}_s - \boldsymbol{\omega}_s \times \mathbf{X}_s \Leftrightarrow \begin{cases} \dot{x}_s = -v_x - \omega_y z_s + \omega_z y_s \\ \dot{y}_s = -v_y - \omega_z x_s + \omega_x z_s \\ \dot{z}_s = -v_z - \omega_x y_s + \omega_y x_s \end{cases} \quad (4.16)$$

which relates the velocity of a 3-D point  $\mathbf{X}_s$  to the sensor spatial velocity  $\mathbf{v}_M$ , (4.14) becomes

$$\dot{\rho} = v_x \frac{2x_s(\rho-1) - d(\rho+1)}{\ell_1^2} + v_y \frac{2y_s(\rho-1)}{\ell_1^2} + \omega_z \frac{y_s d(\rho+1)}{\ell_1^2}. \quad (4.17)$$

Eventually referring to (4.15), the interaction matrix related to  $\rho$  in the planar scene that we consider is identified as:

$$\mathbf{L}_\rho = \begin{bmatrix} \frac{2x_s(\rho-1) - d(\rho+1)}{\ell_1^2} & \frac{2y_s(\rho-1)}{\ell_1^2} & 0 & 0 & 0 & \frac{y_s d(\rho+1)}{\ell_1^2} \end{bmatrix}. \quad (4.18)$$

In connection with the geometric configuration, we consider only the non-zero terms of this matrix so that  $k = 3$ . In this planar scene,  $\mathbf{v}_M$  is replaced by  $\mathbf{u}_M = (v_x, v_y, \omega_z)$  so that  $\mathbf{L}_\rho$  can be reduced to

$$\mathbf{L}_\rho = \begin{bmatrix} \frac{2x_s(\rho-1) - d(\rho+1)}{\ell_1^2} & \frac{2y_s(\rho-1)}{\ell_1^2} & \frac{y_s d(\rho+1)}{\ell_1^2} \end{bmatrix}. \quad (4.19)$$

For more consistence with the scene configuration and the reference frame  $\mathcal{F}_m$ , the interaction matrix  $\mathbf{L}_\rho$  is redefined as a function of  $\ell$  as

$$\mathbf{L}_\rho = \begin{bmatrix} \frac{2x_s(\rho-1) - d(\rho+1)}{\ell^2 + \frac{d^2}{4} - dx_s} & \frac{2y_s(\rho-1)}{\ell^2 + \frac{d^2}{4} - dx_s} & \frac{y_s d(\rho+1)}{\ell^2 + \frac{d^2}{4} - dx_s} \end{bmatrix}, \quad (4.20)$$

from (4.1) that implies that  $\ell_1^2 = \ell^2 + \frac{d^2}{4} - dx_s$ . This matrix contains terms that are unknown, namely the source position  $x_s$ ,  $y_s$ , and  $\ell$  that is dependent on the two previous parameters as stated in (4.2). The approximated interaction matrix is then

$$\widehat{\mathbf{L}}_\rho = \begin{bmatrix} \frac{2\widehat{x}_s(\rho-1) - d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{2\widehat{y}_s(\rho-1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{\widehat{y}_s d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} \end{bmatrix}. \quad (4.21)$$

However, regarding the interaction matrix, the approximation of these parameters should be chosen carefully. Indeed,  $\widehat{\mathbf{L}}_\rho$  has elements of infinite value as soon as  $\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s = 0$ . Thus  $\widehat{y}_s$  and  $\widehat{x}_s$  should be selected so that  $\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s > 0$ . Furthermore from (4.10) or (4.11) as soon as  $\widehat{x}_s$  (respectively  $\widehat{y}_s$ ) is set, one can immediately deduce  $\widehat{y}_s$  (respectively  $\widehat{x}_s$ ) and then  $\widehat{\ell} = \sqrt{\widehat{x}_s^2 + \widehat{y}_s^2}$ . Nonetheless, the questioning about fixing the values of these parameters remains open. We address this issue in Section 4.2.4, by considering the Lyapunov stability of the controller as the matrix approximation criterion. As stated in the previous chapter, the Lyapunov global asymptotic stability of the control scheme ensures the convergence of the controller [CH06].

The latter interaction matrix (4.21) can also be expressed for ILD measured in dB. Indeed when considering the ILD  $\rho_{dB}$  defined as

$$\rho_{dB} = 10 \log_{10} \rho, \quad (4.22)$$

the time variation of the ILD expressed in dB is given by

$$\dot{\rho}_{dB} = \frac{10\dot{\rho}}{\rho \ln(10)}. \quad (4.23)$$

By injecting (4.15) into the latter equation, the interaction matrix related to  $\rho_{dB}$  is therefore defined as

$$\mathbf{L}_{\rho_{dB}} = \frac{10}{\rho \ln(10)} \mathbf{L}_{\rho}. \quad (4.24)$$

Eventually the approximated interaction matrix  $\widehat{\mathbf{L}}_{\rho_{dB}}$  can be explicitly developed as

$$\widehat{\mathbf{L}}_{\rho_{dB}} = \frac{10}{\ln(10)} \begin{bmatrix} 2\widehat{x}_s(10^{\frac{\rho_{dB}}{10}} - 1) - d(10^{\frac{\rho_{dB}}{10}} + 1) & 2\widehat{y}_s(10^{\frac{\rho_{dB}}{10}} - 1) & \widehat{y}_s d(10^{\frac{\rho_{dB}}{10}} + 1) \\ 10^{\frac{\rho_{dB}}{10}} (\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s) & 10^{\frac{\rho_{dB}}{10}} (\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s) & 10^{\frac{\rho_{dB}}{10}} (\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s) \end{bmatrix}. \quad (4.25)$$

It should be noticed that the expressions of  $\widehat{\mathbf{L}}_{\rho_{dB}}$  in (4.25) and  $\widehat{\mathbf{L}}_{\rho}$  in (4.21) lead to equivalent control loop. Depending on how ILDs are expressed one might choose the corresponding interaction matrix. In this thesis, we based our development on  $\widehat{\mathbf{L}}_{\rho}$  that has a simpler form, and does not require additional calculation.

## 4.2.2 Control scheme

From the latter interaction matrix, we can then formulate a task which consists in positioning the robot so that given conditions characterized by the acoustic features are satisfied. The task is performed by considering a single ILD measurement  $\rho(t)$  extracted from the sound signal and by minimizing the error  $\|e(t)\|$  in the same process as described in Chapter 3.2.1. Hence the task is characterized by an error

$$e(t) = \rho(t) - \rho^* \quad (4.26)$$

where  $\rho^*$  denotes the measurements for the desired ILD value. From (4.15) and with a perfectly known interaction matrix as (4.20), it is possible to design a control scheme with a purpose of exponential decoupled decrease of the error [SELB91] in which the error follows

$$\dot{e} = -\lambda e, \quad (4.27)$$

where  $\lambda > 0$  is a gain that tunes the time to convergence. As stated in the previous section the actual matrix  $\mathbf{L}_{\rho}$  is not fully known since some of its elements depend on the sound source position. Hence we design a control scheme governed by the approximation of the interaction matrix  $\widehat{\mathbf{L}}_{\rho}$  in (4.21), in which the velocity of the microphones is computed from (4.15) [CH06] as

$$\mathbf{u}_{\mathbf{M}} = -\lambda \widehat{\mathbf{L}}_{\rho}^+ e. \quad (4.28)$$

In this latter equation  $\widehat{\mathbf{L}}_{\rho}^+ \in \mathbb{R}^{3 \times 1}$  is the Moore-Penrose pseudo-inverse of  $\widehat{\mathbf{L}}_{\rho}$  (see Chapter 3.2.1). Thus the control scheme can be explicitly written as

$$\mathbf{u}_{\mathbf{M}} = -\lambda \frac{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s}{\widehat{N}_1 + \widehat{N}_2 + \widehat{N}_3} \begin{bmatrix} 2\widehat{x}_s(\rho - 1) - d(\rho + 1) \\ 2\widehat{y}_s(\rho - 1) \\ \widehat{y}_s d(\rho + 1) \end{bmatrix} (\rho - \rho^*), \quad (4.29)$$

where

$$\begin{cases} \widehat{N}_1 = 2\widehat{x}_s(\rho - 1) - d(\rho + 1) \\ \widehat{N}_2 = 2\widehat{y}_s(\rho - 1) \\ \widehat{N}_3 = \widehat{y}_s d(\rho + 1). \end{cases} \quad (4.30)$$

### 4.2.3 Task analysis

From the interaction matrix in (4.21), it is then possible to achieve positioning tasks with the control scheme (4.29). However it is necessary to analyze and understand what is the task performed through the control scheme developed in the previous section. As introduced in Chapter 3.2.4, we make use of virtual linkages in order to exhibit the motions that should be expected from the control scheme. In this approach, a vector subspace  $\mathbf{S}^*$  is defined so that it represents all motions of the sensors for which the ILD  $\rho$  remains constant. To some extent this approach can be seen as similar to the development produced in Section 4.1.2, in which  $\mathcal{C}$  characterizes the motions of the source that give the same ILD measurement. When considering only one sound source,  $\mathbf{L}_\rho$  given by (4.20) is a rank-one matrix. Consequently  $\mathbf{S}^* \in \mathbb{R}^{3 \times 2}$  implies a virtual link of class 2:

$$\mathbf{S}^* = \text{Ker } \mathbf{L}_\rho. \quad (4.31)$$

Hence the subspace is defined as

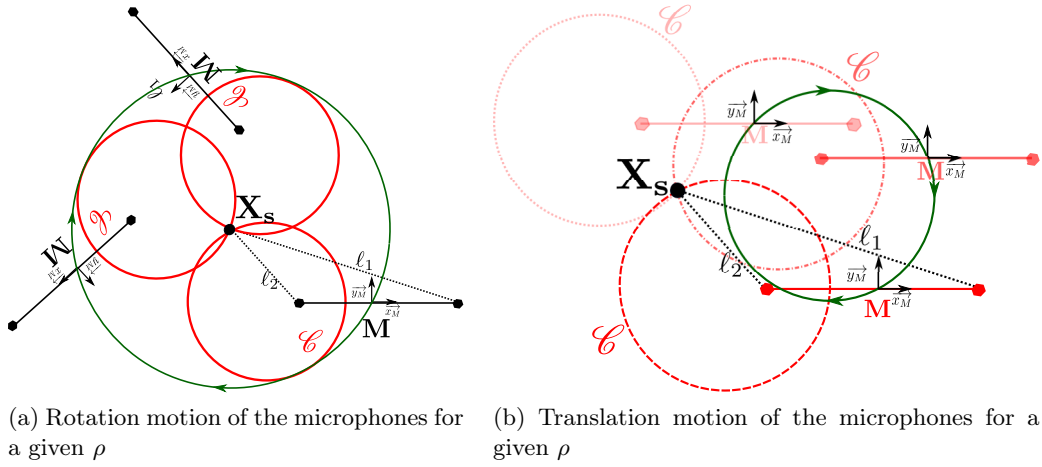
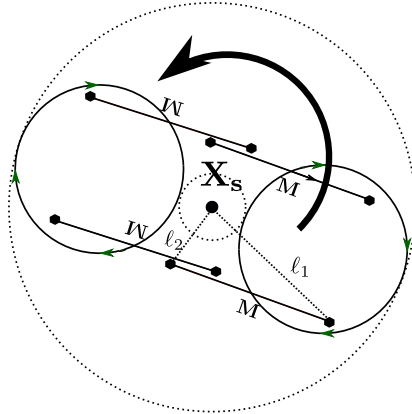
$$\mathbf{S}^* = [\mathbf{u}_{\mathbf{M}_1}^* \quad \mathbf{u}_{\mathbf{M}_2}^*] = \begin{bmatrix} v_{x1} & v_{x2} \\ v_{y1} & v_{y2} \\ \omega_{z1} & \omega_{z2} \end{bmatrix}. \quad (4.32)$$

When considering the interaction matrix  $\mathbf{L}_\rho$  given in (4.20), we obtain:

$$\mathbf{S}^* = \begin{bmatrix} y_s d(1 + \rho) & 2y_s(\rho - 1) \\ 0 & d(\rho + 1) + 2x_s(1 - \rho) \\ d(\rho + 1) + 2x_s(1 - \rho) & 0 \end{bmatrix}. \quad (4.33)$$

Although it is not trivial to infer the geometrical motions that come out from (4.31), they can be deduced by geometric construction. The first vector  $\mathbf{v}_{\mathbf{M}_1}^*$  refers to a circular motion of the microphones for which  $\rho = \frac{\ell_2^2}{\ell_1^2}$  remains constant. Thence it can be immediately deduced that this vector refers to a circular motion around the source, in which  $\ell_1$  and  $\ell_2$  remain constant as depicted in Figure 4.3. Concretely, from a given orientation of the microphones with respect to the source, moving around the sound source by keeping the same relative orientation does not change the ILD value. Indeed in this configuration the circles  $\mathcal{C}$  all intersect on the actual source position. The second vector  $\mathbf{v}_{\mathbf{M}_2}^*$  refers to a translation. The microphones motion corresponds to a circle that imply a relative motion of  $\mathbf{X}_s$  on  $\mathcal{C}$ . By combining these two motions, we obtain an annulus  $\mathcal{D}$  characterized by an inner radius  $r_1 = \left| \frac{d}{2} - \frac{d\sqrt{\rho^*}}{1+\sqrt{\rho^*}} \right|$  and an outer radius  $r_2 = r_1 + 2c_r$ , as depicted in Figure 4.4. In the case when  $\rho^* = 1$  (i.e.,  $\ell_1 = \ell_2$ ), we get  $r_1 = 0$  and  $r_2 = \infty$  which results in an infinite area  $\mathcal{D}$  around the source. This particular configuration lead to the same result as for ITD-based tasks. As a consequence any linear combination of  $\mathbf{v}_{\mathbf{M}_1}^*$  and  $\mathbf{v}_{\mathbf{M}_2}^*$  implies infinite pose for which  $\rho = \rho^*$ , which let us build the following proposition:

**Proposition 1** *For each position of the microphones  $\mathbf{M} \in \mathcal{D}$ , there exists at least one orientation  $\theta_{\mathbf{M}}$  of the microphones so that  $\rho = \rho^*$ .*


 Figure 4.3: Motions implied by  $S^*$ 

 Figure 4.4: The admissible poses of the microphones for a given  $\rho$  shape an annulus

Hence, a typical task involving ILDs mainly consists in orienting the robot towards a given direction with respect to the source location while being or reaching  $\mathcal{D}$ . Interestingly, the task of facing a sound source (*i.e.*,  $\rho^* = 1$ ), that is of high interest in robot audition, can be achieved independently of microphones location since  $\mathcal{D} = \infty$ . Hence, this typical task mainly consists in orienting the robot towards a given direction with respect to the source location. This result implies that controlling only one DOF, namely the rotation  $\omega_z$  is enough to complete any task involving  $\rho^* = 1$ . Hence, the task defined by ILDs in AS is consistent with the exploitation of ILD cues for human listeners, since interaural cues are physiologically related to an azimuth direction as exposed in Chapter 2.1.2. However as stated in Section 4.1.2, this cue does not explicitly provide an azimuth direction (*i.e.*, circle  $\mathcal{C}$ ), which makes it complex to exploit for sound source localization algorithms, while AS can still benefit from this information.

#### 4.2.4 Stability analysis

In this part, we study the stability conditions of the closed-loop control system with respect to the approximated interaction matrix  $\widehat{\mathbf{L}}_\rho$ . The stability properties tailor the approximation of the unknown parameters in the interaction matrix  $\widehat{\mathbf{L}}_\rho$ , since they condition the convergence of the control scheme (see Chapter 3.2.2). Since  $\widehat{\mathbf{L}}_\rho \in \mathbb{R}^{1 \times 3}$ , the closed-loop control system is globally asymptotically stable when

$$\mathbf{L}_\rho \widehat{\mathbf{L}}_\rho^+ > 0. \quad (4.34)$$

In a first step, in order to ease the comprehension and the reading of this analysis, let us redefine the interaction matrix  $\widehat{\mathbf{L}}_\rho$  as

$$\widehat{\mathbf{L}}_\rho = \left[ \begin{array}{ccc} \frac{\widehat{N}_1}{(\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s)} & \frac{\widehat{N}_2}{(\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s)} & \frac{\widehat{N}_3}{(\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s)} \end{array} \right], \quad (4.35)$$

where  $\widehat{N}_1, \widehat{N}_2, \widehat{N}_3$  are defined from (4.30). Similarly  $\mathbf{L}_\rho$  is defined with  $N_1, N_2, N_3$ . From this convention of writing, (4.34) becomes

$$\mathbf{L}_\rho \widehat{\mathbf{L}}_\rho^+ = \frac{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s}{\ell^2 + \frac{d^2}{4} - dx_s} \frac{N_1 \widehat{N}_1 + N_2 \widehat{N}_2 + N_3 \widehat{N}_3}{\widehat{N}_1^2 + \widehat{N}_2^2 + \widehat{N}_3^2} > 0. \quad (4.36)$$

Knowing that  $\widehat{\ell}_1^2 > 0$  or equivalently  $\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s > 0$ , (4.36) is true if and only if  $N_1 \widehat{N}_1 + N_2 \widehat{N}_2 + N_3 \widehat{N}_3$  is greater than 0. Thus one can rewrite the condition by developing the latter expression as:

$$4y_s \widehat{y}_s (\rho - 1)^2 + y_s \widehat{y}_s d^2 (\rho + 1)^2 + (2x_s (\rho - 1) - d(\rho + 1))(2\widehat{x}_s (\rho - 1) - d(\rho + 1)) > 0 \quad (4.37)$$

It is then clear that when  $E_1 = E_2$  (*i.e.*,  $x_s = 0$  and  $\rho = 1$ ), the stability is guaranteed as soon as  $sign(y_s) = sign(\widehat{y}_s)$ . However when  $\rho \neq 1$ , some supplementary analysis is required. To do so, let us rearrange the condition (4.37) to :

$$(\rho - 1)^2 (4y_s \widehat{y}_s + 4x_s \widehat{x}_s) + (\rho + 1)^2 d^2 (1 + y_s \widehat{y}_s) - (\rho + 1)(\rho - 1) 2d(x_s + \widehat{x}_s) > 0. \quad (4.38)$$

By dividing (4.38) by  $(\rho + 1)^2$ , we obtain a quadratic function  $f(Z)$  depending on the ratio  $\rho$  that is

$$f(Z) = Z^2 (4y_s \widehat{y}_s + 4x_s \widehat{x}_s) - 2Zd(x_s + \widehat{x}_s) + d^2 (1 + y_s \widehat{y}_s) > 0, \quad (4.39)$$

with  $Z = \frac{\rho - 1}{\rho + 1}$  and  $Z \in ] -1, 1[$ . It can be noticed that  $sign(Z) = sign(x_s)$  from the geometrical properties outlined in Section 4.1.2. Indeed we have  $x_s > 0$  when  $\rho > 1$ , and  $x_s < 0$  when  $\rho < 1$ . From this result, one can deduce that if  $\widehat{x}_s$  and  $\widehat{y}_s$  are perfectly known,  $f(Z)$  is symmetric with respect to the bisection of the microphones ( $y = 0$ ). Indeed, for opposite value of  $x_s$  (*i.e.*, inverse value of  $\rho$  and opposite value of  $Z$ ), we obtain  $f(z) = f(-z)$ . Although  $\widehat{x}_s$  and  $\widehat{y}_s$  are not known, we can still exploit this symmetry property. Let us consider two opposite values  $z_1$  and  $z_2$  characterized by

$$\begin{cases} f(z_1) = z_1^2 (4y_s \widehat{y}_s + 4x_s \widehat{x}_s) + 2dz_1(x_s + \widehat{x}_s) + d^2 (1 + y_s \widehat{y}_s) \\ f(z_2) = z_2^2 (4y_s \widehat{y}_s + 4x_s \widehat{x}_s) + 2dz_2(x_s + \widehat{x}_s) + d^2 (1 + y_s \widehat{y}_s) \end{cases} \quad (4.40)$$

and  $z_1 = -z_2$ . Actually the symmetric nature of  $f(Z)$  (i.e.,  $f(z_1) = f(z_2)$ ) is preserved if  $\hat{x}_s$  and  $\hat{y}_s$  are selected as constant and with the same sign as  $x_s$  and  $y_s$ , respectively. Under this condition, knowing that  $f(Z)$  is a parabola from its quadratic nature, the vertex of  $f(Z)$  is the point at  $(0, f(0))$ . As a consequence, to fulfill the condition (4.39), it is sufficient to demonstrate that (i),  $f(Z)$  is a convex parabola, and (ii), the vertex value  $f(0)$ , which is a minimum when (i) is ensured, is greater than 0.

(i). First, it can be demonstrated that  $f(Z)$  is convex if

$$(4y_s\hat{y}_s + 4x_s\hat{x}_s) > 0. \quad (4.41)$$

A sufficient condition to fulfill (4.41) consists in setting  $\text{sign}(\hat{y}_s) = \text{sign}(y_s)$  and  $\text{sign}(\hat{x}_s) = \text{sign}(x_s)$ , which has already been assumed.

(ii). In this context, the vertex value  $f(0)$  is greater than 0 when

$$d^2(1 + y_s\hat{y}_s) > 0. \quad (4.42)$$

The latter condition is immediately completed when  $\text{sign}(\hat{y}_s) = \text{sign}(y_s)$ .

Consequently the Lyapunov stability condition of  $\widehat{\mathbf{L}}_\rho$  expressed through (4.39) is ensured as soon as  $\text{sign}(\hat{y}_s) = \text{sign}(y_s)$  and  $\text{sign}(\hat{x}_s) = \text{sign}(x_s)$ . One can underline that  $\text{sign}(\hat{x}_s) = \text{sign}(x_s)$  is obtained directly by the measurement of the ILD  $\rho$ . Actually  $\text{sign}(x_s) = \text{sign}(\rho - 1)$ . However  $\text{sign}(\hat{y}_s) = \text{sign}(y_s)$  cannot be ensured from the ILD measurements because of the properties of the circle  $\mathcal{C}$  depicted in Section 4.1.2. The latter condition directly refers to the *front/back ambiguity* that is unfortunately inherent to the binaural cues. Despite the *front/back ambiguity*, these excellent stability conditions confirm that the exact sound source location is not required for the controller. The stability conditions are not restrictive at all and let us foresee an infinite convergence domain.

In the context depicted until now, we assumed ideal acoustic conditions in free-field, that simplify our model but that are unfortunately not realistic. Indeed as discussed in Chapter 1.2 and in Chapter 2.3, one main issue of robot audition concerns the consistency and the modeling of auditory features in perturbed environment. Nevertheless, this study emphasizes the potential robustness of the controller, that is tolerant to rough approximations of the interaction matrix. This interesting property allows us to envisage that the control scheme may also be robust to approximate acoustic modeling. This hypothesis is verified in the next section, that is articulated around the experimental validations of our approach in real world conditions.

## 4.2.5 Experimental validations

### 4.2.5.1 Preliminaries: robot modeling and control scheme

For the experiments, we consider a non-holonomic unicycle robot *Pioneer 3DX* endowed with two omnidirectional microphones as illustrated on Figure 4.5. Schematically, the system consists in a planar robot controlled in the horizontal plane with the



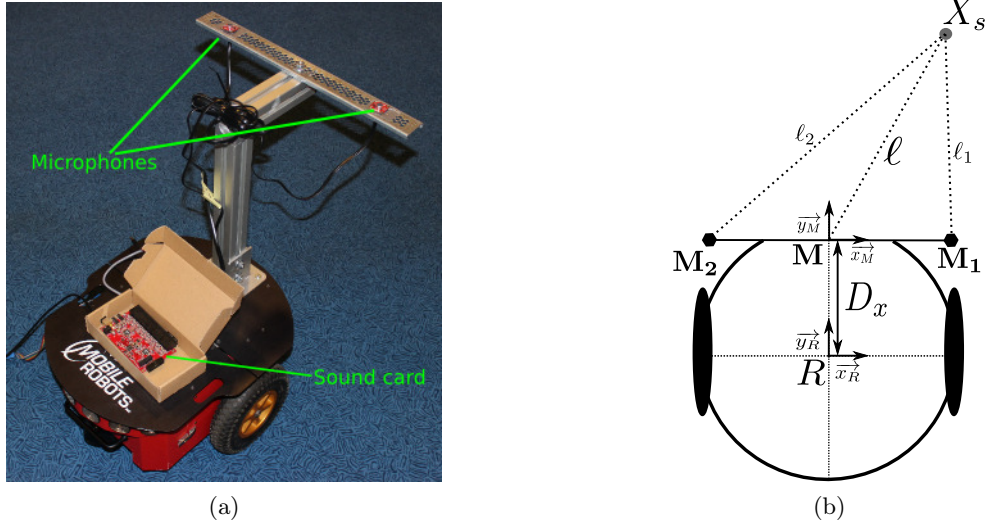


Figure 4.5: modeling of the robotic platform

two microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$ . In addition to  $\mathcal{F}_m$  attached to the microphones, we define the frame  $\mathcal{F}_r(\vec{x}_R, \vec{y}_R)$  attached to the robot.  $D_x$  denotes the distance between the center of the robot  $\mathbf{R}$  and the midpoint  $\mathbf{M}$  of the microphones. The robot can be controlled upon two DOF: the control input  $\dot{\mathbf{q}}$  is given by  $(u, \omega)$ , respectively the translation velocity along  $\vec{y}_R$  and the angular velocity around  $\vec{z}_R$ .

Consequently the relationship between the variation of the ILD  $\rho$  and the control input  $\dot{\mathbf{q}}$  is:

$$\dot{\rho} = \mathbf{J}_\rho \dot{\mathbf{q}} \quad (4.43)$$

where  $\mathbf{J}_\rho$  is the Jacobian feature matrix introduced in Chapter 3.2.1 that is equal

$$\mathbf{J}_\rho = \mathbf{L}_\rho \mathbf{J}_r, \quad (4.44)$$

$\mathbf{J}_r$  being the robot Jacobian. From the model given in Figure 4.5, the robot Jacobian is equal to

$$\mathbf{J}_r = \begin{bmatrix} 0 & D_x \\ 1 & 0 \\ 0 & 1 \end{bmatrix}. \quad (4.45)$$

Hence, the control scheme of the robot is (see Chapter 3)

$$\dot{\mathbf{q}} = -\lambda \widehat{\mathbf{J}}_\rho^+(\rho - \rho^*). \quad (4.46)$$

Furthermore, from the previous task analysis, we demonstrated that a task considering the ILD  $\rho$  mainly specifies a desired orientation of the robot w.r.t the sound source. Therefore a control input characterized by the angular velocity  $\omega$  only is sufficient to achieve any task involving one sound source. In the following, we consider

this particular case when  $u$  would always be equal to 0. Hence the Jacobian matrix  $\widehat{\mathbf{J}}_\rho$  reduces to

$$\widehat{\mathbf{J}}_\rho = \frac{D_x(2\widehat{x}_s(\rho - 1) - d(\rho + 1)) + \widehat{y}_s d(\rho + 1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s}, \quad (4.47)$$

and the control input is then

$$\dot{\mathbf{q}} = -\lambda \frac{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s}{D_x(2\widehat{x}_s(\rho - 1) - d(\rho + 1)) + \widehat{y}_s d(\rho + 1)} (\rho - \rho^*). \quad (4.48)$$

#### 4.2.5.2 Experimental results

The sets of experiments conducted in this section and the following are performed in a room with a reverberation time  $\text{RT}_{60} \approx 580$  ms. The microphones instrumenting the robot were connected to a sound card 8SoundsUSB [AC<sup>+</sup>] that processes the signal in real time. The sound card operates at a sampling frequency of 48 kHz, and provides frames of 256 samples. The sound energy is computed from 10 consecutive frames  $w$  (*i.e.*, 50 ms). The parameters given in Table 4.1 were used for all experiments. In a first step, we do not account for the *front-back ambiguity*, and thus assume that the sound source is in the front side of the robot (*e.g.*  $y_s > 0$ ).

$d$	0.31 m
$D_x$	0.3 m
$\widehat{y}_s$	1 m
$\widehat{x}_s$	$\text{sign}(\rho - 1) \times 1$ m
$\lambda$	0.5

Table 4.1: Experimental settings

The first experiment simply consists in facing the sound source. The desired ILD is thus set to  $\rho^* = 1$ . The sound source is a loudspeaker emitting a white Gaussian noise. The SNR at the desired pose is around 20 dB in presence of typical noise such as computer noise and ventilation in the room. As illustrated in Figure 4.6, the control scheme allows to face the sound source starting from a random orientation of the robot. The magnitude of the error follows an exponential decrease behavior as expected from the control scheme. It should also be noted that our approach is straightforward: the control input computed from (4.48) simply consists in a scalar corresponding to the angular velocity of the robot. Furthermore, our approach is performed without signal enhancement, modeling of the environment nor tracking, which explains the low computation cost of the control system.

#### 4.2.5.3 Addressing the front-back ambiguity

Although the previous experiment confirmed the validity of our approach, we assumed that the sound source was in the front side of the robot, which can be restrictive. The *front-back ambiguity* remains an issue that cannot be addressed from the

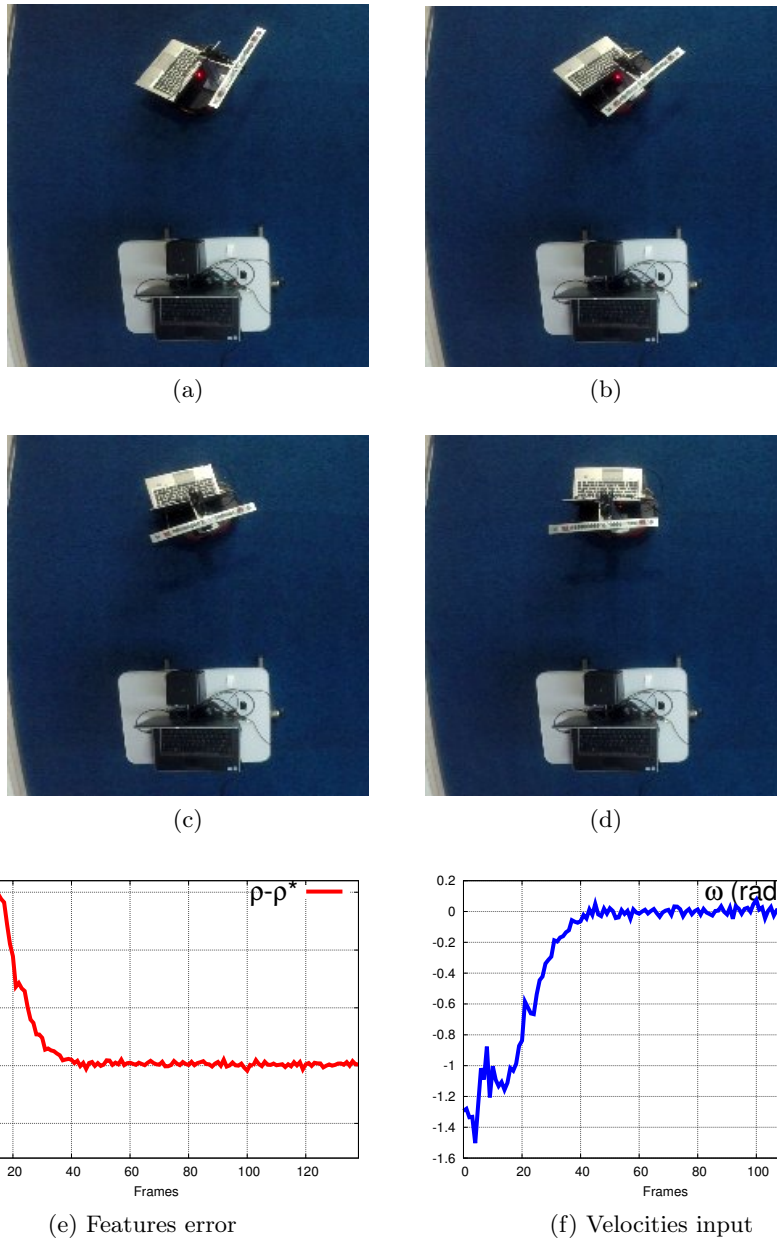


Figure 4.6: A typical task using the ILD  $\rho$  consists in facing the sound source from a random orientation

binaural cues. The development in Chapter 2.1.2 showed that human listeners disambiguate the location of the source from monaural cues (*i.e.*, action of the pinnae through the HRTFs) and from head motion. The first solution, based on monaural cues, cannot be applied to our context since the robot is endowed with free-field microphones, that is to say HRTFs depends only on the azimuth angle. Yet, the motion of the robot can be used in order to dissipate the ambiguity. A simple method would

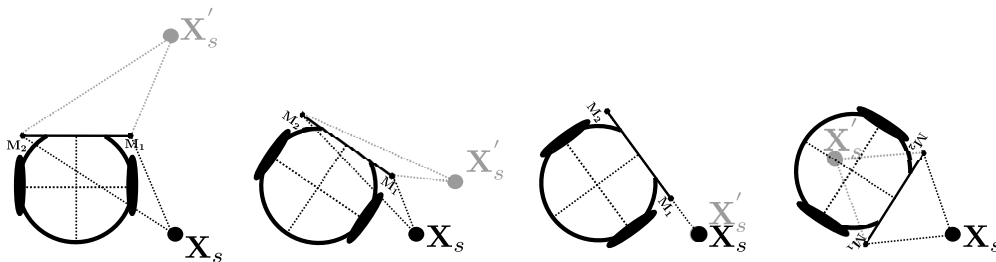


Figure 4.7: From left to right a position task that consists to face the sound source  $\mathbf{X}_s$  located on the back-side of the robot: because of the front-back ambiguity the robot turns with respect to the phantom source  $\mathbf{X}'_s$ . This behavior naturally allows to complete successfully the task, since the motion of the robot repositions the actual source on the front-side

consist in a state-machine inferring the side of the source from the evolution of the ILD with respect to the motion of the robot. This kind of approach has been applied in [MMWB15, MMB15] for instance. More sophisticated techniques have also been proposed in the literature such as stochastic filtering in [PDA12] or [NCVC16] that respectively use an unscented Kalman filter and mixture of Kalman filter. These two approaches fuse the motion of the robot to the cues measurements to infer the sound location without ambiguity.

Our approach is designed for situations when the sound source is in the front side (*i.e.*,  $y_s > 0$ ) which explains that the interaction matrix is parameterized with  $\hat{y}_s > 0$  so that the stability conditions are ensured. Hence, the control scheme may not be stable when the sound source is located on the back-side since  $\text{sign}(\hat{y}_s) \neq \text{sign}(y_s)$ , and the robot may not converge towards the desired pose. Under this hypothesis, the robot moves with respect to a phantom sound source (symmetric to the actual source) as illustrated in Figure 4.7. The robot moves towards an opposite direction to the one required to the completion of the task which increases the magnitude of the error  $e = \rho - \rho^*$ . This characteristic that could be seen as a limitation of the control scheme, turns out to be actually an advantage. Indeed, the stability conditions induce that a sufficient way to complete a given task is obtained when  $\text{sign}(\hat{y}_s) = \text{sign}(y_s)$ , that is, the source located in the front side since we fixed  $\hat{y}_s > 0$ . For ILD-based task that consists in turning the robot towards a given direction with respect to the source until the condition  $e = \rho - \rho^* = 0$  is met, it can be inferred that the latter condition cannot be fulfilled when the source is located in the back since the magnitude of the error increases. Consequently, the rotation motion will necessarily lead to  $y_s > 0$ . Concretely the magnitude of the error increases in a first phase, then decreases towards zero as soon as  $y_s > 0$  which enables the completion of the task. Thus, the robot always reaches a pose that corresponds to a configuration  $y_s > 0$ , in which the task can be completed similarly to the previous experiment. The *front-back ambiguity* is then inherently solved by our control scheme: without any assumption about the source location, it is possible to complete positioning tasks based on ILD measurements.

This result has been confirmed by experiments in which we observe the predicted behavior discussed above. The experiment illustrated by Figure 4.8, shows a first phase where the error of the task increases until  $y_s > 0$ . Subsequently, from this configuration, an exponential decrease of the error magnitude can be observed until the robot faces the sound source.

#### 4.2.5.4 Addressing the case of a moving sound source

Eventually, we can tackle the problem related to a moving sound source. Considering robot audition state-of-the-art, dealing with moving sources is a bottleneck of sound source localization, since it requires a tracking and a model of source motions. By contrast, AS is a closed-loop control, that naturally induces flexibility and reactivity to any modification of the acoustic perception, thanks to the real-time feedback. Hence, any motion of the sound source will have an impact on the control input. Indeed when the source moves laterally, the magnitude of the error  $e = \rho - \rho^*$  increases, which immediately generates a motion in order to reduce the error. Hence, this type of control scheme is particularly suitable for tracking a moving source. The experiment carried in Figure 4.9 sums up all the potential of AS. Initially, the source is in the back of the robot. This issue is correctly solved by the control scheme, and the robot is able to face the sound source. Afterwards, the sound source is moved laterally and arbitrarily while being accurately tracked by the robot in real time. The motion of the robot maintains the source in the *auditory fovea*. However, fast motions of the sound source may entail some delay in the control scheme to compensate the error. Nevertheless this error could be reduced by advanced control techniques. If we recall the principle of task function given in Chapter 3.2.1, the time derivative of  $\rho$  is exactly defined as

$$\dot{\rho} = \dot{e} = \mathbf{L}_e \mathbf{v}_M + \frac{\partial e}{\partial t}, \quad (4.49)$$

in which  $\frac{\partial e}{\partial t}$  characterizes the time variation of  $e$  due to the unknown source motion. Until now, to derive the control scheme from the ILD variation, we assumed a motionless source, considering  $\frac{\partial e}{\partial t} = 0$ . Thus, in order to avoid this tracking error,  $\frac{\partial e}{\partial t}$  should be integrated in the control scheme. With a purpose of exponential decoupled decrease of the error ( $\dot{e} = -\lambda e$ ), the velocity of the microphones is therefore defined as

$$\mathbf{v}_M = -\lambda \widehat{\mathbf{L}}_e^+ e - \widehat{\mathbf{L}}_e^+ \frac{\partial e}{\partial t}, \quad (4.50)$$

in which  $\widehat{\frac{\partial e}{\partial t}}$  is an estimation of the time variation of  $e$  due to the source motion. Eventually the variation of the error is obtained by injecting (4.50) into (4.49) as:

$$\dot{e} = -\lambda \mathbf{L}_e \widehat{\mathbf{L}}_e^+ e - \mathbf{L}_e \widehat{\mathbf{L}}_e^+ \frac{\partial e}{\partial t} + \frac{\partial e}{\partial t}. \quad (4.51)$$

From the latter equation it can be seen that the tracking error due to the motion of the source can be compensated under the condition that

$$\mathbf{L}_e \widehat{\mathbf{L}}_e^+ \frac{\partial e}{\partial t} = \frac{\partial e}{\partial t}. \quad (4.52)$$

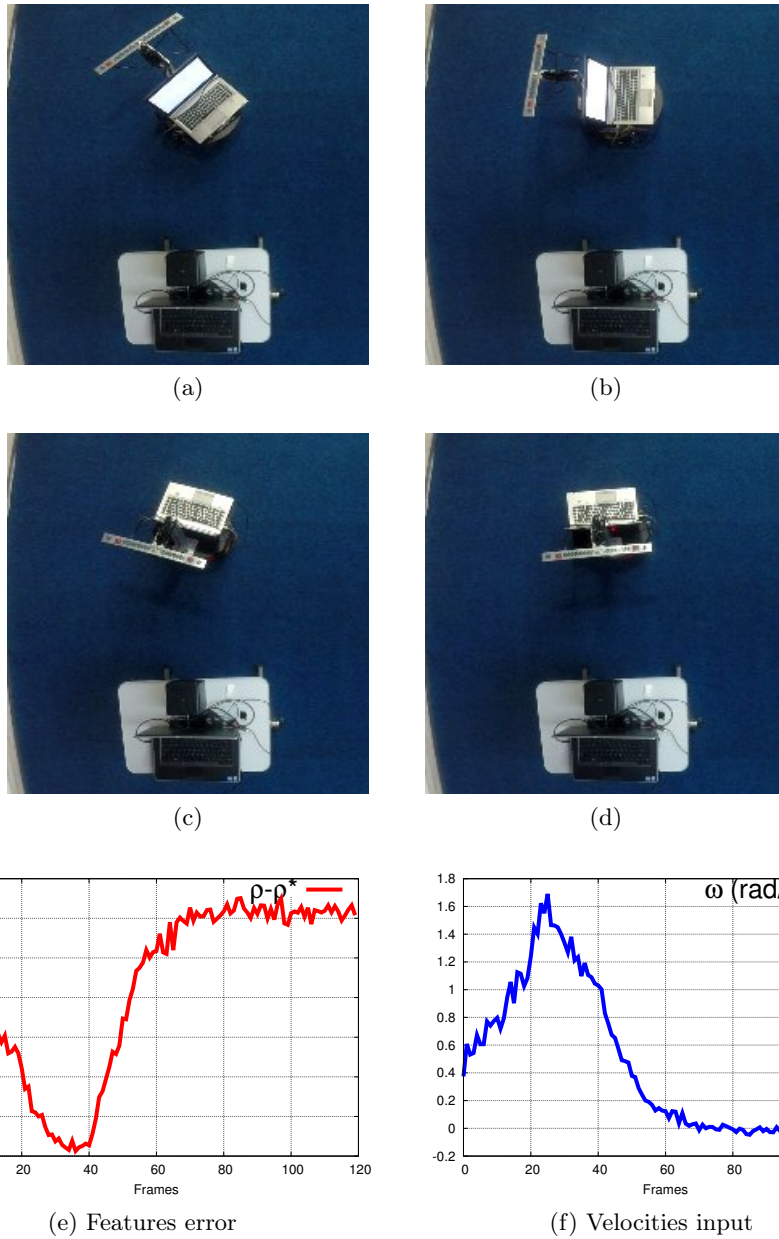
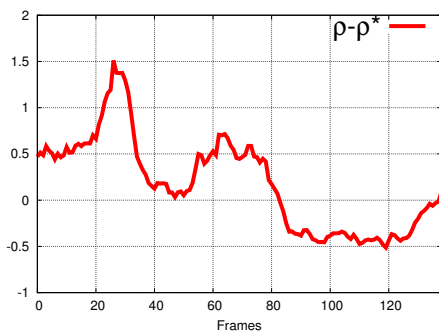
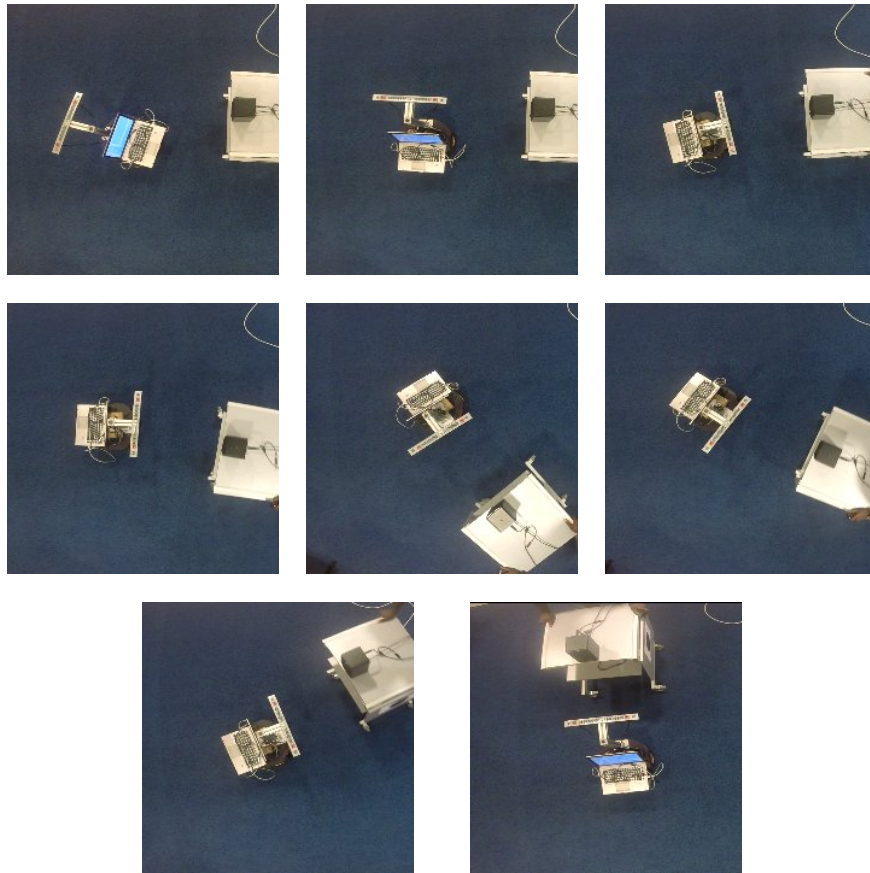


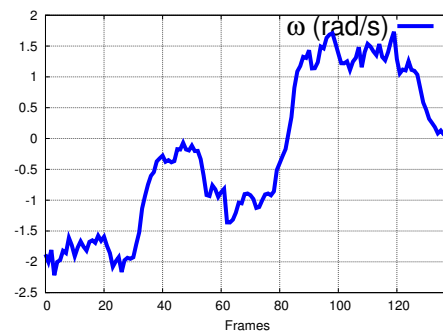
Figure 4.8: The *front-back ambiguity* is inherently solved by the control scheme: In a first step the absolute value of the error increases until the configuration corresponds to a model when  $y_s > 0$ . In a second phase the robot converges towards a correct pose with an exponential decrease of the magnitude of the error.

Several methods have been developed in control literature in order to estimate  $\widehat{\frac{\partial e}{\partial t}}$ . To name a few of them, a Kalman filter is proposed in [CG93], while a predictive control scheme is used in [GdM02] and an integrator in [CH08].





(i) Features error



(j) Velocities input

Figure 4.9: An experiment summing up all configurations (reading from left to right, top to bottom): first the front-back ambiguity is solved, in a second phase, the robot converges towards a satisfying pose, and finally the speaker is moved randomly while being accurately tracked by the robot.

### 4.2.6 Evaluation and limitations of the ILD-based task

In order to evaluate the performance of the ILD-based task, we tested the method through simulations that could guarantee repeatability and ground truth references to assess of the accuracy of the task. The simulation environment is based on *Roomsimove* [BOV12], a room acoustics simulator, that allows to parametrize the room size, reverberation, and the location of the source and microphones. From this tool, we designed a room of 8 m×6 m in which a sound source  $\mathbf{X}_s$  is located. Similarly to the first experiment presented in this section, the simulated tasks consisted in facing the sound source from several poses of the robot as depicted in Figure 4.10. The task was repeated for several initial poses that are varied by 20° around the sound source for a given distance  $\ell$  within the range  $[0^\circ, 180^\circ]$ . This decomposition yielded to 10 initial poses. It should be noted that the initial orientation is fixed, and only the initial positions are varied. We principally evaluated the orientation error of the final pose of the robot with respect to the desired task (facing the source). Such evaluation was repeated for several distance  $\ell \in \{0.5, 1, 2, 3\}$  and several reverberation times  $\text{RT}_{60} \in \{0, 50, 100, 200\}$  ms. The results are summarized in Figure 4.10, where the mean absolute error, in degree, is reported.

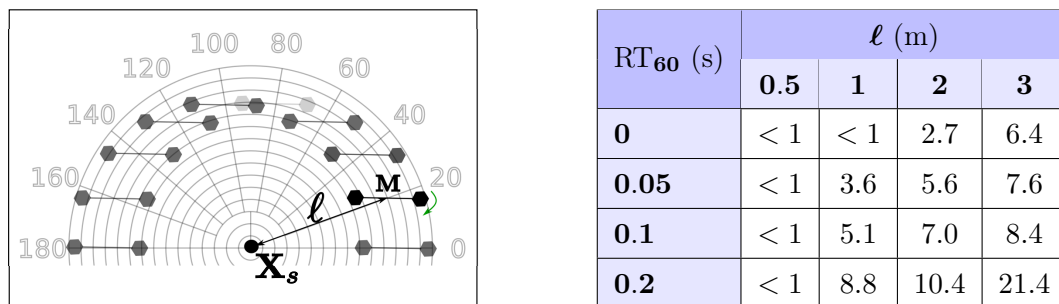


Figure 4.10: A simulated task that consists in orienting the robot in the direction of the sound source, from different poses in a 8 m × 6 m room. The final absolute mean error, in degree, is calculated for several reverberation times ( $\text{RT}_{60}$ ) and distances to the source.

The principal information that can be drawn from these simulations assesses that an ILD-based positioning task is accurate only in the close neighborhood of the sound source. This constraint can be explained by two distinct properties of ILDs that can be interpreted from the results. First the simulations confirm that increasing the distance between the microphones and the source degrades the task accuracy. This is confirmed by the results in anechoic conditions, from which a decrease of accuracy can be observed as soon as the microphones get further from the source. This is caused by the limitations inherent to ILD definition  $\rho = \frac{\ell_2}{\ell_1}$ . From the scene configuration described in Figure 4.1, the relationship between  $\ell_1$  and  $\ell_2$  is given by

$$\ell_2^2 = \ell_1^2 + 2dx_s \quad (4.53)$$

while  $\ell_1^2 = \ell^2 + \frac{d^2}{4} - dx_s$  and  $\ell_2^2 = \ell^2 + \frac{d^2}{4} + dx_s$ . When the robot is far from the sound source,  $\ell$  (respectively each  $\ell_i$ ) becomes large in comparison with  $d$ , the



inter-microphone distance. As a result, it comes out that  $\ell_1^2 \rightarrow \ell_2^2$  and  $\rho \rightarrow 1$ . Thus the energy difference between the two microphones becomes too small and the variation of the ILD  $\rho$  is not significant anymore: large robot motions induced a small variation of ILD which explains the results in anechoic conditions. This result also confirms that ILD measurements in free-field conditions are more relevant for a wide inter-microphones distance, which is unfortunately not compatible with the robotic context. By contrast, head-mounted systems are less affected by this limitation (see Chapter 6), because of the additional sound attenuation of the head (at high frequencies).

Furthermore, as already stated in [BG05] or [CCW06], the results exposed in Figure 4.10 confirm that ILD measurements are highly sensitive to reverberation. Systematic bias corrupts  $\rho$  when the microphones are far from the sound source, especially for positions close to walls. Indeed, with reverberation each microphone  $\mathbf{M}_i$  perceive actually an energy defined by

$$E_i = E_{i_{src}} + E_{i_{rev}} \quad (4.54)$$

where  $E_{i_{src}}$  is the source contribution while  $E_{i_{rev}}$  is produced by the virtual sound sources generated by echoes. These virtual sound sources interfere with the signal received from the actual sound source, with a constructive or destructive additional contribution. The measured ILD is modified as follow:

$$\rho = \frac{E_{1_{src}} + E_{1_{rev}}}{E_{2_{src}} + E_{2_{rev}}}. \quad (4.55)$$

This kind of erroneous measurement can have some influence on the accuracy of the task since the error  $e$  is then equal to

$$e = \frac{E_{1_{src}} + E_{1_{rev}}}{E_{2_{src}} + E_{2_{rev}}} - \rho^*. \quad (4.56)$$

Therefore if  $E_{i_{rev}}$  is significant enough with respect to  $E_{i_{src}}$ , the error  $e$  in the control loop is biased. One can deduce that a sufficient condition to obtain accurate ILD measurements (resp. accurate error  $e$ ) is a high Direct-to-Reverberant Ratio (DRR) that limits the impact of  $E_{i_{rev}}$  over  $E_{i_{src}}$ .

Hence under realistic conditions, that is to say in presence of reverberation, and if the robot is far from a given source, the performance of ILD-based positioning task may be drastically decreased. To overcome this limitation, we introduce the sound energy  $E_{\mathbf{M}}$  as an additional feature corresponding to a distance cue. The variation of this cue can be integrated in the control scheme in order to regulate the distance to the sound source by setting a desired energy level. In this way, a positioning task can be performed accurately from an initial configuration relatively far from the source.

### 4.3 Integrating the absolute level of energy as a distance cue

The distance of observation for accurate ILDs is in the near-field as demonstrated above. However the near-field assumption cannot be guaranteed in uncontrolled

environments. In this ultimate section of the chapter, we introduce a distance cue that can bring the robot to the near-field zone. Namely, this cue is the absolute level of sound energy. The sound level is naturally a measure of distance. As stated in Chapter 2.2, it is complex to deduce a metric value of distance from this cue. Indeed the sound level depends on the intrinsic level of the emitting sources and is also dependent of the room acoustic that may add destructive or constructive interferences. The distance cues are not often used in robot audition because of these limitations although some studies showed that coarse approximations could be obtained by combining binaural cues to signal amplitude information [Rod10] or by estimating the DRR [LC10]. In contrast to approaches based on localization, AS allows us to exploit distance cues such as the absolute level of energy. In this paradigm, we are not attached to extract a given metric measurement related to the sound level but rather to use the variation of the sound energy, related to the sound decay, so as to position a robot at a given distance to the sound source.

### 4.3.1 Energy estimation

For a first step, let us analyze the consistency of sound energy with respect to the reverberation issue, through AS framework. Reverberation as denoted in the previous section drastically decreases the performance of the ILD-based positioning task from far-field. It is then necessary to prove that the absolute energy level is less affected by reverberation and thus can overcome this limitation. For more versatility and flexibility, we consider measurements of the absolute energy in between the microphones, that is to say at the position  $\mathbf{M}$ . From (4.4), the sound energy  $E_{\mathbf{M}}$  received by the point  $\mathbf{M}$  is given by

$$E_{\mathbf{M}} = \frac{1}{\ell^2} \int_{t=0}^w a^2(t - \frac{\ell}{c}) dt. \quad (4.57)$$

In order to prove that reverberation has a minor effect on the sound energy, let us consider the reflective perturbations discussed above for the ILD. Assuming that the reference value  $E_{\mathbf{M}}^*$  is experimentally measured beforehand, both  $E_{\mathbf{M}}(t)$  and  $E_{\mathbf{M}}^*$  include the effect of reverberation  $E_{rev}$  caused by echoes. Therefore the energy level error in the control loop is equal to

$$e_{\mathbf{M}} = E_{\mathbf{M}_{src}} + E_{\mathbf{M}_{rev}} - (E_{\mathbf{M}_{src}}^* + E_{\mathbf{M}_{rev}}^*). \quad (4.58)$$

If  $E_{\mathbf{M}}^*$  is measured so that the sensors are in the near-field (*i.e.*, high DRR), as it is targeted in order to overcome ILD limitations, one can deduce that  $E_{\mathbf{M}_{src}}^* \gg E_{\mathbf{M}_{rev}}^*$ . In the far-field it is expected that the level of sound energy is lower than in the near-field because of the sound decay phenomenon detailed in Chapter 1.2. It can then be written that  $E_{\mathbf{M}} < E_{\mathbf{M}_{src}}^*$ . Thus while approaching the desired pose  $E_{\mathbf{M}}^*$ , we can simplify (4.58) by using (4.57) as:

$$e_{\mathbf{M}} \underset{s \rightarrow s^*}{=} \left( \frac{1}{\ell^2} - \frac{1}{\ell^{*2}} \right) \int_{t=0}^w a^2(t - \frac{\ell}{c}) dt. \quad (4.59)$$

Hence reverberation has a minor effect on the energy level cue, in the configuration in which the robot has to reach or stay in the near-field zone. From then on, since the energy level in  $\mathbf{M}$  is not directly available, we just approximate  $E_{\mathbf{M}}$  as the mean value of the energy received by each microphones:

$$E_{\mathbf{M}} \approx \frac{E_1 + E_2}{2}. \quad (4.60)$$

### 4.3.2 Energy level modeling

In a similar process as the one developed in Section 4.2 for modeling  $\mathbf{L}_{\rho}$ , the interaction matrix related to the absolute sound energy  $\mathbf{L}_{E_{\mathbf{M}}}$  can be inferred. Let us start from (4.2), specifying the distance to the source as  $\ell = \sqrt{x_s^2 + y_s^2}$ . The time variation of this distance is defined by:

$$\dot{\ell} = \frac{x_s \dot{x}_s + y_s \dot{y}_s}{\ell}. \quad (4.61)$$

By injecting the kinematic equation (4.16) in (4.61), the interaction matrix related to  $\ell$  is easily obtained as

$$\mathbf{L}_{\ell} = \begin{bmatrix} -\frac{x_s}{\ell} & -\frac{y_s}{\ell} & 0 \end{bmatrix}. \quad (4.62)$$

With the assumption that the source emits a constant signal during the time frame  $w$ , from (4.57) the time variation of  $E_{\mathbf{M}}$  is

$$\dot{E}_{\mathbf{M}} = -2E_{\mathbf{M}} \frac{\dot{\ell}}{\ell}. \quad (4.63)$$

Therefore the interaction matrix  $\mathbf{L}_{E_{\mathbf{M}}}$  related to the sound energy perceived in  $\mathbf{M}$  is given by

$$\mathbf{L}_{E_{\mathbf{M}}} = -2 \frac{E_{\mathbf{M}}}{\ell} \mathbf{L}_{\ell} = E_{\mathbf{M}} \begin{bmatrix} \frac{2x_s}{\ell^2} & \frac{2y_s}{\ell^2} & 0 \end{bmatrix}. \quad (4.64)$$

Once again, an approximation of the latter interaction matrix is necessary since the sound source location is unknown:

$$\widehat{\mathbf{L}}_{E_{\mathbf{M}}} = E_{\mathbf{M}} \begin{bmatrix} \frac{2\widehat{x}_s}{\widehat{\ell}^2} & \frac{2\widehat{y}_s}{\widehat{\ell}^2} & 0 \end{bmatrix}. \quad (4.65)$$

It can be noticed that the sufficient stability conditions of a task based on  $\widehat{\mathbf{L}}_{E_{\mathbf{M}}}$  are the same as the ILD described in Section 4.2.4. Indeed

$$\mathbf{L}_{E_{\mathbf{M}}} \widehat{\mathbf{L}}_{E_{\mathbf{M}}}^+ = \frac{\widehat{\ell}^2}{\ell^2} \frac{x_s \widehat{x}_s + y_s \widehat{y}_s}{\widehat{x}_s^2 + \widehat{y}_s^2}, \quad (4.66)$$

is positive as soon as  $\text{sign}(\widehat{x}_s) = \text{sign}(x_s)$  and  $\text{sign}(\widehat{y}_s) = \text{sign}(y_s)$ . Hence  $\widehat{x}_s$ ,  $\widehat{y}_s$  and  $\widehat{\ell}$  can be estimated with the same parameters as these used in  $\widehat{\mathbf{L}}_{\rho}$ .

### 4.3.3 Control scheme and task analysis

For the rank-one interaction matrix  $\mathbf{L}_{E_M}$  of the energy level, the subspace  $\mathbf{S}^* \in \mathbb{R}^{3 \times 2}$  implying a virtual link of class 2 is:

$$\mathbf{S}^* = \begin{bmatrix} 0 & -y_s \\ 0 & x_s \\ 1 & 0 \end{bmatrix}. \quad (4.67)$$

The first motion implied by  $\mathbf{S}^*$  is a pure rotation. Indeed the distance  $\ell$  between  $\mathbf{M}$  and the source is invariant with respect to the orientation of the microphones. The second column illustrates all translations for which the distance  $\ell$  to the sound source is unchanged. Namely, it refers to a circular motion around the sound source. Consequently by combining these two types of motion,  $\mathbf{S}^*$  describes a circle of radius  $\ell$  around the sound source with unconstrained orientation of the microphones, what is illustrated by Figure 4.11. This result emphasizes that the ILD and the energy level are complementary features. Indeed, in contrast with energy level, ILD measurements control microphones orientation while the distance is constrained by the energy level.

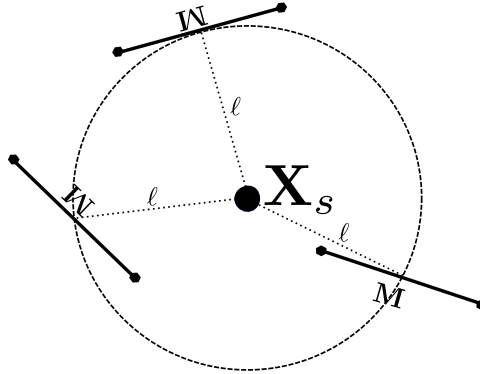


Figure 4.11: Admissible poses of the microphones considering the sound energy level as acoustic features

Then one can define an interaction matrix  $\widehat{\mathbf{L}}_{\rho E}$  combining the ILD  $\rho$  to the energy level  $E_M$  by simply stacking  $\widehat{\mathbf{L}}_{\rho}$  and  $\widehat{\mathbf{L}}_{E_M}$  as follows:

$$\widehat{\mathbf{L}}_{\rho E} = \begin{bmatrix} \frac{2\widehat{x}_s(\rho-1)-d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{2\widehat{y}_s(\rho-1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{\widehat{y}_s d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} \\ \frac{2E_M \widehat{x}_s}{\widehat{\ell}^2} & \frac{2E_M \widehat{y}_s}{\widehat{\ell}^2} & 0 \end{bmatrix}. \quad (4.68)$$

From the latter interaction matrix, the control scheme related to the velocity of the microphones is defined as

$$\mathbf{u}_M = -\lambda \widehat{\mathbf{L}}_{\rho E}^+ \mathbf{e} \quad (4.69)$$

where the error vector is defined as  $\mathbf{e} = [\rho - \rho^*, E_M - E_M^*]^\top$ . For this control scheme integrating the energy level, the vector subspace  $\mathbf{S}^* \in \mathbb{R}^{3 \times 1}$  implying a virtual link

of class 1, is defined as:

$$\mathbf{S}^* = \begin{bmatrix} y_s \\ -x_s \\ 1 \end{bmatrix}. \quad (4.70)$$

In this case, the motion described by this subspace refers to a circular motion around the sound source at a distance  $\ell$  maintaining the same orientation of the microphones with respect to the source (see Figure 4.12). Consequently with a good selection of

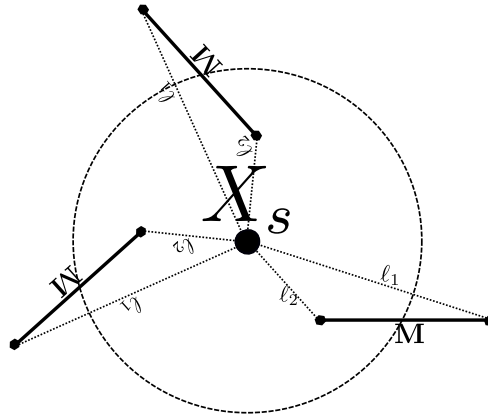


Figure 4.12: Admissible poses of the microphones considering the sound energy level and the ILD as acoustic features

the desired features values, it is possible to reach a pose around the sound source by only considering the measurement of the energy and the ILD. Such a task can be referred to a coarse-to-fine approach. When the robot is far from the sound source, the ILD measurement gives a rough approximation of the actual direction of the source. However as the robot is moving closer to the sound source, the ILD measurement is refined, and the robot motion becomes more accurate. Hence, from rough and approximated feature measurements, a virtuous cycle is created from the closed-loop control, that allows to achieve difficult tasks (see Section 4.3.4). Eventually, it should be emphasized that until now, no additional signal processing, nor filtering has been employed for the control scheme. Additionally, the feature measurements and the control scheme are independent from any additional tracking method, as long as the sound source of interest is predominant in the environment. The computation cost of this AS framework is drastically decreased, when compared to classic localization methods in adverse conditions. Actually, the most complex calculation of the framework consists in computing the pseudo-inverse of  $\widehat{\mathbf{L}}_{E_M} \in \mathbb{R}^{2 \times 3}$ .

### 4.3.4 Experimental validations

#### 4.3.4.1 Preliminaries: control scheme and experimental setup

Once again, we consider the 2-DOF robot modeled in Section 4.2.5.1. The control input to this robot is given by

$$\dot{\mathbf{q}} = -\lambda \widehat{\mathbf{J}}_{\rho \mathbf{E}}^{-1} \mathbf{e} \quad (4.71)$$

with

$$\widehat{\mathbf{J}}_{\rho \mathbf{E}} = \begin{bmatrix} \frac{2\widehat{y}_s(\rho-1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{D_x(2\widehat{x}_s(\rho-1) - d(\rho+1)) + \widehat{y}_s d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} \\ \frac{2E_{\mathbf{M}}\widehat{y}_s}{\widehat{\ell}^2} & \frac{2D_x E_{\mathbf{M}}\widehat{x}_s}{\widehat{\ell}^2} \end{bmatrix}. \quad (4.72)$$

Apart from the degenerate cases ( $\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s = 0$  and  $\widehat{\ell}^2 = 0$ ), singularities of the interaction matrix could appear as soon as  $\widehat{y}_s = D_x$  or  $\widehat{y}_s = 0$ . In these particular situations, the determinant of this Jacobian matrix given by

$$|\widehat{\mathbf{J}}_{\rho \mathbf{E}}| = \frac{2E_{\mathbf{M}}\widehat{y}_s d(D_x - \widehat{y}_s)(1 + \rho)}{\widehat{\ell}^2(\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s)}. \quad (4.73)$$

would be equal to 0, and the interaction matrix would not be invertible. Hopefully, in practice the sound source position is approximated so that  $\widehat{y}_s > D_x$ .

The experiments were conducted on the *Pioneer 3DX* with the same configuration given in Section 4.2.5.1 and illustrated in Figure 4.5. The global processing time of one iteration was also negligible ( $< 10$  ms). The acoustic conditions are detailed in the sections devoted to each experiment. The parameters given in Table 4.2 were used for all experiments. An adaptive gain  $\lambda(x)$  in which  $x$  refers to the norm of the error  $\mathbf{e}$  is used to smooth and speed up the robot motion.

$d$	0.31 m
$D_x$	0.3 m
$\widehat{y}_s$	1 m
$\widehat{x}_s$	$\text{sign}(\rho - 1) \times 1$ m
$\lambda$	$0.45e^{(-1.5x)}$

Table 4.2: Experimental settings

#### 4.3.4.2 Typical positioning tasks

The first tests were conducted in a room characterized by a reverberation time  $\text{RT}_{60} \approx 580$  ms. The sound source is a loudspeaker emitting a white Gaussian noise. The desired energy level is learned by placing the robot at 80 cm in front of the loudspeaker, while the desired ILD is set so that  $\rho^* = 1$ . The SNR at the desired pose is around 20 dB. The loudspeaker being directional, the admissible poses of the microphones were not in a circular configuration as stated in Figure 3.2, but they were shaped by the directivity properties of the speaker.

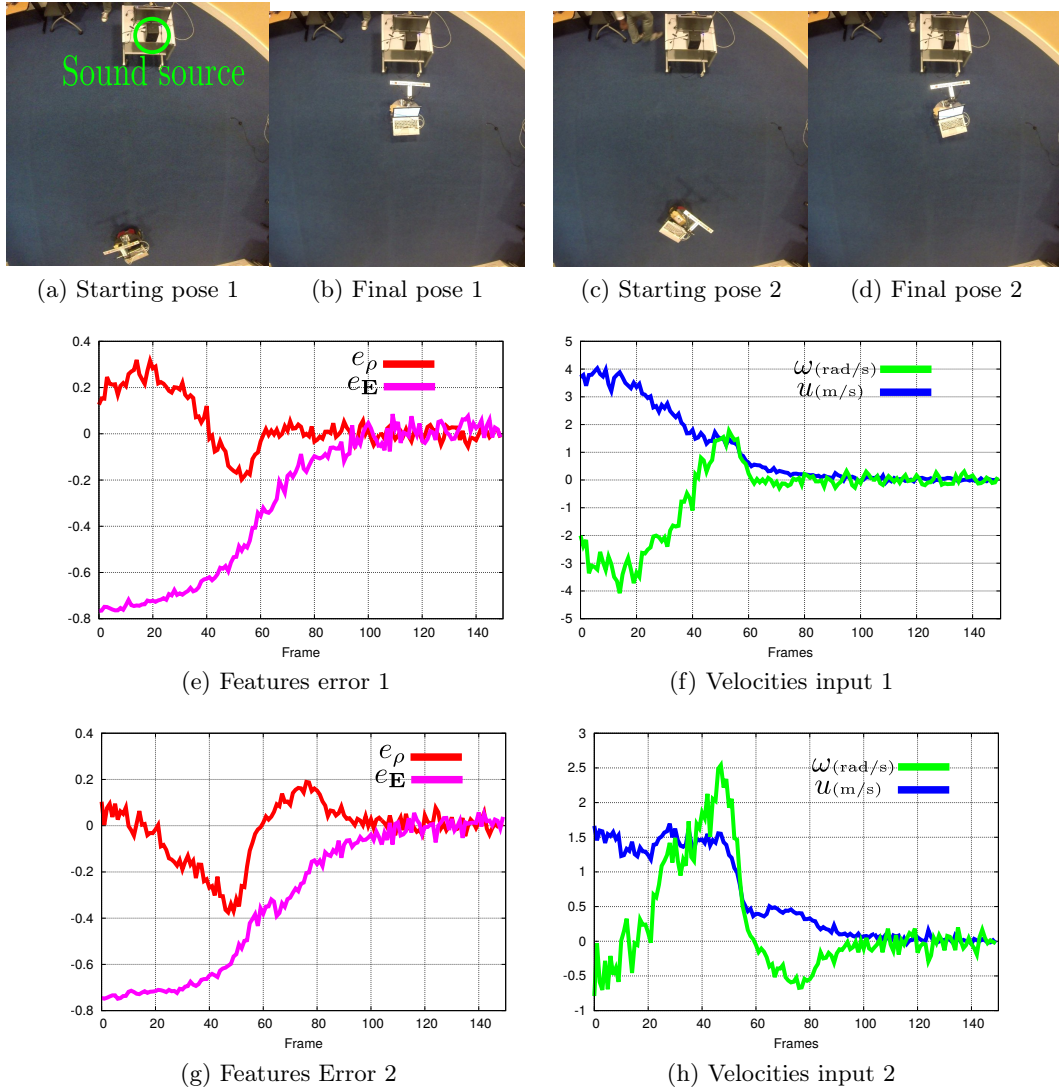


Figure 4.13: Typical positioning tasks from two different starting poses

First we showed the validity of our approach with a static sound source, by starting the robot from different poses. As illustrated in Figure 4.13, the control scheme allows to reach a pose satisfying the desired features from various initial poses. Actually, as long as the difference of energy between the microphones is perceptible at the initial pose, the control scheme is able to position the robot in a desired configuration. A fine tuning of  $\lambda$  or  $w$  can also improve the sensitivity of the control scheme to small differences in the energy level. By increasing the size of  $w$ , the difference of energy recorded by the pair of microphones is more significant. And  $\lambda$  tunes the time to convergence.

In the second part of the experiment, we considered a moving sound source. As shown in Figure 4.14, the robot is able to follow a sound source, with exactly the

same control scheme as before. As expected, when the robot is far from the sound source, it can be observed on Figure 4.14e that the ILD is not always accurate. Indeed we can notice some abrupt changes in the ILD error curve. But as soon as the robot gets closer to the sound source, the ILD value is correct, and the task can be correctly completed.

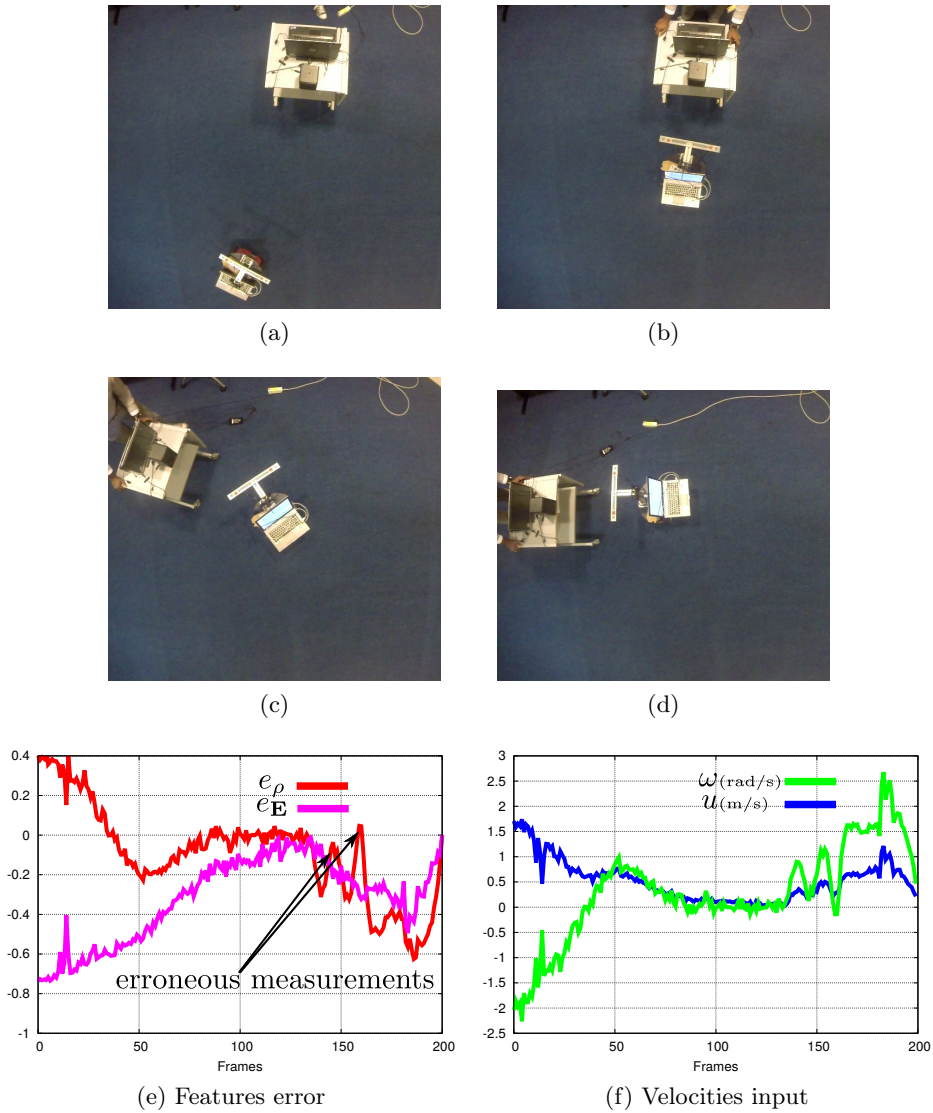


Figure 4.14: Following a moving sound source

#### 4.3.4.3 Robustness and flexibility in long range navigation

In a second experiment, we conducted a long range navigation test. Starting from the previous room, we moved the source through different environments of the laboratory with the robot pursuing the sound source in real time. Thus several acoustic



conditions were encountered during the navigation task as described in Figure 4.15.

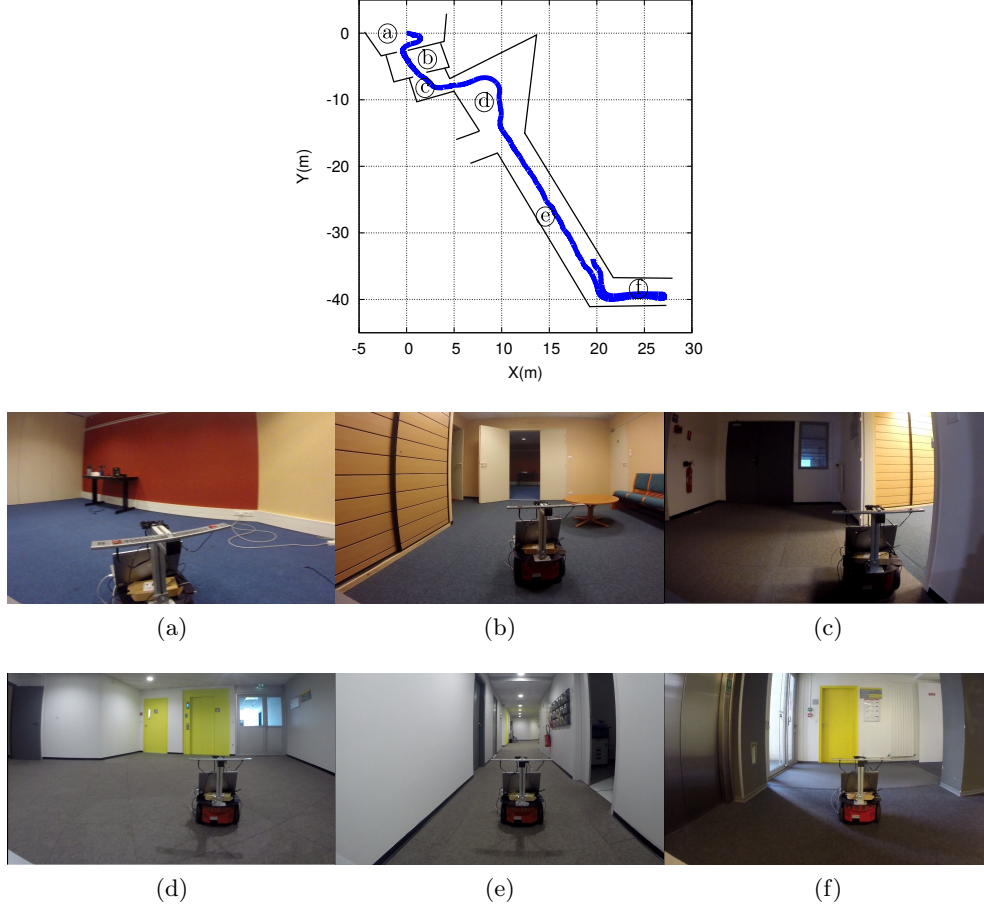


Figure 4.15: Odometry data from the navigation task in indoor environment. The acoustic conditions for each location are respectively:  
 (a)  $RT_{60} \approx 580\text{ms}$   $SNR \approx 20\text{dB}$ , (b)  $RT_{60} \approx 620\text{ms}$   $SNR \approx 20\text{dB}$ , (c)  $RT_{60} \approx 680\text{ms}$   $SNR \approx 16\text{dB}$ , (d)  $RT_{60} \approx 880\text{ms}$   $SNR \approx 13\text{dB}$ , (e)  $RT_{60} \approx 700\text{ms}$   $SNR \approx 18\text{dB}$ , (f)  $RT_{60} \approx 620\text{ms}$   $SNR \approx 17\text{dB}$ .

The SNR varied from 20 dB to 13 dB while the reverberation rate changed from 580 ms to 880 ms depending on the location. The ambient noise was mainly caused by the ventilation systems and a server room in (d). Nonetheless the robot never lost the track of the sound source despite dynamic and challenging conditions. Indeed the rooms and corridors crossed by the robot have different shapes, and are built with different materials and clutter. For instance the corridor (e) is narrow and produces strong first order echoes. Besides the echoes are not necessarily symmetric, because of the office doors that were open. In (f) different surrounding materials (glass door, metallic door of the elevator, walls...) were producing several types of echoes that could disturb the control scheme.

Consequently, from this experiment it can be emphasized that the proposed control scheme is robust and flexible enough to deal with indoor real world environment.

Indeed, since our method does not rely on any tracking or filtering of the signal, there is no tuning nor parameter dependent on the acoustic environment. Thus our approach is robust to environment changes.

#### 4.3.4.4 Cooperative application

The framework has also been tested in a cooperative task involving two robots. This time, instead of using a loudspeaker to generate the sound, we used the propellers of an unmanned aerial vehicle (UAV) as the sound source. Indeed most UAVs are known to be noisy. However, in a conventional sound localization scheme, this sound is considered more as a disturbance than a feature to exploit [BSFL14][FON+13], while our approach takes advantage of this noise. In this experiment an UAV (mikrokopter MK-Quadro), remotely controlled, led the unicycle ground robot just by the sound naturally emitted from the propellers. Nonetheless it should be noted that the sound emitted by the UAV was not stationary nor omnidirectional. Indeed the UAV was oscillating during the flight and the sound was produced by the four propellers of the UAV. Nevertheless, despite these unfavourable conditions, the ground robot was able to follow the UAV (see Figure 4.16), even if the motion of the robot was less smooth than in the previous experiments. This basic experimental scenario confirmed that our approach is relevant and suitable for cooperative tasks involving several robots. Furthermore, the control scheme based on only two microphones and the low computation cost of the framework make us believe that this kind of control scheme can be embedded on different types of robots. More evolved and complex tasks could then be achieved involving cooperation between aerial and ground robots or formation control of swarm robots.

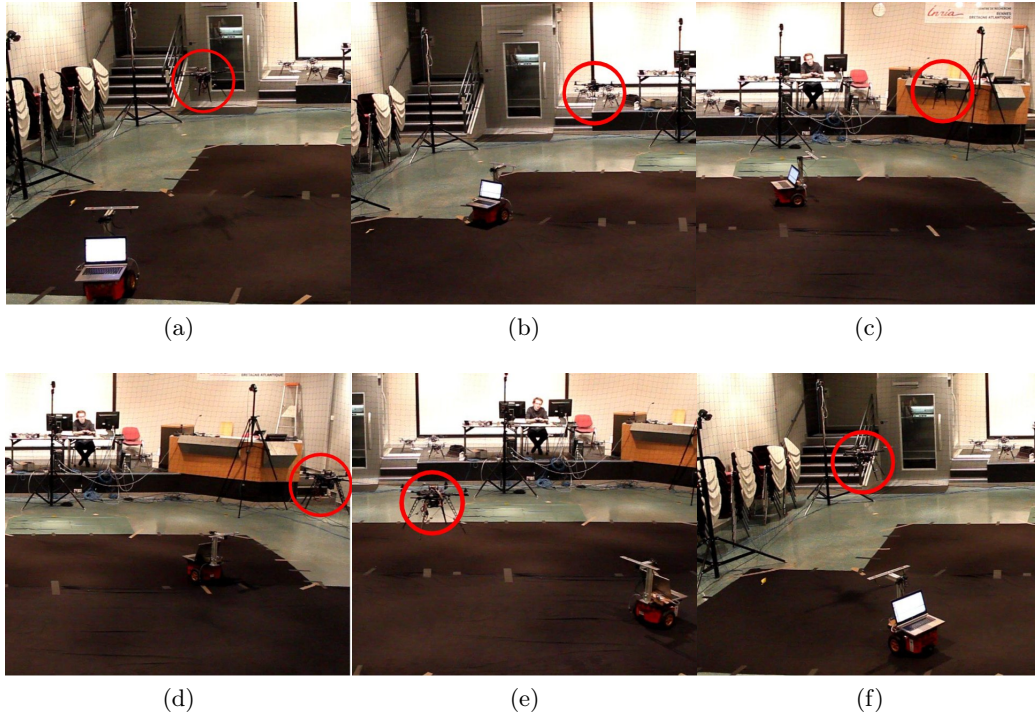


Figure 4.16: An example of cooperative application where an aerial robot (circled in red) is leading a ground robot with the sound emitted by its propellers

### 4.3.5 Numerical evaluation

Eventually, the control scheme was tested through simulations. The simulated environments and conditions are strictly identical to the evaluation performed in Section 4.2.6. This time, the task consists in approaching, at a given distance, the sound source in order to face it. Hence  $\rho^*$  is set to 1, while  $E_M^*$  is measured experimentally in anechoic conditions as the energy measurement when  $\rho = 1$  and  $\ell = 0.5$  m. It should be emphasized that  $E_M^*$  has been measured only once, and the same value of  $E_M^*$  is used for all acoustic conditions. The results are summarized in Table 4.3, where the mean absolute error of orientation, in degree, and the mean absolute error of the range positioning, in cm, are reported. These results show impressive improvements compared to the task of only facing the sound source. For all starting poses, even in the far-field and with reverberation, the robot is always able to face to the sound source accurately as the errors reported are all lesser than  $1^\circ$ . It can be also noted that as expected and demonstrated in Section 4.3.1, the energy level as a range cue is little affected by reverberation, with negligible errors varying from 1 to 4 cm under moderate reverberation. Furthermore, interestingly, the energy level is consistent over changing conditions. Despite an initial measurement  $E_M^*$  performed under anechoic conditions, the range error remains lower than 5 cm for a reverberation level  $RT_{60} = 0.2$  s. Of course, more accurate results could be obtained if  $E_M^*$  was measured in the same acoustic conditions as the task environment. These excellent

RT <sub>60</sub> (s)	$\ell$ (m)							
	0.5		1		2		3	
<b>0</b>	< 1	(1.47)	< 1	(1.47)	< 1	(1.48)	< 1	(1.47)
<b>0.05</b>	< 1	(1.77)	< 1	(1.77)	< 1	(1.78)	< 1	(1.8)
<b>0.1</b>	< 1	(1.94)	< 1	(2.05)	< 1	(1.95)	< 1	(1.96)
<b>0.2</b>	< 1	(3.90)	< 1	(3.90)	< 1	(3.96)	< 1	(3.98)

Table 4.3: A simulated task that consists in approaching and facing of the sound source ( $\rho^* = 1$ ,  $E_M^* \equiv \ell^* = 0.5$  m), from different poses in a  $8 \times 6$  m<sup>2</sup> room. The final absolute mean error, in degree, and the final range error (bracketed values), in cm, are calculated for several reverberation times (RT<sub>60</sub>) and distances to the source.

properties substantially explain the good results obtained in experiments, in which our approach is robust and flexible to changing and real world environments.

## 4.4 Conclusion

This chapter globally emphasizes several benefits of the AS paradigm. First, the two solutions presented are based on features (ILD, energy as distance cue), that are challenging to exploit under binaural localization paradigm. For instance most work exploiting ILDs conjointly estimates ITDs, in order to map them with an azimuth direction from a learning process [MvdPK11]. However, these cues can be fully exploited with a control scheme based on their variation.

To do so, we evaluated geometrically the information gathered by ILD measurements. From ILD measurements, the sound source location is expressed through a circle depending on the ratio of energy. This circle is subject to the *front-back ambiguity*, since the ILD circle is centered somewhere on the extension of the microphones interaural axis. As a second part, we focused on the controller of the sensor-based scheme, more exactly on the relationship between the ILD variation with respect to the robot motion. As a result, an approximated interaction matrix could be extracted from an ILD cue. From the design of a sensor-based control scheme, the task analysis confirmed that ILD information could be exploited to orient the robot with respect to a sound source direction. Eventually the stability analysis showed that the source location is not required for this positioning task. Actually, there exists an infinite number of solutions that guarantee the convergence of the controller to complete such tasks.

The second interesting benefit of AS comes from its ability to deal with real world conditions thanks to the flexibility and the robustness of the approach. The framework described above has been validated experimentally in real conditions despite an initial model based on ideal conditions. In our experiments conducted on a mobile robot, we showed that from our approach a robot is able to face accurately a

sound source, to track a moving sound source, and more impressively to cope with the *front-back ambiguity*, without any information about the source location.

In the second part of this chapter, we integrated a distance cue to our model to overcome the limitation inherent to ILDs: accuracy decreases for a distant source. This distance cue is the energy level measured by the microphones. ILDs allow orienting the robot towards the sound source while the signal energy controls the distance between the robot and the sound source. By combining these two features the robot is able to follow a moving sound source, despite erroneous ILD measurements, as confirmed in several experiments.

Furthermore, this framework is directly performed on raw measurements without tracking, filtering or signal enhancement. As a result, the computational cost of the control scheme is small, less than 10 ms in our experiments. The different experiments described in this chapter emphasized the robustness of the control scheme towards reverberation and variability of the environment. The different use cases showed the applicability of the method in various topic that could take advantage of the auditory sense:

- A mobile robot navigated by pursuing a sound source through indoor environment while facing different level of reverberation and noise. This kind of tasks has potential applicability to a wide a range of field such as security patrolling or delivery services.
- An aerial robot has been able to guide a ground robot only with the sound of its propellers. In this case, applications in the field of multi-robot and cooperative tasks, or search-and-rescue mission can be targeted.

Globally the ILD-based control framework presented in this chapter could help fulfilling the requirements for robots endowed with hearing sense that are: embeddability, real time processing, flexibility to broad environments, robustness to noise and reverberation. But on the other hand, one constraint of this framework is the use of a continuous and slowly varying sound signal. Hence, this control framework is not particularly adapted to non-stationary, intermittent sources such as speech. Furthermore, until now we assumed a single source in the acoustic scene. For acoustic scenes considering speech or multiple sources, ITDs are more relevant cues. These cues are the subject of the next chapter that introduces an ITD-based control framework.

## Chapter 5

# ITD-based aural servo

In the continuity of the previous chapter centered on ILDs, we introduce in this chapter the modeling of a sensor-based control framework using ITD measurements. ITDs are certainly the most studied auditory cues for localization in machine hearing. With a wide variety of techniques [PDA12, NMOK03] and microphones topology [VMRL03, KND06], the estimation of this cue concentrated a lot of effort from the robot audition community. However the results remain contrasted when facing dynamic realistic environments. Generally each method for estimating ITDs is tuned for a particular environment and context. By contrast, this chapter demonstrates that in the AS paradigm, ITD cues can still be exploited despite changing environment or rough approximations in acoustic scene modeling. The control framework is based on a relatively common ITD estimation algorithm: GCC-PHAT.

In a binaural configuration characterized by a pair of microphones in a free-field area, a geometrical analysis is performed in Section 5.1 in order to establish the link between an ITD measurement and a sound source location.

In Section 5.2, from this analysis, two interaction matrices are extracted from the variation of ITDs with respect to the microphones motions. These matrices are respectively extracted from a rigorous geometrical model of ITDs and an approximation obtained from the far-field assumption commonly adopted in sound source localization. The control scheme and the positioning task are then analyzed from these interaction matrices. Thereafter, the stability properties, in the Lyapunov sense, are studied to demonstrate the feasibility of such an approach. Eventually experimental results and numerical evaluations support the robustness and flexibility properties of the proposed AS framework.

In the last segment of this chapter, we focus in Section 5.3 on the context of multi-sources configurations. Control schemes considering two, three or more sound sources are given. Experimental results and numerical evaluations are provided as well.

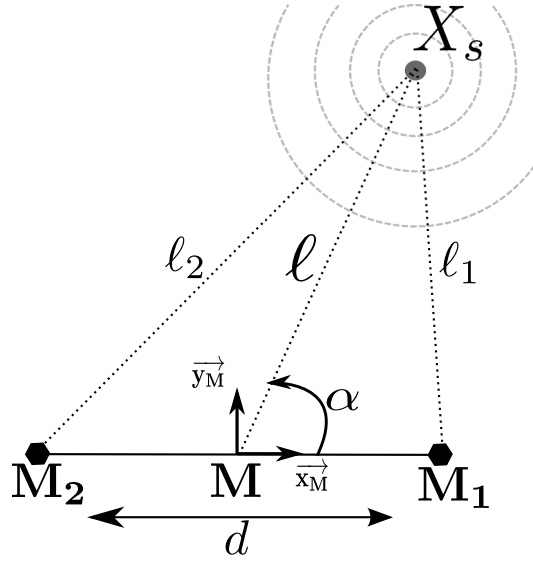


Figure 5.1: Geometric configuration of the considered system, that includes a source  $\mathbf{X}_s$  emitting a spherical sound wave, and a pair of microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$ .

## 5.1 ITD modeling

### 5.1.1 Scene configuration

In a first phase, let us start from the configuration used for ILD-based control. We then recall the scene configuration that assumes a motorized pair of microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$  in an area free of obstacle and separated by a distance  $d$  illustrated in Figure 5.1. An omni-directional sound source  $\mathbf{X}_s$  is continuously emitting and we assume that linear acoustics hypotheses hold. In the following,  $\mathbf{X}_s$  is shaped as a point that generates a sound wave uniformly in all directions. A frame  $\mathcal{F}_m(\mathbf{M}, \vec{x}_M, \vec{y}_M)$  originating from the midpoint  $\mathbf{M}$  of the pair of microphones. Considering a planar scene (*i.e.*,  $z_M = 0$ ), the Cartesian coordinates of each microphone are respectively  $\mathbf{M}_1(\frac{d}{2}, 0)$  and  $\mathbf{M}_2(-\frac{d}{2}, 0)$ . The sound source  $\mathbf{X}_s(x_s, y_s)$  is located at a distance  $\ell_i$  from each microphone  $\mathbf{M}_i$ . These distances  $\ell_i$  are respectively given by

$$\begin{cases} \ell_1 = \sqrt{(x_s - d/2)^2 + y_s^2} \\ \ell_2 = \sqrt{(x_s + d/2)^2 + y_s^2}. \end{cases} \quad (5.1)$$

Moreover, let  $\ell$  be the distance between  $\mathbf{M}$  and  $\mathbf{X}_s$ , while  $\alpha$  is the incident angle of the sound source with respect to the microphones axis.  $\alpha$  is also known as the direction of arrival (DOA) of the source. The sound source position is thus characterized by the following relationships:

$$\begin{cases} x_s = \ell \cos \alpha \\ y_s = \ell \sin \alpha \end{cases} \quad \text{or} \quad \begin{cases} \alpha = \text{atan2}(y_s, x_s) \\ \ell = \sqrt{x_s^2 + y_s^2}. \end{cases} \quad (5.2)$$



This configuration is then endowed with 3 DOF in the horizontal plane, that are the translations along  $\vec{x}_M$  and  $\vec{y}_M$  axis, and the rotation around  $\vec{z}_M$ .

### 5.1.2 Geometrical properties of the ITD

The first part of this chapter is dedicated to the study of an ITD  $\tau$ , that corresponds to the time difference in sound arrival between two microphones. Let us focus first on the properties of this cue in the geometric configuration described above. From the Figure 5.1, the sound wave emitted by  $\mathbf{X}_s$  reaches each microphone  $\mathbf{M}_i$  at a time  $t_i$  given by

$$t_i = \frac{\ell_i}{c}, \quad (5.3)$$

in which  $c$  is the sound celerity. As a result the ITD  $\tau$  between the pair of microphones is given by

$$\tau = t_2 - t_1 = \frac{\ell_2}{c} - \frac{\ell_1}{c} \quad (5.4)$$

From the measurement of  $\tau$  and the knowledge of the latter relationship, one can deduce the sound location. For this purpose, by injecting (5.1) in (5.4) we obtain:

$$\tau = \frac{\sqrt{(x_s + \frac{d}{2})^2 + y_s^2} - \sqrt{(x_s - \frac{d}{2})^2 + y_s^2}}{c}. \quad (5.5)$$

The latter equation develops as

$$\frac{\sqrt{(x_s + \frac{d}{2})^2 + y_s^2}}{c\tau} = \frac{c\tau + \sqrt{(x_s - \frac{d}{2})^2 + y_s^2}}{c\tau}, \quad (5.6)$$

that is equivalent to

$$\frac{(x_s + \frac{d}{2})^2 - (x_s - \frac{d}{2})^2}{(c\tau)^2} - 2\sqrt{\frac{(x_s - \frac{d}{2})^2 + y_s^2}{(c\tau)^2}} = 1. \quad (5.7)$$

After simplification we obtain

$$4 \left( \frac{x_s^2 (d^2 - (c\tau)^2)}{(c\tau)^2} - y_s^2 \right) = d^2 - (c\tau)^2. \quad (5.8)$$

Finally the equation is reduced to

$$\frac{x_s^2}{\frac{(c\tau)^2}{4}} - \frac{y_s^2}{\frac{d^2 - (c\tau)^2}{4}} = 1. \quad (5.9)$$

This equation corresponds to an hyperbola  $\mathcal{H}$ . Hence there are infinite solutions to characterize a potential location of the source as depicted in Figure 5.2. It should also be mentioned that the same reasoning can be developed for 3D scenes (see [APH14]), in which  $\mathcal{H}$  becomes an hyperboloid.



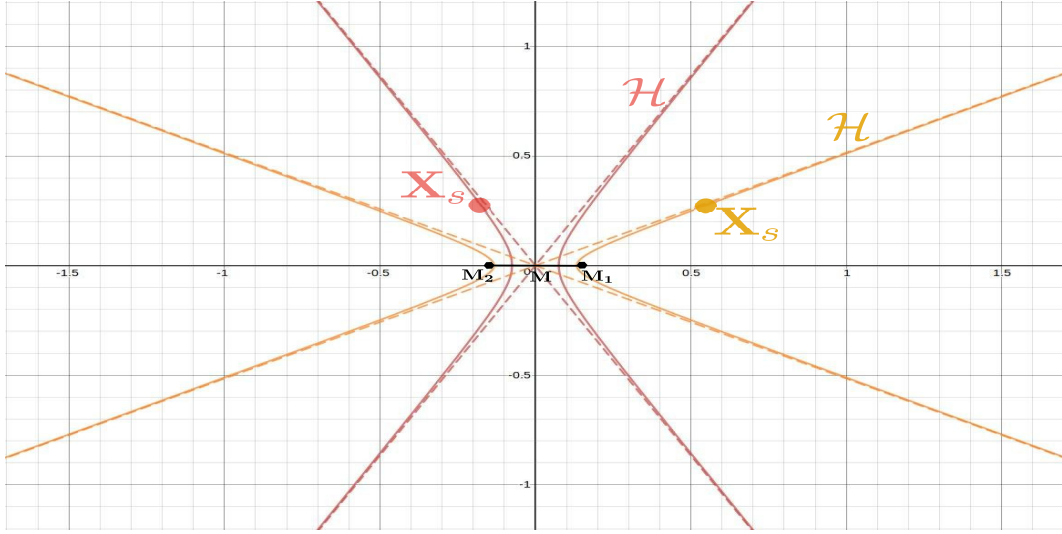


Figure 5.2: For a given sound source  $\mathbf{X}_s$ , the ITD geometrically refers to an hyperbola  $\mathcal{H}$ , for which front/back and left/right ambiguity should be solved. The asymptotes values (dotted line) of the hyperbola can serve as an approximation (far-field assumption) of the sound direction.

Several properties result from (5.9) in the planar case. First this hyperbola exists only if

$$-\frac{d}{c} < \tau < \frac{d}{c} \text{ and } \tau \neq 0. \quad (5.10)$$

As a result when  $\tau = 0$ , the hyperbola  $\mathcal{H}$  degenerates to a straight line that corresponds to the bisector of the microphone pair (*i.e.*  $x_s = 0$ ). This geometrical representation states that the sound source is "somewhere" on the hyperbola  $\mathcal{H}$ , or on the bisector of the microphones when  $\tau = 0$ . The position of the sound source can be refined by resolving the ambiguities related to  $\mathcal{H}$ . First, there is an ambiguity about determining on which branch of the hyperbola the sound source is located, for a given ITD. Fortunately, this ambiguity can be solved from the sign of the ITD  $\tau$ . Indeed from (5.1), (5.7) can be rewritten

$$\frac{2x_s d}{(c\tau)^2} - \frac{2\ell_1}{c\tau} = 1. \quad (5.11)$$

With some rearrangements, (5.11) becomes

$$x_s = \frac{c\tau(c\tau + 2\ell_1)}{2d}. \quad (5.12)$$

Knowing from (5.4) that  $c\tau = \ell_2 - \ell_1$ , the sound source lateral location with respect to the microphones frame is determined by

$$x_s = \frac{c\tau(\ell_1 + \ell_2)}{2d}. \quad (5.13)$$

In plain, if  $\tau < 0$  (*i.e.* the source is closer to  $\mathbf{M}_2$  than  $\mathbf{M}_1$ ), then the source is located on the left branch, else the source is on the right branch. As for the front or back position of the sound source with respect with the microphones axis, it can be related to the sign of the hyperbola  $\mathcal{H}$  asymptotes given by

$$y = \pm \frac{\sqrt{d^2 - (c\tau)^2}}{c\tau} x \quad (5.14)$$

However, the asymptotes' sign is independent from the sign of the ITD and hence cannot be determined. This indeterminacy draws the already discussed concept of *front-back ambiguity*. By extension, in 3D scenes, this ambiguity related to an hyperboloid leads to the so-called cone of confusion inherent to mammals sound localization as discussed in Chapter 2.1.2.

Eventually, one can also emphasize that for distant sound sources, that is, for a source located at a pose where the hyperbola  $\mathcal{H}$  has reached its asymptotic value (see Figure 5.2), (5.14) can be written as

$$\tan \alpha = \pm \frac{\sqrt{d^2 - (c\tau)^2}}{c\tau}, \quad (5.15)$$

in which  $\alpha$  is the DOA (see Figure 5.1). From trigonometry properties, one can develop (5.15) as

$$\frac{1 - \cos^2 \alpha}{\cos^2 \alpha} = \frac{d^2 - (c\tau)^2}{(c\tau)^2}, \quad (5.16)$$

that can be simplified as

$$(c\tau)^2 = d^2 \cos^2 \alpha. \quad (5.17)$$

Finally one obtains

$$\tau = A \cos \alpha \quad (5.18)$$

where  $A = d/c$ . Equation (5.18) corresponds to the definition of ITDs with respect to DOAs under the far-field assumption. Thus it is possible to geometrically differentiate the near-field from the far-field by comparing how close the hyperbola is from its asymptotes, by using a distance criterion. Additionally, this geometrical modeling stresses the limitation of the far-field assumption that is commonly assumed in sound source localization (*e.g.* planar wavefront) [VMRL03, NOK01]. Indeed, assuming the far-field condition, the accuracy of the DOAs obtained from ITDs decrease as the source gets closer to the microphones. Hence, rigorously our modeling should be based on (5.4), as it will be seen in Section 5.2.2, in order to control the robot accurately, independently of the distance to the sound source. Nonetheless, in the following, we also derive a model based on the ITD under the far-field approximation in (5.18).

Although the far-field assumption cannot always be guaranteed in the scene, this kind of model may highlight one of the key points of the sensor-based control scheme: these two models are equivalent from a control point of view. This property will be discussed exhaustively in Section 5.2.6.1.

## 5.2 An ITD-based interaction with a single source

### 5.2.1 ITD interaction matrix under the far-field assumption

For now on, this section is concerned about linking the motions of the robot to ITDs variation. Indeed building the interaction matrix is the foundation of the sensor-based framework. In the configuration discussed above, one can draw a model from the DOA  $\alpha$  that is connected to the ITD  $\tau$  through (5.18). The angle  $\alpha$  can be expressed with respect to  $\mathbf{X}_s(x_s, y_s)$  as :

$$\alpha = \text{atan2}(y_s, x_s). \quad (5.19)$$

Thus the time derivative of  $\alpha$  can be expressed as

$$\dot{\alpha} = \frac{\dot{y}_s x_s - \dot{x}_s y_s}{\ell^2}. \quad (5.20)$$

Equation (5.20) can also be expressed as a matrix relationship:

$$\dot{\alpha} = \mathbf{L}_\alpha \mathbf{v}_M \quad (5.21)$$

where  $\mathbf{L}_\alpha$  is the interaction matrix related to  $\alpha$  and  $\mathbf{v}_M = (v_x, v_y, v_z, \omega_x, \omega_y, \omega_z)$  the spatial velocity of the microphones pair. Once again we define  $\mathbf{u}_M = (v_x, v_y, \omega_z)$  that corresponds to the controlled DOF of the robot as described in Figure 5.1. In this equation, we can notice that  $v_z = 0, \omega_x = 0, \omega_y = 0$ , which limits the motion induced by the interaction matrix in the azimuth plane. From the kinematic equation

$$\dot{\mathbf{X}}_s = -\mathbf{v}_s - \boldsymbol{\omega}_s \times \mathbf{X}_s \Leftrightarrow \begin{cases} \dot{x}_s = -v_x - \omega_y z_s + \omega_z y_s \\ \dot{y}_s = -v_y - \omega_z x_s + \omega_x z_s \\ \dot{z}_s = -v_z - \omega_x y_s + \omega_y x_s \end{cases} \quad (5.22)$$

we obtain

$$\dot{\alpha} = \frac{-v_y x_s + v_x y_s - \omega_z (y_s^2 + x_s^2)}{x_s^2 + y_s^2}. \quad (5.23)$$

Eventually referring to (5.21), the interaction matrix related to  $\alpha$  is identified as:

$$\mathbf{L}_\alpha = \begin{bmatrix} \frac{y_s}{\ell^2} & -\frac{x_s}{\ell^2} & -1 \end{bmatrix} \quad (5.24)$$

in order to be consistent with  $\mathbf{u}_M$  (only the non-zero terms are considered). Thereafter, since the sound source position  $\mathbf{X}_s(x_s, y_s)$  is unknown, one can reshape the interaction matrix by using (5.2) so that  $\mathbf{L}_\alpha$  becomes

$$\mathbf{L}_\alpha = \begin{bmatrix} \frac{\sin \alpha}{\ell} & -\frac{\cos \alpha}{\ell} & -1 \end{bmatrix}. \quad (5.25)$$

This result can subsequently be exploited to obtain the interaction matrix related to  $\tau$ . Under the far-field assumption, the time derivative of (5.18) is:

$$\dot{\tau} = -A \sin \alpha \dot{\alpha}. \quad (5.26)$$

Considering that  $\alpha \in [0; \pi]$  (*i.e.*  $\sin \alpha > 0$ ), the interaction matrix  $\mathbf{L}_{\tau_f}$  under the far-field assumption is then given by

$$\mathbf{L}_{\tau_f} = -A \sin \alpha \mathbf{L}_\alpha = \begin{bmatrix} -\frac{A \sin^2 \alpha}{\ell} & \frac{A \sin \alpha \cos \alpha}{\ell} & A \sin \alpha \end{bmatrix}. \quad (5.27)$$

From (5.18) and the application of the trigonometric properties, the interaction matrix  $\mathbf{L}_{\tau_f}$  is eventually reduced to

$$\mathbf{L}_{\tau_f} = \begin{bmatrix} -\frac{\nu^2}{A\ell} & \frac{\tau\nu}{A\ell} & \nu \end{bmatrix} \quad (5.28)$$

where  $\nu = \sqrt{A^2 - \tau^2}$ . Similarly to binaural localization, the distance  $\ell$  that appears in  $\mathbf{L}_\alpha$  and  $\mathbf{L}_{\tau_f}$  is unknown. The actual interaction matrices related to  $\tau$  or  $\alpha$  are approximated as follows:

$$\widehat{\mathbf{L}}_\alpha = \begin{bmatrix} \frac{\sin \alpha}{\widehat{\ell}} & -\frac{\cos \alpha}{\widehat{\ell}} & -1 \end{bmatrix} \quad \text{and} \quad \widehat{\mathbf{L}}_{\tau_f} = \begin{bmatrix} -\frac{\nu^2}{A\widehat{\ell}} & \frac{\tau\nu}{A\widehat{\ell}} & \nu \end{bmatrix}. \quad (5.29)$$

This kind of approximation is comparable to image-based visual servoing where the depth of a point, which is unknown, appears in the interaction matrix of an image point. An overview on these two matrices highlights the non-admissible values for approximating their parameters. Indeed, for both matrices  $\widehat{\ell} \neq 0$  is a necessary condition to avoid elements of infinite values in the interaction matrix. Infinite values would nullify the control input. In the same way, the relation  $|\tau| < A$  related to the existence of the hyperbola  $\mathcal{H}$  appears in the interaction matrix of  $\tau$  as a condition of non-singularity (*i.e.*  $\mathbf{u}_M = \infty$ ).

### 5.2.2 ITD interaction matrix without assumption

An interaction matrix  $\mathbf{L}_{\tau_r}$  can also be obtained from the exact modeling of  $\tau$  given by (5.4). That is to say  $\mathbf{L}_{\tau_r}$  is computed without any assumption about the sound source being in the far-field or near-field. In this case the time derivative of  $\tau$  is

$$\dot{\tau} = \frac{1}{c}(\dot{\ell}_2 - \dot{\ell}_1) \quad (5.30)$$

By injecting (5.1) in (5.30), the derivative of  $\tau$  develops as

$$\dot{\tau} = \frac{1}{c} \left( \frac{2\dot{x}_s x_s + 2\dot{y}_s y_s + d\dot{x}_s}{2\ell_2} - \frac{2\dot{x}_s x_s + 2\dot{y}_s y_s - d\dot{x}_s}{2\ell_1} \right). \quad (5.31)$$

By using the kinematic equation given in (5.22), the latter equation becomes

$$\dot{\tau} = v_x \frac{x_s \tau - \frac{A}{2}(\ell_1 + \ell_2)}{\ell_1 \ell_2} + v_y \frac{y_s \tau}{\ell_1 \ell_2} + \omega_z \frac{\frac{A}{2}(\ell_1 + \ell_2) y_s}{\ell_1 \ell_2}. \quad (5.32)$$

Finally  $\mathbf{L}_{\tau_r}$  can be extracted from (5.32):

$$\mathbf{L}_{\tau_r} = \begin{bmatrix} \frac{x_s \tau - \frac{A}{2}(\ell_1 + \ell_2)}{\ell_1 \ell_2} & \frac{y_s \tau}{\ell_1 \ell_2} & \frac{\frac{A}{2}(\ell_1 + \ell_2) y_s}{\ell_1 \ell_2} \end{bmatrix}. \quad (5.33)$$

Similarly to (5.29), unknown parameters depending on  $\ell$  ( $x_s$ ,  $y_s$  and  $\ell_i$  are all function of  $\ell$  and  $\alpha$ ) and thus related to the sound source location, appears in  $\mathbf{L}_{\tau_r}$ . The approximated feature interaction matrix is then

$$\widehat{\mathbf{L}}_{\tau_r} = \begin{bmatrix} \frac{\widehat{x}_s \tau - \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)}{\widehat{\ell}_1 \widehat{\ell}_2} & \frac{\widehat{y}_s \tau}{\widehat{\ell}_1 \widehat{\ell}_2} & \frac{\frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2) \widehat{y}_s}{\widehat{\ell}_1 \widehat{\ell}_2} \end{bmatrix}. \quad (5.34)$$

For this type of interaction matrix, any approximation should respect the condition  $\widehat{\ell}_i \neq 0$  in order to avoid elements of infinite values.

It should also be noticed that this interaction matrix (5.33) can easily be linked to the interaction matrix under the far-field assumption (5.28). Under the far-field assumption, the following hypothesis  $\ell_1 \approx \ell_2 \approx \ell$  holds. Hence (5.33) can be rewritten as

$$\mathbf{L}_{\tau_r} \simeq \begin{bmatrix} \frac{x_s \tau - A\ell}{\ell^2} & \frac{y_s \tau}{\ell^2} & \frac{y_s A\ell}{\ell^2} \end{bmatrix}. \quad (5.35)$$

Afterwards, by replacing  $\tau$ ,  $x_s$  and  $y_s$  in (5.35) using (5.18) and (5.2), the interaction matrix becomes

$$\mathbf{L}_{\tau_r} \simeq \begin{bmatrix} -\frac{A \sin^2 \alpha}{\ell} & \frac{A \sin \alpha \cos \alpha}{\ell} & A \sin \alpha \end{bmatrix} = \mathbf{L}_{\tau_f} \quad (5.36)$$

that corresponds to (5.27), the interaction matrix with a far-field assumption.

### 5.2.3 Control scheme

These interaction matrices can then be used to design a control scheme that consists in positioning the robot so that a given condition characterized by  $\tau = \tau^*$  is satisfied. A positioning task is performed by considering a single ITD measurement  $\tau(t)$  extracted from the sound signal and by minimizing the error  $\|e(t)\|$ . Hence this task is characterized by an error

$$e(t) = \tau(t) - \tau^* \quad (5.37)$$

where  $\tau^*$  denotes the measurements for the desired ITD value. Similarly to the ILD case, we design a control scheme governed by the approximation of the interaction matrix  $\widehat{\mathbf{L}}_{\tau_j}$ ,  $j \in \{r, f\}$  in (5.29) or (5.34), in which the velocity of the microphones is computed from (5.21) as (see Chapter 3)

$$\mathbf{u}_M = -\lambda \widehat{\mathbf{L}}_{\tau_j}^+ e. \quad (5.38)$$

We recall that  $\lambda > 0$  is the gain that tunes the time to convergence. In the latter equation  $\widehat{\mathbf{L}}_{\tau_j}^+ \in \mathbb{R}^{3 \times 1}$  is the Moore-Penrose pseudo-inverse of  $\widehat{\mathbf{L}}_{\tau_j}$  (see Chapter 3.2.1). Thus, when considering the far-field assumption, the control scheme can be explicitly written as

$$\mathbf{u}_M = -\lambda \frac{A^2 \widehat{\ell}^2}{\nu^2 + \tau^2 \nu^2 + A^2 \widehat{\ell}^2 \nu^2} \begin{bmatrix} -\frac{\nu^2}{A\widehat{\ell}} \\ \frac{\tau \nu}{A\widehat{\ell}} \\ \nu \end{bmatrix} (\tau - \tau^*). \quad (5.39)$$

Otherwise, when considering the more general interaction matrix given in (5.34), the control scheme becomes

$$\mathbf{u}_M = \frac{-\lambda \widehat{\ell}_1 \widehat{\ell}_2}{(\widehat{y}_s \tau)^2 + \left(\frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)\widehat{y}_s\right)^2 + \left(\widehat{x}_s \tau - \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)\right)^2} \begin{bmatrix} \widehat{x}_s \tau - \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2) \\ \widehat{y}_s \tau \\ \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)\widehat{y}_s \end{bmatrix} (\tau - \tau^*). \quad (5.40)$$

#### 5.2.4 Task analysis

We analyze, here, the ITD-based positioning from (5.25), that offers more flexibility and is easier to interpret. From this interaction matrix it is possible to achieve positioning tasks so that  $\alpha$  reaches a particular desired value  $\alpha^*$ . By using only one feature from a single sound source, it can be intuitively expected that several sensor poses exist so that  $\alpha = \alpha^*$ , that is a direct consequence of ITDs properties detailed previously. Actually only one DOF can be controlled by using one sound source. However, it is possible to extend the approach to control more DOF by simply increasing the number of sound sources as it will be performed for Section 5.3. Here we are attached to describe this task geometrically from virtual linkages approach given in Chapter 3.

This analysis is based on the vector subspace  $\mathbf{S}^*$ , where each column characterizes a motion for which  $\alpha$  stays constant. When considering only one sound source,  $\mathbf{L}_\alpha$  given by (5.25) is a rank-one matrix. Consequently  $\mathbf{S}^* \in \mathbb{R}^{3 \times 2}$  implies a virtual link of class 2:

$$\mathbf{S}^* = \text{Ker } \mathbf{L}_\alpha. \quad (5.41)$$

Hence, this subspace is defined as

$$\mathbf{S}^* = \begin{bmatrix} \mathbf{u}_{M_1}^* & \mathbf{u}_{M_2}^* \end{bmatrix} = \begin{bmatrix} v_{x1} & v_{x2} \\ v_{y1} & v_{y2} \\ \omega_{z1} & \omega_{z2} \end{bmatrix}. \quad (5.42)$$

which corresponds analytically to

$$\mathbf{S}^* = \begin{bmatrix} \cos \alpha & \ell \sin \alpha \\ \sin \alpha & -\ell \cos \alpha \\ 0 & 1 \end{bmatrix}. \quad (5.43)$$

The first column of  $\mathbf{S}^*$  induces a translation motion along the sound source direction, that is along  $\overrightarrow{M\mathbf{X}_s}$ . The second column describes a rotation around  $\overrightarrow{z_M}$  axis combined with a translation. This last motion can be geometrically represented by a circle of radius  $\ell$  centered on  $\mathbf{X}_s$ . Hence, similarly to the ILD case, such task consists principally in orienting the robot with respect to the source location.

Thus, any linear combination of these two motions implies infinite poses to complete a task  $\alpha = \alpha^*$  (see Figure 5.3), from which we set up a first proposition.

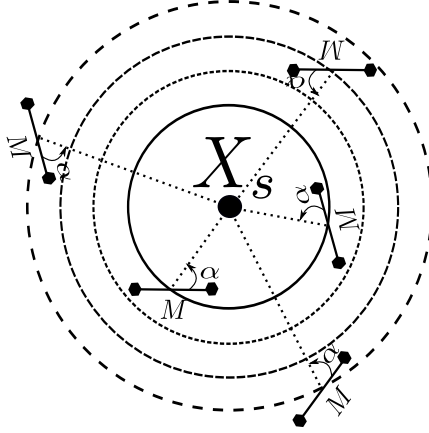


Figure 5.3: With one sound source several poses exists for a given positioning task expressed by  $\alpha = \alpha^*$ . These poses lie on circles of radius  $\ell$  centered on the sound source location.

**Proposition 2** *For each position of the microphones defined by  $\mathbf{M}$ , there exists one orientation  $\theta_{\mathbf{M}}$  of the microphones such that  $\alpha = \alpha^*$ .*

### 5.2.5 Stability analysis

In this part, we analyze the stability properties of our sensor-based control through the interaction matrix modeled in (5.34) and (5.29). The stability properties tailor the approximation of the unknown parameters in the feature Jacobian matrix. With the same reasoning developed for ILD context, in the previous chapter, the Lyapunov global asymptotic stability condition is obtained when

$$\mathbf{L}_\tau \widehat{\mathbf{L}}_\tau^+ > 0, \quad (5.44)$$

knowing that  $\widehat{\mathbf{L}}_\tau \in \mathbb{R}^{1 \times 3}$ . In order to determine the stability conditions, let us focus first on the Jacobian matrix related to the DOA  $\alpha$ . In this case, the stability is obtained when

$$\mathbf{L}_\alpha \widehat{\mathbf{L}}_\alpha^+ > 0. \quad (5.45)$$

From (5.28) and (5.29), we have

$$\mathbf{L}_\alpha \widehat{\mathbf{L}}_\alpha^+ = \begin{bmatrix} \frac{\sin \alpha}{\ell} & -\frac{\cos \alpha}{\ell} & -1 \end{bmatrix} \begin{bmatrix} \frac{\widehat{\ell} \sin \alpha}{1 + \widehat{\ell}^2} \\ -\frac{\widehat{\ell} \cos \alpha}{1 + \widehat{\ell}^2} \\ -\frac{\widehat{\ell}^2}{1 + \widehat{\ell}^2} \end{bmatrix}. \quad (5.46)$$

Equation (5.46) can be simplified to

$$\mathbf{L}_\alpha \widehat{\mathbf{L}}_\alpha^+ = \frac{\widehat{\ell} + \ell \widehat{\ell}^2}{\ell(1 + \widehat{\ell}^2)}. \quad (5.47)$$

Thus, it appears that for ensuring the stability condition  $\mathbf{L}_\alpha \widehat{\mathbf{L}}_\alpha^+ > 0$ , it is sufficient to set  $\widehat{\ell} > 0$ . However, this demonstration is valid under the condition that the DOA injected in  $\widehat{\mathbf{L}}_\alpha$  is not too coarse. Indeed if we consider an approximation  $\widehat{\alpha}$  estimated from  $\tau$ , (5.47) is more exactly written as

$$\mathbf{L}_\alpha \widehat{\mathbf{L}}_\alpha^+ = \frac{\widehat{\ell} \sin \alpha \sin \widehat{\alpha} + \widehat{\ell} \cos \alpha \cos \widehat{\alpha} + \ell \widehat{\ell}^2}{\ell(1 + \widehat{\ell}^2)}. \quad (5.48)$$

Considering that  $\ell > 0$  and  $\widehat{\ell} > 0$ , the stability is ensured as soon as  $\text{sign}(\cos \widehat{\alpha}) = \text{sign}(\cos \alpha)$  and  $\text{sign}(\sin \widehat{\alpha}) = \text{sign}(\sin \alpha)$ . This interesting result confirms that even with rough approximation of  $\alpha$ , the controller should be able to converge towards the desired configuration. Generally, any technique of ITD estimation (referenced in Chapter 2.2), and by extension DOA estimation, provides sufficiently accurate results so that  $\text{sign}(\cos \widehat{\alpha}) = \text{sign}(\cos \alpha)$ . The second condition  $\text{sign}(\sin \widehat{\alpha}) = \text{sign}(\sin \alpha)$  is as expected related to the *front-back ambiguity* inherent to ITDs. As a consequence, under the far-field assumption, the stability of the controller based on  $\widehat{\mathbf{L}}_{\tau_r}$  in (5.29) is ensured as soon as  $\widehat{\ell} > 0$ , the DOA  $\alpha$  and then the ITD  $\tau$  is not too coarse (see (5.18) and Figure 5.2) and the *front-back ambiguity* is solved.

For generalized stability properties (*i.e.*, valid in the near-field), the Lyapunov stability should also be evaluated on (5.34). First, to ease the comprehension and the reading of the analysis, let us denote the interaction matrix  $\widehat{\mathbf{L}}_\tau$  as

$$\widehat{\mathbf{L}}_\tau = \begin{bmatrix} \frac{\widehat{N}_1}{\widehat{\ell}_1 \widehat{\ell}_2} & \frac{\widehat{N}_2}{\widehat{\ell}_1 \widehat{\ell}_2} & \frac{\widehat{N}_3}{\widehat{\ell}_1 \widehat{\ell}_2} \end{bmatrix} \quad (5.49)$$

where  $\widehat{N}_1 = \widehat{x}_s \tau - \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)$ ,  $\widehat{N}_2 = \widehat{y}_s \tau$  and  $\widehat{N}_3 = \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)\widehat{y}_s$ . Similarly  $\mathbf{L}_\tau$  can be defined with  $N_1$ ,  $N_2$ ,  $N_3$ ,  $\ell_1$  and  $\ell_2$ . From this convention of writing, (5.44) becomes

$$\mathbf{L}_{\tau_r} \widehat{\mathbf{L}}_{\tau_r}^+ = \frac{\widehat{\ell}_1 \widehat{\ell}_2 N_1 \widehat{N}_1 + N_2 \widehat{N}_2 + N_3 \widehat{N}_3}{\ell_1 \ell_2 \widehat{N}_1^2 + \widehat{N}_2^2 + \widehat{N}_3^2}. \quad (5.50)$$

Considering of course  $\ell_i > 0$  and  $\widehat{\ell}_i > 0$ , the Lyapunov stability is ensured as soon as

$$N_1 \widehat{N}_1 + N_2 \widehat{N}_2 + N_3 \widehat{N}_3 > 0. \quad (5.51)$$

The latter equation can be developed as:

$$y_s \widehat{y}_s \tau^2 + \frac{A^2}{4} y_s \widehat{y}_s (\widehat{\ell}_1 + \widehat{\ell}_2) (\ell_1 + \ell_2) + (x_s \tau - \frac{A}{2}(\ell_1 + \ell_2)) (\widehat{x}_s \tau - \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)) > 0. \quad (5.52)$$

Hence, it can be deduced that a sufficient condition of stability is obtained with  $\text{sign}(\widehat{x}_s) = \text{sign}(x_s)$  and  $\text{sign}(\widehat{y}_s) = \text{sign}(y_s)$ . Indeed (5.52) is always positive for these conditions, since  $(\widehat{x}_s \tau - \frac{A}{2}(\widehat{\ell}_1 + \widehat{\ell}_2)) < 0$  and  $(x_s \tau - \frac{A}{2}(\ell_1 + \ell_2)) < 0$  owing to  $\frac{\ell_1 + \ell_2}{2} \geq x_s$  and  $|\tau| < A$  (see (5.10)).

In the end, the stability conditions of the controller, assuming the far-field approximation or the proper modeling of the ITD are exactly the same. Indeed, from



(5.2) the conditions  $sign(\hat{x}_s) = sign(x_s)$  and  $sign(\hat{y}_s) = sign(y_s)$  can be rewritten as:

$$\begin{cases} sign(\hat{\ell} \cos \hat{\alpha}) = sign(\ell \cos \alpha) \\ sign(\hat{\ell} \sin \hat{\alpha}) = sign(\ell \sin \alpha) \end{cases} \quad (5.53)$$

which amounts to  $\hat{\ell} > 0$ ,  $sign(\cos \hat{\alpha}) = sign(\cos \alpha)$  and  $sign(\sin \hat{\alpha}) = sign(\sin \alpha)$ , the conditions described for the control scheme based on the far-field assumption. Both ITDs models stress that sufficient stability conditions are obtained when the source location is disambiguated on the hyperbola  $\mathcal{H}$ . The conditions  $sign(\hat{x}_s) = sign(x_s)$  and  $sign(\hat{y}_s) = sign(y_s)$  are sufficient. The first condition is easily fulfilled from the sign of  $\tau$  as discussed above in (5.13). However, the second condition related to the *front-back ambiguity* cannot be ensured from ITD measurements. Nonetheless, experimentally, it may be possible to reach a pose where all the above conditions are completed, as already demonstrated in Chapter 4.2.5. Eventually, the conditions resulting from the current study underline infinite solutions ensuring global asymptotic stability, for both cases of ITD modeling. From these results, one can notice the strong similarity between the stability properties for ITDs and ILDs. For both cues, a task mainly consists in orienting a robot towards a given azimuth direction. Hence, as soon as the actual source orthant is known (*i.e.*  $sign(\hat{x}_s) = sign(x_s)$  and  $sign(\hat{y}_s) = sign(y_s)$ ), an adequate robot motion can be found so that the error converges towards 0.

## 5.2.6 Experimental results

### 5.2.6.1 Distance of observation evaluation

Before experimenting the proposed control scheme, it is essential to evaluate the difference between the interaction matrix based on the far-field assumption  $\widehat{\mathbf{L}}_{\tau_f}$  and  $\widehat{\mathbf{L}}_{\tau_r}$  extracted from the actual geometry of the ITD. We then tested these two models through simulations. The simulation environment designed from *Roomsimove* [BOV12] consists in an anechoic room in which a positioning task is performed with respect to a speech sound  $\mathbf{X}_s$ . We considered anechoic conditions, since these simulations aim principally to evaluate the accuracy of these two models with respect to the distance of sound observation. The positions of  $\mathbf{X}_s$  are defined so that  $\alpha$  varied by  $18^\circ$  at given distance  $\ell$  within the range  $[18^\circ, 162^\circ]$  as depicted in Figure 5.4. When considering one sound source, since the task mainly consists in orienting the robot towards a given direction, we controlled only the rotational velocity, which lead to

$$\omega_{z_f} = -\lambda\nu(\tau - \tau^*), \quad (5.54)$$

where  $\omega_{z_f}$  is the control based on the far-field assumption. Similarly the control ensued from the rigorous geometrical definition of ITDs  $\omega_{z_r}$  is defined as:

$$\omega_{z_r} = -\lambda \frac{A}{2} \frac{(\hat{\ell}_1 + \hat{\ell}_2) \hat{y}_s}{\hat{\ell}_1 \hat{\ell}_2} (\tau - \tau^*). \quad (5.55)$$

In (5.54), the parameters to be estimated are set to  $\hat{x}_s = -1 \times \text{sign}(\tau)$  and  $\hat{y}_s = 1$ .  $\hat{\ell}_1$  and  $\hat{\ell}_2$  can afterwards be deduced by using (5.1). The task consists in facing the sound source, which imply that  $\tau^* = 0$ . The ITDs  $\tau_i$  are estimated in real-time, for both configurations, by using GCC-PHAT algorithm (see Chapter 2.2.3). The orienting task was repeated from far-field and near-field distances. More explicitly  $\ell$  was set respectively at 0.5 m of the microphones in the near-field and 2 m for the far-field with  $d = 0.3$  m. In the results illustrated in Table 5.1, we also compared these models to an open loop localization approach, in which the accuracy of the azimuth is evaluated from the AEG model (see Chapter 2.2.2). The absolute orientation errors (azimuth errors in the localization case) in degree is reported.

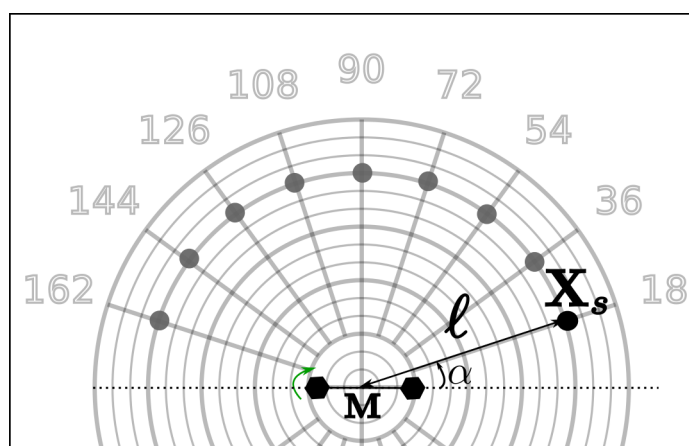


Figure 5.4: A simulated task that consists in facing a sound source ( $\tau^* = 0$ ) in an anechoic room. The sound source positions are varied so that  $\alpha \in [18^\circ, 162^\circ]$  by step of  $18^\circ$  and  $\ell \in \{0.5, 2\}$  m for a near-field and far-field observation distance.

The simulations show that both control schemes  $\omega_{z_r}$  and  $\omega_{z_f}$  give excellent results. This result could be expected from the stability conditions that are exactly the same for the two approaches. From the stability perspective, the actual interaction matrix related to the far-field assumption  $\mathbf{L}_{\tau_f}$  is just an approximation of  $\mathbf{L}_{\tau_r}$ , so that it can be written

$$\mathbf{L}_{\tau_f} = \widehat{\mathbf{L}}_{\tau_r}. \quad (5.56)$$

The values  $\ell$  (and  $\tau$ ) are directly reflected in  $\widehat{\mathbf{L}}_{\tau_r}$  through  $\hat{x}_s$ ,  $\hat{y}_s$  and  $\hat{\ell}_i$ . Hence, without violating the stability conditions of  $\mathbf{L}_{\tau_r}$ , it is ensured that a control scheme based on  $\mathbf{L}_{\tau_f}$  will converge under near-field conditions. This outcome is also valid for  $\widehat{\mathbf{L}}_{\tau_f}$  under satisfying stability conditions. Consequently the ITD-based positioning task can be performed accurately under a far-field assumption in the near-field. Eventually, the results exhibited by our approach can also be compared to a more conventional positioning task based on localization. Conversely to our method, in the near-field, the localization accuracy is particularly degraded, because of the far-field assumption. This result corroborates the geometrical modeling of ITDs detailed in Section 5.1.2, in which the hyperbola  $\mathcal{H}$  does not fit with its asymptotes in the near-field. Although the simulation results of localization remains acceptable in the

		$\alpha$ (degrees)								
		18	36	54	72	90	108	126	144	162
Near-field	$\mathbf{u}_f$	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1
	$\mathbf{u}_r$	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1
	<b>loc</b>	7.1	8.2	6.3	3.9	0	3.9	6.8	8.2	7.1
Far-field	$\mathbf{u}_f$	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1
	$\mathbf{u}_r$	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1	< 1
	<b>loc</b>	< 1	1.1	< 1	< 1	< 1	< 1	< 1	1.1	< 1

Table 5.1: Absolute error of the orientation task described in Figure 5.4 considering a control scheme using a far-field assumption  $\omega_{z_f}$  or a control scheme without assumption  $\omega_{z_r}$ . These two methods are compared to a baseline approach based on localization **loc**.

near-field (in anechoic conditions), when facing real environments the errors may be accentuated by reverberation, noises or other perturbations.

As a direct consequence of these results, we can already emphasize the potential robustness of our control scheme towards approximations or inaccurate estimations. In the following, all the experiments and analysis are based on  $\widehat{\mathbf{L}}_{\tau_f}$ , that has a more convenient form. For more simplicity, we rename  $\mathbf{L}_{\tau_f}$  as  $\mathbf{L}_{\tau}$ .

### 5.2.6.2 Preliminaries: robot modeling and control scheme

To apply the proposed framework, the same system as in Chapter 4.2.5 is modeled: a non-holonomic unicycle robot endowed with two microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$  as illustrated on Figure 5.5. In this context, we aim to control the two DOFs of the robot characterized by the control input  $\dot{\mathbf{q}}$  given by  $(u, \omega)$ , respectively the translation and rotation velocities. In addition to  $\mathcal{F}_m$  related to the microphones, the frame  $\mathcal{F}_r(\vec{x}_R, \vec{y}_R)$  is attached to the robot.  $D_x$  denotes the distance between the center of the robot  $\mathbf{R}$  and the midpoint  $\mathbf{M}$  of the microphones.

From this modeling, the control input  $\dot{\mathbf{q}}$  can be computed from the following relationship:

$$\dot{\tau} = \mathbf{J}_{\tau} \dot{\mathbf{q}} \quad (5.57)$$

where  $\mathbf{J}_{\tau}$  is the Jacobian feature matrix introduced in Chapter 3.2.1 that is equal to

$$\mathbf{J}_{\tau} = \mathbf{L}_{\tau} \mathbf{J}_{\mathbf{r}}. \quad (5.58)$$

$\mathbf{J}_{\mathbf{r}}$  is the robot Jacobian obtained from Figure 5.5 as

$$\mathbf{J}_{\mathbf{r}} = \begin{bmatrix} 0 & D_x \\ 1 & 0 \\ 0 & 1 \end{bmatrix}. \quad (5.59)$$

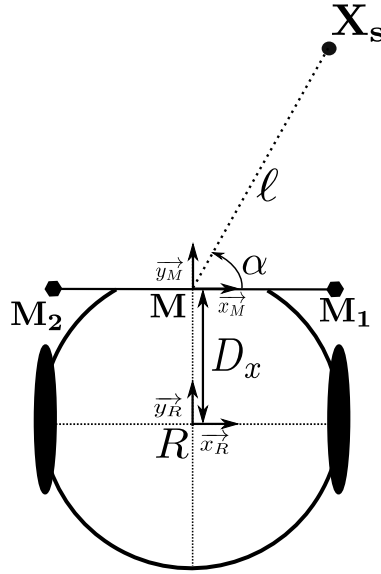


Figure 5.5: modeling of the robotic platform

Hence, the control scheme of the robot is

$$\dot{\mathbf{q}} = -\lambda \widehat{\mathbf{J}}_{\tau}^{\dagger} (\tau - \tau^*). \quad (5.60)$$

As mentioned in the task analysis, a control input characterized by the angular velocity  $\omega$  only is sufficient to achieve any task involving one sound source. In that particular case where  $u$  is always equal to 0 and the Jacobian matrix reduces to

$$\widehat{\mathbf{J}}_{\tau} = \frac{A\widehat{\ell}\nu - D_x\nu^2}{A\widehat{\ell}}. \quad (5.61)$$

The control input becomes

$$\dot{q} = -\lambda \frac{1}{\nu \left(1 - \frac{D_x\nu}{A\widehat{\ell}}\right)} (\tau - \tau^*) \quad (5.62)$$

Excluding the degenerate case where  $\alpha = 0$  or  $\alpha = 180$  (i.e.,  $\nu = 0$ ), singularities of the interaction matrix could in principle occur if  $\widehat{\ell} < D_x$ . Hopefully, this is impossible in practice. Indeed, knowing that  $0 \leq \nu \leq A$ , the denominator of (5.62) can never vanish as soon as  $\widehat{\ell} > D_x$ . In the same way, the global asymptotic stability of the system is satisfied as soon as  $\ell > D_x$  and  $\widehat{\ell} > D_x$ .

When the two DOF of the robot ( $u, \omega$ ) are controlled, the Jacobian matrix is given by

$$\widehat{\mathbf{J}}_{\tau} = \begin{bmatrix} \frac{\tau\nu}{A\widehat{\ell}} & \frac{A\widehat{\ell}\nu - D_x\nu^2}{A\widehat{\ell}} \end{bmatrix} \quad (5.63)$$

and the control input is obtained as:

$$\dot{q} = -\lambda \frac{\tau^2 + (\nu D_x - A\widehat{\ell})^2}{A^2\widehat{\ell}^2} \begin{bmatrix} \frac{\tau\nu}{A} \\ -\nu\left(\frac{\nu D_x}{A} - \widehat{\ell}\right) \end{bmatrix} (\tau - \tau^*) \quad (5.64)$$

The interaction matrix is not singular as soon as  $\hat{\ell} > 0$ , which corresponds to the Lyapunov stability condition already demonstrated previously.

### 5.2.6.3 ITD estimation and tracking

Prior to the experimental validations, it is essential to characterize the sound processing and more particularly ITDs estimation. Indeed unlike ILDs that can be computed uniquely by integrating the sound signal over a particular time window, ITDs are estimated from a more complex process. Recalling the methods of ITDs estimation detailed in Chapter 2.2.3, from GCC-PHAT [KC76], the ITD  $\tau$  can be obtained by comparing the temporal signals  $x_1(t)$  from the microphone  $\mathbf{M}_1$  and  $x_2(t)$  from  $\mathbf{M}_2$  in the spectral domain with:

$$\mathbf{R}_{1,2}(\tau) = \max_l \sum_f^F \psi(f) X_1(f, l) X_2^*(f, l) e^{j\varphi(\tau)}. \quad (5.65)$$

$X_1(f, l)$  and  $X_2^*(f, l)$  are respectively the Fourier transform of  $x_1(t)$  and the conjugate of the Fourier transform of  $x_2(t)$ . The maximum peak of the GCC-PHAT function gives an estimation of the actual ITD and can therefore be written as:

$$\hat{\tau} = \underset{\tau}{\operatorname{argmax}} \mathbf{R}_{1,2}(\tau) \quad (5.66)$$

with  $\tau \in [-A, A]$  corresponding to a sound direction of arrival  $\alpha$  from  $0^\circ$  to  $180^\circ$ . In ideal conditions, the maximum peak of the latter function should correspond to the ITD of the sound source as exposed in Figure 5.6a.

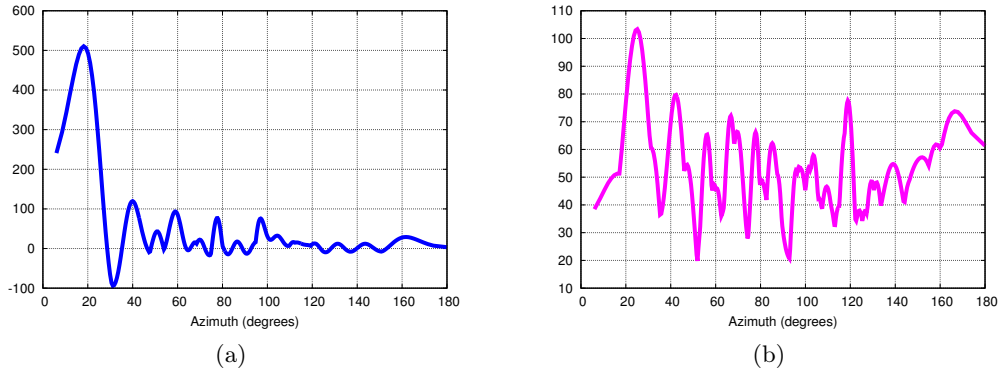


Figure 5.6: Output of the GCC-PHAT function under anechoic conditions (a) and reverberant conditions (b).

In real world conditions, the result of the GCC-PHAT function is more contrasted. Instead of having one dominating peak, which makes immediate the estimation of the actual ITD, several plausible but spurious peaks appear because of reverberation and noise (see Figure 5.6b). These conditions may also alter the actual peak related

to the source that would be less sharp and accurate. Fortunately, this problem can be handled by our control framework that is robust to approximate ITDs.

Therefore, under these conditions,  $p$  peaks ( $p > 1$ ) may be considered among which the correct peak should be found. Furthermore, when considering specific signals such as speech, it is very likely to record sparse data (temporally and spectrally). Not all the frames processed contain relevant information, since speech is a non-stationary and intermittent signal.

This emphasizes the need of a tracking algorithm to detect correct ITDs. Independently of the intrinsic values of the features, AS, as many other sensor-based approaches, is built upon a correct tracking of the input features. Tracking ITDs corresponding to an actual source among a set of observations may be an issue for the correct achievement of a desired task. If ITDs that do not correspond to the source are tracked the positioning task would certainly fail. In robot audition literature, the problem of tracking ITDs is quite new, with the recent interest in *active audition* and moving source/robot as stated in Chapter 1.2. The common approach consists in probabilistic techniques such as the particle filter used in [VMR07] and [MP10]. These two approaches are complemented with a speech activity detection to account for periods of silence. In [EMN<sup>+</sup>15] a probability hypothesis density tracker is proposed in order to skip the activity detection. The aforementioned approaches are relying on array-based approach for more robustness. In binaural context [PDA12] and [NCVC16] are based on a mixture of Kalman filters. However these methods have only been partially validated in real environments [PBD<sup>+</sup>14]. Being able to accurately track dynamic ITDs is still a challenging task in robot audition.

Yet, AS can provide added value to this tracking problem. One of the benefit of coupling motion and perception, as in AS, lies in the prediction that limits the scope of erroneous measurements. More specifically, the Jacobian matrix  $\mathbf{J}_\tau$  can be used to infer the evolution of the tracked ITDs in the next time frame. Given  $\tau$  as the state  $x$  and the velocity  $\dot{\mathbf{q}}$  applied to the robot, a local prediction model based on the Jacobian matrix (5.58) is simply given as follows:

$$\begin{cases} \dot{x}(k) = \widehat{\mathbf{J}}_\tau \dot{\mathbf{q}} \\ x(k+1) = x(k) + T_e \dot{x}(k) \end{cases} \quad (5.67)$$

in which  $T_e$  refers to the sampling time of the control loop. Nonetheless the predicted  $\tau$  is not as accurate as the genuine ITD, because of the approximation  $\widehat{\mathbf{J}}_\tau$  used in (5.60) in which  $\ell_i = \widehat{\ell}_i$ . Several methods such as state space observer [DLORG08] could be used to obtain a better estimation of  $\widehat{\ell}_i$ , but the closed-loop control scheme is sufficiently robust to cope with a rough approximation of  $\ell_i$ . Despite this rough approximation, the prediction step gives useful information that can also be used when the source is not active. If no ITD measurement "fit" with the predicted value, it is very likely that the source is inactive. Hence, the unavailable ITD could be replaced by this prediction. Without complex tracking methods, it should then be possible to track the correct ITDs in realistic environments.

In our application, the tracking procedure is divided into two steps. In the first step, the goal is to find the correct ITDs with respect to the number of active sound sources defined beforehand. This step assumes that the microphones and the sound

source(s) are not moving. Then the set of peaks obtained from a GCC function is observed during a given number of time frames. A history pooling algorithm based on the Euclidean distance between ITDs is then applied to detect the frequency of appearance of a given ITD  $\tau_i$ . By combining this frequency to the average appearance rank of each  $\tau_i$ , the most probable ITD is then selected. This method is applied to the initial pose to retrieve  $\tau(t_0)$  but could also be applied to retrieve each  $\tau^*$  when the desired pose is not characterized by an obvious ITD. For some simple cases, each  $\tau^*$  is defined manually, for instance for a task that consists in orienting the robot towards the sound source (*e.g.*,  $\tau^* = 0$ ). The second step is used during the robot motion, and consists in finding the genuine ITDs among the set of observations given by the GCC-PHAT function. Knowing the ITDs previous value, we simply select the closest peak in the current frame, by taking into account our predicted ITDs.

It is clear that this tracking method is not optimal. More evolved solutions could complement our approach with the methods available in the literature (particle filtering, probabilistic data association...). They would, however, increase the computational cost of the control scheme. We have preferred the simple and efficient solution exposed above in order to stress the added value of our approach, and more particularly the robustness of the control scheme to inaccuracies and/or punctual errors.

#### 5.2.6.4 Experimental setup

In order to further validate the effectiveness of our approach we conducted experiments on the *Pioneer 3DX* robot. Two microphones connected to a sound card 8SoundsUSB [AC<sup>+</sup>] were used. The sound card operates at a frequency of 48 kHz, and provides windows frames of 256 samples. The ITD is computed from 10 consecutive windows frames (*e.g.* 50 ms), that are sub-sampled at 16 kHz. Processing the sound signal at a frequency of 16 kHz gives two advantages: better results are obtained from uttered speech that is less sparse at low frequencies and the processing time is reduced with less samples to analyze. Consequently, each control loop is processed at an average frequency of 12 Hz. The tests were conducted in a room with a reverberation time  $RT_{60} \approx 580$  ms. Moreover, the measured SNR is around 20 dB

$d$	0.31 m
$c$	343 m.s <sup>-1</sup>
$\widehat{\ell}$	1 m
$A$	0,00090379 s
$\lambda(x)$	$5e^{(-4000x)}$

Table 5.2: Experimental settings

in presence of typical noise such as computer noise and ventilation in the room. The parameters used for the experiments are given in Table 5.2. An adaptive gain  $\lambda(x)$

in which  $x$  refers to the norm of the error  $e$  is used to smooth the robot motion.

### 5.2.6.5 Typical ITD-based positioning task

In this experiment we considered one sound source that corresponds to a female voice recording of 10 s played in loop. The task to be performed is to face the sound source,  $\tau^*$  is set to 0. The two DOF are controlled from the control input given by (5.64). The experiment can be divided into two steps.

In the first step the robot is randomly oriented with respect to the loudspeaker. Nonetheless this orientation is set so that the source stays in the front-side of the robot. In this part of the experiment the robot correctly positioned itself in the direction of the static sound source, as illustrated by the Figure 5.7.

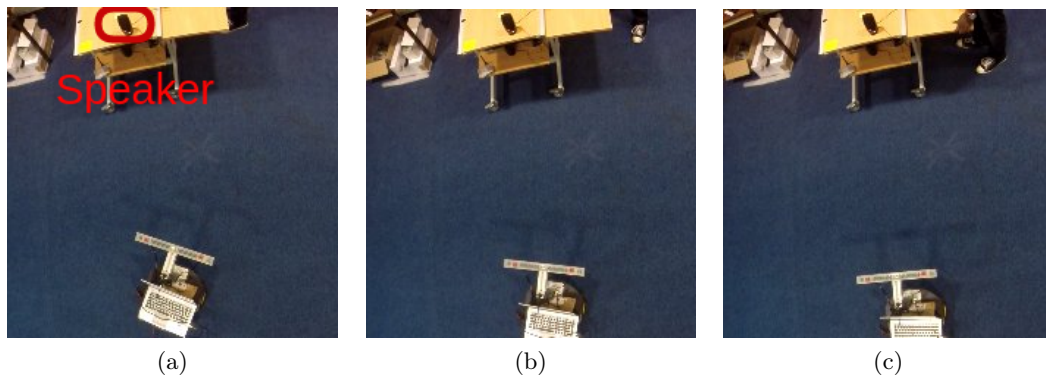


Figure 5.7: First phase of the experiment: the robot accurately orients towards the sound source

Subsequently the sound source was moved from one side in the environment and at different distances. As a result, the robot constantly moved in the direction of the source (see Figure 5.8). The robot accurate behavior is supported by the data exposed in Figure 5.9, that are extracted from the experiment. During the first 5 s, the error follows an exponential decrease pattern as modeled in the control scheme. In the following, when the source moves, the error is compensated by the robot motion. As already mentioned in the previous chapter, in the ILD case, there is still a small tracking error, that is illustrated by the error curve that do not stay at 0. Nonetheless, the task is correctly achieved since the lag between the robot orientation and the desired one did not exceed  $10^\circ$  despite the low dynamic response of the mobile robot. Knowing that no specific information about the source motion is used in the control scheme, the behavior of the robot remains satisfactory. Furthermore, this experiment acknowledges the relevance of the tracking method. We can notice in Figure 5.9c several false measurements represented by the green dots while the actual ITD is accurately tracked. These spurious measurements may be caused by reverberation, or simply noise. This noise can also be noticed by the aligned green dots at  $\alpha = 90^\circ$  during all the frames of the task. This diffuse noise is certainly caused by the ventilation system in the room. It can be noted that



no particular signal enhancement has been used for this experiment. Unlike the experiments proposed in the previous chapter, we are not attached here to address the *front-back ambiguity*. The singularities related to ITD hyperbola (*i.e.*,  $\tau \pm A$  or  $\alpha = 0^\circ$ ,  $\alpha = 180^\circ$ ) would make the control scheme unstable.

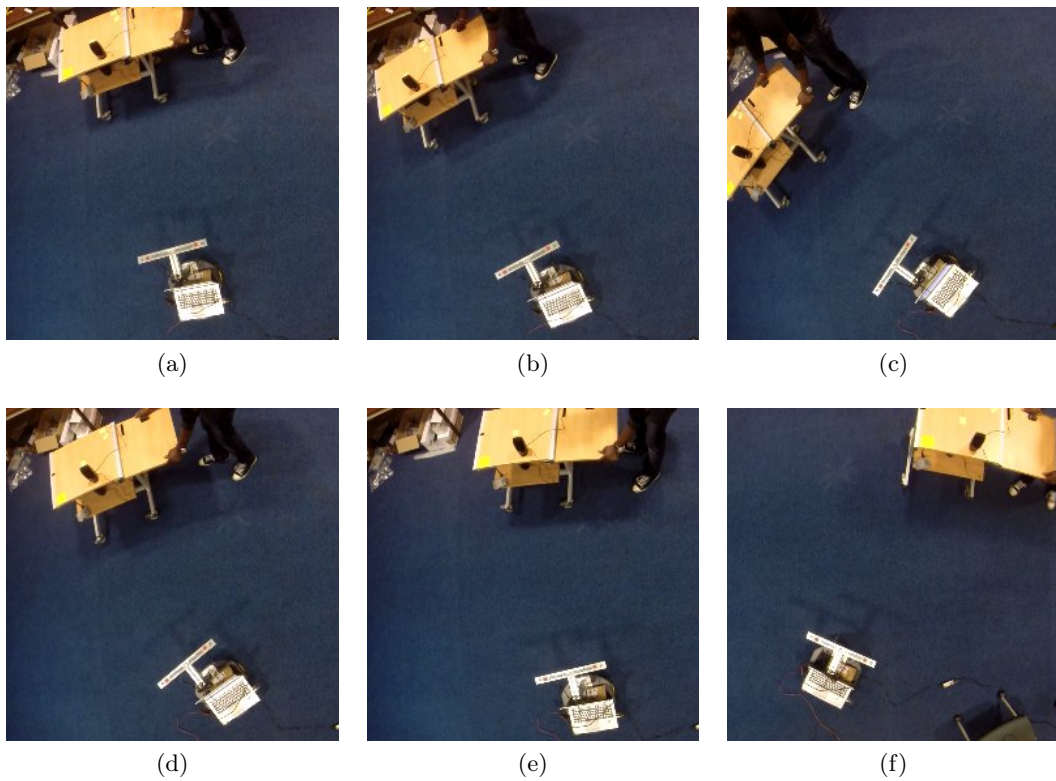


Figure 5.8: Second phase of the experiment: the robot is accurately tracking the sound source that is moving

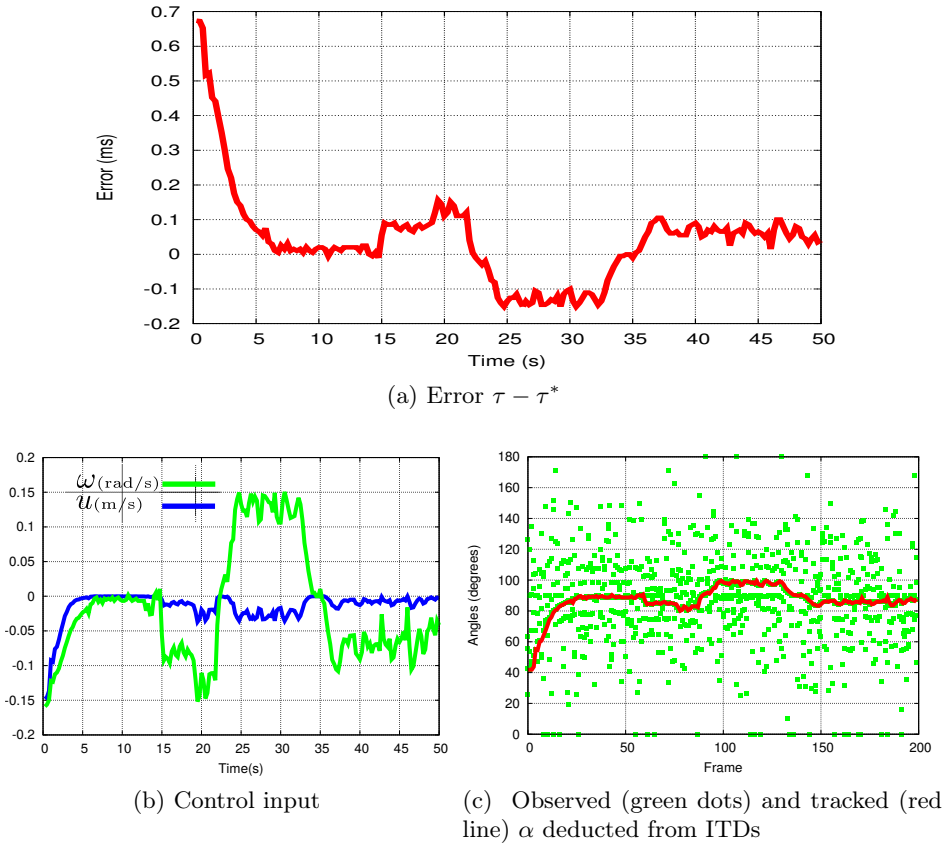


Figure 5.9: Experimental result of the ITD-based task illustrated in Figures 5.7 and 5.8

### 5.2.7 Numerical evaluation

The proposed framework has also been evaluated in a numerical simulation. An acoustic room of 5 by 5 m<sup>2</sup> is simulated, in which (a sound source corresponding to the same female speech as in the previous experiments) is played in loop. Each ITD is extracted from a frame of 100 ms. This frame size was chosen for real-time purpose. The experiment is repeated several times with 12 source locations so that the azimuth angle at the starting pose of the microphones varies between 30° and 150° with a resolution of 10° (the azimuth angle 90° is not considered). The goal of this task is to orient a robot towards a given sound source (i.e.  $\tau^* = 0$  or  $\alpha = 90^\circ$ ). The correct achievement of the tasks as well as the erroneous and missing ITD are evaluated. A measurement is considered erroneous if the error between the actual ITD and the estimated ITD leads to an error of 5° or more in the corresponding DOA  $\alpha$ . A task is said to be completed when  $\tau = \tau^*$  at the end of the servo.

#### Robustness to noise and reverberation

Table 5.3 sums up the results of the simulation. As expected the performance is decreased in presence of a high level of noise coupled with reverberation. Under 20 dB

SNR(dB)	RT <sub>60</sub> (s)				Total
	0	0.05	0.1	0.2	
30	1	1	1	1	1
25	1	1	1	1	1
20	1	0,916	0,667	0	0,646
15	1	0,667	0,416	0,083	0,541
10	0,667	0,167	0	0	0,208
<b>Total</b>	0,944	0,625	0,514	0,347	

Table 5.3: Rate of achieved tasks considering several reverberation and noise conditions

of SNR, the task is affected by reverberation. However, most of the time, the robot did not reach the correct pose because of a wrong initialization of ITD measurements (e.g  $\alpha(t = 0) = 90^\circ$ ). These failures stress the importance of being able to track the correct ITDs and more importantly to initialize correctly the tracking algorithm. Hence, a more elaborated initialization step could improve the rate of completion of the task for low SNR. Globally, we notice that this approach is more sensitive to a high level of noise than to reverberation. This result fits with the method chosen for the ITD calculation (GCC-PHAT), known to be less robust to noise. The PHAT processor whitens the signal for all frequencies without any selectivity of frames containing noise or speech. Approaches such as RWPHAT (see Chapter 2.2.3) would undeniably improve these results. On the other hand, in a context suitable for GCC-PHAT, that is to say with a high SNR, the positioning task is accurately performed.

#### **Robustness to erroneous measurements**

A more thorough study of the previous results confirms that the control scheme is able to cope with erroneous/missing measurements. Among the achieved tasks, we focus now on the rate of erroneous/missing ITDs. The Figure 5.10 illustrates this rate for a noise level of 30 and 25 dB. Lower SNR are not considered in this figure since the different number of achieved tasks would bias the global analysis. Although these tasks were all achieved correctly, several erroneous or missing measurements are faced by the control scheme mainly because of speech pauses. As expected, this rate of erroneous ITDs increases logically with higher level of noise and reverberation. But the control scheme is still able to complete the task by using the prediction derived from the interaction matrix, even for cases where nearly 20% of the measurements are missing or erroneous. This result shows the effectiveness of our method to cope with punctual erroneous measurements.

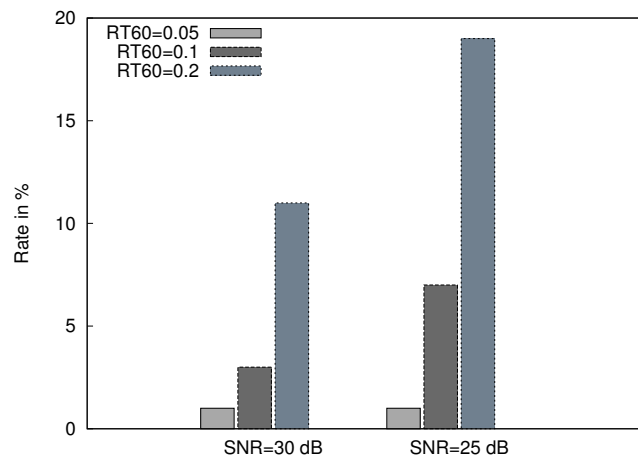


Figure 5.10: Average rate of missing/erroneous ITD considered for correctly achieved tasks for 30 and 25 dB of SNR

### 5.3 Multi-source tasks

In the second part of this chapter, we address scenarios where several sound sources are active in the scene. Here, the idea is to define a positioning task with respect to the ITDs extracted from each sound source. Under the far-field assumption that relates each ITD  $\tau_i$  to the DOA  $\alpha_i$  such positioning tasks can be referred to the field of beaconing or bearing-only homing/navigation. This kind of approach is bio-inspired and such process can be observed in nature through the localization system of bees [CC87]. One of the first applications about bearing-based robot control can be found in [LDW91] where the navigation of a mobile robot is governed by the beacon extracted by an array of sonar transducers. In the field of vision-based control several works also exploit beacons. For instance, [Cor03] proposes a planar visual servoing navigation system based on these features. This work inspired more recently [LPCS10] that develops a homing system based only on bearing measurements, without knowledge of the distance to the visual landmarks. This context is also exploited nowadays by the field of multi-robot localization [MPS05] or formation control of ground robots [FM02] or aerial vehicles [FMG<sup>+</sup>12] that is essentially based on bearing measurements. Such aforementioned tasks could also be performed by exploiting sound.

#### 5.3.1 Case of two sound sources

##### 5.3.1.1 Control scheme and task analysis

We consider now the configuration where two sound sources  $\mathbf{X}_{s_1}$  and  $\mathbf{X}_{s_2}$  are present in the acoustic scene. In this configuration the interaction matrix is obtained by

merely stacking (5.25) for each sound source:

$$\mathbf{L}_\alpha = \begin{bmatrix} \frac{\sin \alpha_1}{\ell_1} & -\frac{\cos \alpha_1}{\ell_1} & -1 \\ \frac{\sin \alpha_2}{\ell_2} & -\frac{\cos \alpha_2}{\ell_2} & -1 \end{bmatrix}, \quad (5.68)$$

which is equivalent to

$$\mathbf{L}_\tau = \begin{bmatrix} -\frac{\nu_1^2}{A\ell_1} & \frac{\tau_1\nu_1}{A\ell_1} & \nu_1 \\ -\frac{\nu_2^2}{A\ell_2} & \frac{\tau_2\nu_2}{A\ell_2} & \nu_2 \end{bmatrix} \quad (5.69)$$

with the ITDs  $\tau_1$  and  $\tau_2$  respectively related to  $\mathbf{X}_{\mathbf{s}_1}$  and  $\mathbf{X}_{\mathbf{s}_2}$ , and  $\nu_i = \sqrt{A^2 - \tau_i^2}$ . Similarly to the single-source task, we consider a control scheme based on the approximation of the latter interaction matrix  $\widehat{\mathbf{L}}_\tau$ .

Once again, the task related to the use of such a control scheme can be characterized from the virtual linkage approach applied on (5.68). This time  $\mathbf{S}^* \in \mathbb{R}^{3 \times 1}$  implies a class 1 virtual link:

$$\mathbf{S}^* = \begin{bmatrix} \ell_1 \cos \alpha_2 - \ell_2 \cos \alpha_1 \\ \ell_1 \sin \alpha_2 - \ell_2 \sin \alpha_1 \\ \sin(\alpha_1 - \alpha_2) \end{bmatrix}. \quad (5.70)$$

Geometrically  $\mathbf{S}_2^*$  refers to the circumscribed arc of circle characterized by  $\mathbf{X}_{\mathbf{s}_1}$ ,  $\mathbf{X}_{\mathbf{s}_2}$  and  $\mathbf{M}$ , as it can be proved by the inscribed angle theorem. Let  $\delta$  be  $\angle \mathbf{X}_{\mathbf{s}_1} \mathbf{M}^* \mathbf{X}_{\mathbf{s}_2}$  (i.e.  $\delta = \alpha_2^* - \alpha_1^*$ ) from a specified pose as it appears in Figure 5.11a. This theorem exposes that given  $\mathbf{X}_{\mathbf{s}_1}$ ,  $\mathbf{X}_{\mathbf{s}_2}$ , the set of points  $\mathbf{M}$  in the plane for which the angle  $\angle \mathbf{X}_{\mathbf{s}_1} \mathbf{M} \mathbf{X}_{\mathbf{s}_2}$  is equal to  $\delta$  is an arc belonging to the circumscribed circle of  $\mathbf{M}^* \mathbf{X}_{\mathbf{s}_1} \mathbf{X}_{\mathbf{s}_2}$ . Then by considering any point  $\mathbf{M}$  on this arc, from Proposition 2, it is guaranteed to find an orientation  $\theta_{\mathbf{M}}$  such that  $\alpha_1 = \alpha_1^*$  (or  $\alpha_2 = \alpha_2^*$ ). Knowing that  $\delta$  stays constant,  $\alpha_2 = \alpha_2^*$  (or  $\alpha_1 = \alpha_1^*$ ) is obtained from  $\delta = \alpha_2^* - \alpha_1^*$ . This result sets up the following proposition:

**Proposition 3** *For any position  $\mathbf{M}$  on the arc of the circumscribed circle defined by  $\mathbf{X}_{\mathbf{s}_1}$ ,  $\mathbf{X}_{\mathbf{s}_2}$  and  $\mathbf{M}^*$ , there exists one orientation  $\theta_{\mathbf{M}}$  of the microphones so that  $\alpha_1 = \alpha_1^*$  and  $\alpha_2 = \alpha_2^*$ .*

This proposition is valid for any source configuration but the situation where  $\mathbf{M}^*$ ,  $\mathbf{X}_{\mathbf{s}_1}$  and  $\mathbf{X}_{\mathbf{s}_2}$  are aligned. In this particular case where  $\alpha_1^* = \alpha_2^*$  we get:

$$\mathbf{S}_2^* = [\cos \alpha_1 \quad \sin \alpha_1 \quad 0]^\top. \quad (5.71)$$

As expected, we obtain a trajectory similar to the translation motion described for the case of a single source in (5.43), since only one angular data is available (see Figure 5.11b). For the stability, the condition  $\mathbf{L}_\tau \widehat{\mathbf{L}}_\tau^+ > 0$  is of course ensured when  $\widehat{\ell}_i = \ell_i$  since we have  $\mathbf{L}_\tau \widehat{\mathbf{L}}_\tau^+ = \mathbb{I}_2$  in that case. Since  $\ell_i$  is not available, a conventional method to approximate each  $\widehat{\ell}_i$  is to use the distances  $\ell_i^*$  to the sound source at a desired pose. In this case, it is well known that the system is locally asymptotically stable in the neighborhood of the desired pose [CH08]. In our case, with an infinite set of desired poses, a simple choice is to fix  $\widehat{\ell}_1 = \widehat{\ell}_2 = k$ , where  $k$  is an approximation of the distance to the sources at the desired pose.

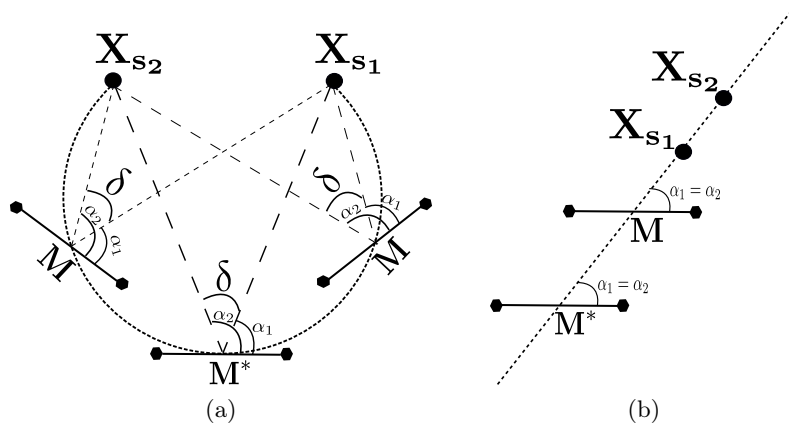


Figure 5.11: With two sound sources (a) several poses exist for given  $\tau_1^*$  and  $\tau_2^*$  on the circumscribed circle defined by the sound sources and the position of the microphones. When the sources are aligned with  $\mathbf{M}$  (b), the circle becomes a line directed by the sources locations.

### 5.3.1.2 Experimental results

From the robot configuration given in Figure 5.5, the control scheme used in this experiment is based on the Jacobian feature matrix extracted from (5.69) as

$$\widehat{\mathbf{J}}_{\boldsymbol{\tau}} = \begin{bmatrix} \frac{\tau_1 \nu_1}{A \widehat{\ell}_1} & \frac{A \widehat{\ell}_1 \nu_1 - D_x \nu_1^2}{A \widehat{\ell}_1} \\ \frac{\tau_2 \nu_2}{A \widehat{\ell}_2} & \frac{A \widehat{\ell}_2 \nu_2 - D_x \nu_2^2}{A \widehat{\ell}_2} \end{bmatrix}. \quad (5.72)$$

The control input  $\dot{\mathbf{q}} = (u, \omega)$  is then computed from

$$\dot{\mathbf{q}} = -\lambda \widehat{\mathbf{J}}_{\boldsymbol{\tau}}^+ (\boldsymbol{\tau} - \boldsymbol{\tau}^*), \quad (5.73)$$

in which  $\boldsymbol{\tau} = [\tau_1, \tau_2]^\top$  and  $\boldsymbol{\tau}^* = [\tau_1^*, \tau_2^*]^\top$ . The settings are exactly the same as the case of a single source task and are given in Table 5.2. The approximated distances are  $\widehat{\ell}_1 = \widehat{\ell}_2 = 1$  m. Furthermore the tracking routine described in Section 5.2.6.3 is used. Additionally to this tracking, a labelling of the retrieved ITDs is necessary when there are two sound sources. The goal is to associate each  $\tau_i$  to the desired  $\tau_i^*$  so that the task can be correctly completed. The labelling problem is trivial to solve in this configuration. If we consider the working space as the half plane in front of the microphones, the ordinality of  $\tau_i(t)$  and  $\tau_i^*$  is the same. Namely if  $\tau_1^* < \tau_2^*$  then  $\tau_1(t)$  should be lesser than  $\tau_2(t)$ . As shown in Figure 5.11a, it is obvious that each pose of the microphones in the environment can be characterized by a circumscribed circle on which the corresponding angles  $\alpha_1$  and  $\alpha_2$  have always the same ordinality.

Hence, in this experiment, besides the female speech we added a second sound source corresponding to a burst of white Gaussian noise of 25 ms followed by 25 ms of silence played in loop. This time, the objective is to reach a pose where  $\tau_1^* = -\tau_2^*$  with  $\alpha_1 = 50^\circ$ . From a pose fulfilling that condition, the system extracted  $\tau_1^*$  and

$\tau_2^*$  in the first step. Afterwards, starting from a pose around 3 meters away from the sources, the system automatically initialized and labelled  $\tau_1(t_0)$  and  $\tau_2(t_0)$ .

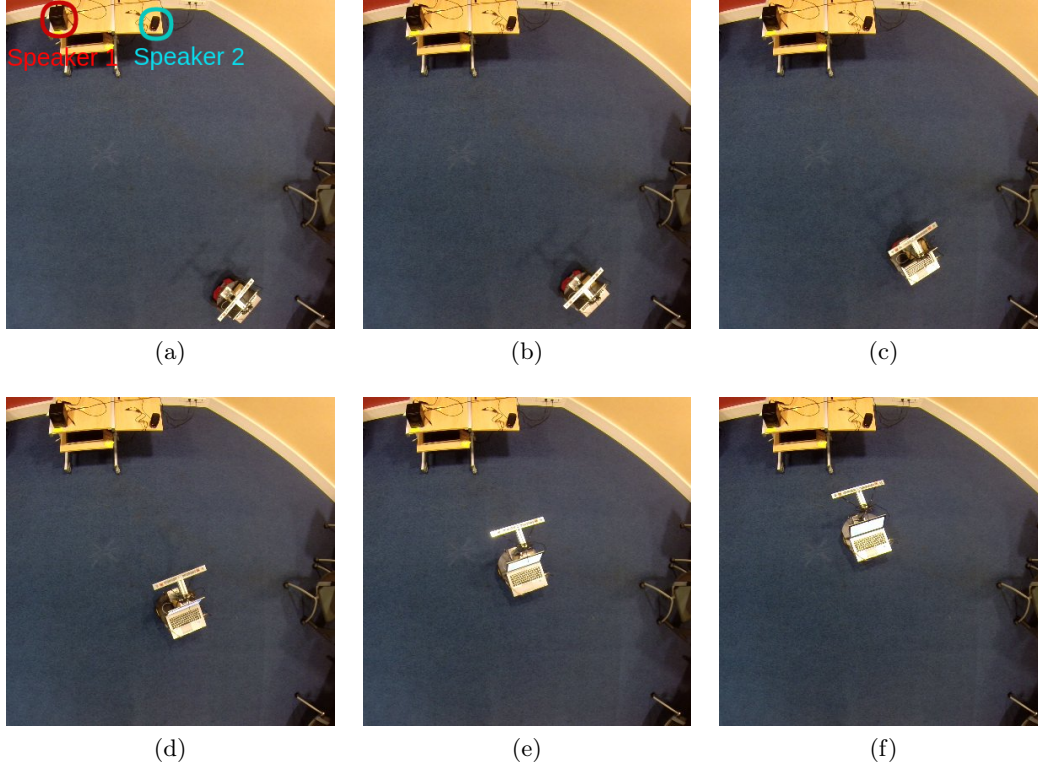


Figure 5.12: With two sound source, the robot heads towards a pose satisfying the given bearing conditions.

Figure 5.12 shows the completion of the task. Data from this experiment (see Figure 5.13) shows corrupted ITD estimations occurring during the robot motion which were coped with with the tracking routine. More precisely spurious ITDs were caused by echoes following the same variation as the actual ITDs. This could be expected since the echoes appear as virtual sound sources. Moreover the noise effect is still present with observation of peaks for  $\alpha = 90^\circ$  during most of the frames. Despite these poor observations, the error of the measured ITDs successfully converged to zero while the robot follows a straight and smooth trajectory towards the circular set of solutions to be reached, with a correct orientation.

However, compared to ITD-based tasks using a single source, this type of task is less robust. The robustness decreases notably because of the issue of tracking each source ITD. When considering intermittent sound sources, the tracking algorithm does not account for the number of active sources at the current frame. Hence, some plausible ITDs issued from the echoes of the active source could be associated to the ITD of the inactive sound source. For this kind of configuration, it is then necessary to have a more reliable tracking that takes into consideration the number of active sound sources. Otherwise, when considering continuous sound sources, the



performance of the positioning task is similar to the single source task.

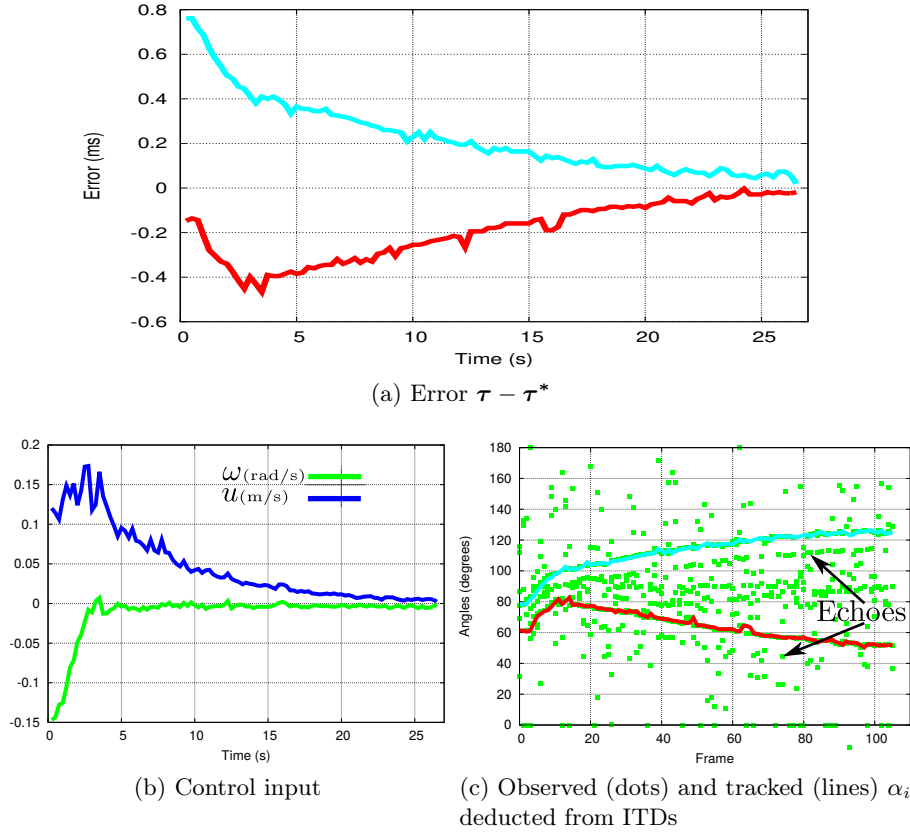


Figure 5.13: Experimental result of the ITD-based task with two sound sources illustrated in Figure 5.12.

### 5.3.2 Case of three sound sources and more

#### 5.3.2.1 Control scheme and task analysis

It is also possible to consider more sound sources. When considering three sound sources  $\mathbf{X}_{s_1}$ ,  $\mathbf{X}_{s_2}$  and  $\mathbf{X}_{s_3}$  the interaction matrix is generally full rank 3:

$$\mathbf{L}_\alpha = \begin{bmatrix} \frac{\sin \alpha_1}{l_1} & -\frac{\cos \alpha_1}{l_1} & -1 \\ \frac{\sin \alpha_2}{l_2} & -\frac{\cos \alpha_2}{l_2} & -1 \\ \frac{\sin \alpha_3}{l_3} & -\frac{\cos \alpha_3}{l_3} & -1 \end{bmatrix} \quad (5.74)$$

For the ITDs  $\tau_1$ ,  $\tau_2$  and  $\tau_3$ , related to each sound source, this interaction matrix becomes

$$\mathbf{L}_\tau = \begin{bmatrix} -\frac{\nu_1^2}{Al_1} & \frac{\tau_1 \nu_1}{Al_1} & \nu_1 \\ -\frac{\nu_2^2}{Al_2} & \frac{\tau_2 \nu_2}{Al_2} & \nu_2 \\ -\frac{\nu_3^2}{Al_3} & \frac{\tau_3 \nu_3}{Al_3} & \nu_3 \end{bmatrix}. \quad (5.75)$$



In most cases, this configuration implies that there exists only one pose where  $\mathbf{s} = \mathbf{s}^*$ . However, depending on the specified desired position or the sound sources configuration, a set of contiguous poses for which  $\mathbf{s} = \mathbf{s}^*$  may exist. Indeed, inspired by the case of two sound sources, a set can be expected if  $\mathbf{X}_{s_1} \mathbf{X}_{s_2} \mathbf{X}_{s_3}$  and  $\mathbf{M}$  design a concyclic quadrilateral. Wherever  $\mathbf{M}$  belongs to the corresponding circumscribed circle, an orientation  $\theta_{\mathbf{M}}$  exists such that  $\mathbf{s} = \mathbf{s}^*$  (see Figure 5.14b) and  $\mathbf{S}^*$  is designed by (5.70). This result can be demonstrated analytically by computing the determinant of the interaction matrix  $\mathbf{L}_{\alpha}$ :

$$|\mathbf{L}_{\alpha}| = \frac{\sin(\alpha_1 - \alpha_2)}{\ell_1 \ell_2} + \frac{\sin(\alpha_2 - \alpha_3)}{\ell_2 \ell_3} + \frac{\sin(\alpha_3 - \alpha_1)}{\ell_1 \ell_3}. \quad (5.76)$$

Besides, from the inscribed angle theorem and the sinus law, it is possible to say that:

$$\frac{\|X_{s_3} X_{s_1}\|}{\sin(\alpha_3 - \alpha_1)} = \frac{\|X_{s_3} X_{s_2}\|}{\sin(\alpha_3 - \alpha_2)} = \frac{\|X_{s_2} X_{s_1}\|}{\sin(\alpha_2 - \alpha_1)}. \quad (5.77)$$

From now on, the determinant of the interaction matrix is equal to:

$$|\mathbf{L}_{\alpha}| = \frac{\sin(\alpha_3 - \alpha_1)(\ell_2 - \frac{\|X_{s_2} X_{s_1}\|}{\|X_{s_3} X_{s_1}\|} \ell_3 - \frac{\|X_{s_3} X_{s_2}\|}{\|X_{s_3} X_{s_1}\|} \ell_1)}{\ell_1 \ell_2 \ell_3}. \quad (5.78)$$

Furthermore, an other property of the cyclic quadrilateral is that the product of the diagonals equals to the sum of the products of opposite sides:

$$\ell_2 \|X_{s_3} X_{s_1}\| = \ell_1 \|X_{s_3} X_{s_2}\| + \ell_3 \|X_{s_2} X_{s_1}\| \quad (5.79)$$

By combining equation (5.78) and (5.79), it appears that  $|\mathbf{L}_{\alpha}| = 0$ . As a result, there is a rank loss of the interaction matrix, and this system is similar to the case of two sound sources. There exists other configurations also leading to a rank loss of the matrix. By analyzing (5.76),  $|\mathbf{L}_{\alpha}|$  is null when:

- $\alpha_1 = \alpha_2 = \alpha_3$ . In that configuration, the sound sources are aligned in the microphones axis as  $\overrightarrow{\mathbf{M} \mathbf{X}_{s_1}} = k_1 \overrightarrow{\mathbf{M} \mathbf{X}_{s_2}} = k_2 \overrightarrow{\mathbf{M} \mathbf{X}_{s_3}}$ , which corresponds to the configuration described by (5.71). But we can also notice that an infinitesimal motion defined by  $v_x, v_y$  or  $\omega_z$  different from sound direction will cancel this singularity.
- $\alpha_1 = \alpha_2$ , then we have  $|\mathbf{L}_{\alpha}| = \frac{\sin \alpha_{12} \cos \alpha_3 - \cos \alpha_{12} \sin \alpha_3}{\ell_2 \ell_3} + \frac{\cos \alpha_{12} \sin \alpha_3 - \sin \alpha_{12} \cos \alpha_3}{\ell_1 \ell_3}$ . As soon as  $\ell_1 = \ell_2$ , the determinant is equal to zero. Concretely, if at least two out of the three sources are juxtaposed, we are in the similar case of only two sound sources.

Apart from the aforementioned configurations, there is a unique solution for a task based on three sound sources (see Fig.5.14a). Geometrically, this result is obtained by considering the three circumscribed arcs of circle defined by  $\mathbf{X}_{s_1} \mathbf{X}_{s_2} \mathbf{M}$ ,  $\mathbf{X}_{s_2} \mathbf{X}_{s_3} \mathbf{M}$  and  $\mathbf{X}_{s_1} \mathbf{X}_{s_3} \mathbf{M}$ . According to Proposition 3,  $\mathbf{M}$  must belong to these three arcs to ensure  $\alpha_1 = \alpha_1^*$ ,  $\alpha_2 = \alpha_2^*$  and  $\alpha_3 = \alpha_3^*$ . As shown in Figure 5.14b, these arcs intersect in only one position when considering a nominal configuration. From this result the following proposition can be set:

**Proposition 4** *In presence of three sound sources in a non-concyclic or non-degenerate configuration with  $\mathbf{M}^*$ , there exists only one pose  $\mathbf{M}$  of the microphones so that  $\alpha_1 = \alpha_1^*$ ,  $\alpha_2 = \alpha_2^*$  and  $\alpha_3 = \alpha_3^*$ .*

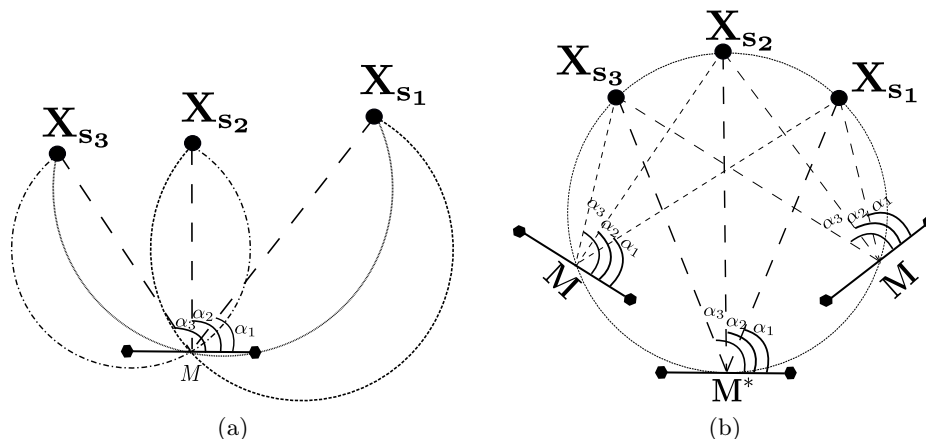


Figure 5.14: With three sound sources (a) a unique pose generally satisfies the conditions given by  $\tau_1^*$ ,  $\tau_2^*$  and  $\tau_3^*$ . When the sources and the microphones position  $\mathbf{M}$  design a concyclic quadrilateral (b), there is set of solution equivalent to the configuration with two sound sources

Eventually, the circumscribed circle built by the sound sources is then a zone where the interaction matrix is singular and thus the controller is not stable. Indeed independently of the approximation  $\widehat{\mathbf{L}}_\tau$ , the condition  $\mathbf{L}_\tau \widehat{\mathbf{L}}_\tau^{-1} > 0$  cannot be ensured since at this position  $|\mathbf{L}_\tau| = 0$ . We denote this zone as a singularity circle. A parallel can be emphasized with IBVS region of instability, where the circumscribed circle is the planar equivalent of the singular cylinder [MR<sup>+</sup>93].

### 5.3.2.2 Using more than three sound sources

Using more than three sound sources can be beneficial by giving more robustness to the control law. The demonstration developed for three sound sources remains applicable and a single pose exists such that  $\mathbf{s} = \mathbf{s}^*$ . Redundant data may ensure the stability of the control scheme when crossing the singularity circle that does not exist anymore. Indeed the singularity of the interaction matrix is cancelled by a fourth sound source positioned in a non-concyclic configuration with the three first ones. On the other hand, only the local asymptotic stability is verified under the condition  $\widehat{\mathbf{L}}_\tau^+ \mathbf{L}_\tau > 0$  (see Chapter 3.2.3), which is ensured when  $\widehat{\ell}_i = \ell_i$  since  $\widehat{\mathbf{L}}_\tau^+ \mathbf{L}_\tau = \mathbb{I}_3$ . Thence with three sound sources in a non-singular configuration or with at least four sources in any other case, all the three robot DOF are constrained. A homing task can be performed without an exact knowledge of the position of the sound sources (an approximation  $\widehat{\ell}_i$  is sufficient). The realization of such task corresponds to bearing-only homing system.

Nonetheless, from an acoustic point of view, achieving such task remains extremely challenging when considering wideband and intermittent signal such as speech. In indoor environment, reverberation, sources (in)activity complicate the tracking and the identification of each ITD. Moreover, only few techniques are able to detect robustly such a number of sound sources. The survey conducted in [BOV12] shows that GCC-PHAT has, at best, a recall rate of 0.7 for 3 sound sources. From this result, it can be inferred that it is likely that all the sources will not be detected in the sound mixture. For all these reasons, we think that such a method should be particularly suitable for narrowband acoustic landmarks, that simplify the aforementioned problems. Each landmark is then detected on a limited and disjoint frequency range, which eases the detection and identification of the ITD. For instance under this hypothesis, in [BSLF16] the authors propose a follower-leader control scheme based on bearing measurements recorded by 4 microphones. In this configuration, the ITD-based positioning task could be fully exploited for multi-robot systems navigation or formation control.

### 5.3.2.3 Numerical evaluation

The developed method was tested through simulations in Roomsimove [BOV12]. We designed a room of 4.55 m  $\times$  3.55 m  $\times$  2.5 m in which three sound sources  $\mathbf{X}_{s_1}$ ,  $\mathbf{X}_{s_2}$  and  $\mathbf{X}_{s_3}$  were positioned as they appear in Figure 5.15. Here, we simulated a holonomic robot that is controlled by the 3 DOF of the azimuth plane  $v_x$ ,  $v_y$  and  $\omega_z$ . First, each of the sound sources emitted in succession so that they could be identified uniquely. Thereafter, the ITDs  $\tau_i$  are estimated from the sound mixtures generated by the three sound sources. In the first two simulations described above we considered the real interaction matrix so that  $\widehat{\mathbf{L}}_{\boldsymbol{\tau}}^{-1} = \mathbf{L}_{\boldsymbol{\tau}}^{-1}$  assuming that each distance  $\ell_i$  is known. The effect of approximating the interaction matrix is discussed afterwards. The control scheme of the task is given by

$$\mathbf{u}_M = -\lambda \mathbf{L}_{\boldsymbol{\tau}}^{-1} (\boldsymbol{\tau} - \boldsymbol{\tau}^*), \quad (5.80)$$

where  $\mathbf{L}_{\boldsymbol{\tau}}$  is defined by (5.75).

We simulated first a homing task with an initial pose defined by  $\mathbf{M}(2, 0.75)$  and  $\theta_M = -5^\circ$  and a home pose defined by  $\mathbf{M}^*(3.1, 1.75)$  and  $\theta_{M^*} = 10^\circ$ . That home pose corresponds to  $\boldsymbol{\alpha}^* = (31^\circ, 85^\circ, 118^\circ)$  and was located outside of the circumscribed circle. In addition, a moderate reverberation was added with a reverberation rate  $RT_{60} \approx 75$  ms and with background white noise that guarantees an SNR of 25 dB. The result of this simulation is shown on Figure 5.15. Despite the uncertainties on the ITD measurements caused by noise and reverberation, the system converged successfully to the home pose with an exponential decrease of the error. Moreover, since the singularity circle was not crossed during the task, the stability of the control law was ensured. As for simulation data, despite the jittery behavior in the velocities curves caused by ITDs estimation, the trajectory remains smooth.

A second scenario was conducted including 99 trials on different reverberation times and uniformly distributed starting poses. With the *front-back ambiguity* implied by the use of only two microphones,  $\theta_M$  was chosen so that the sound sources

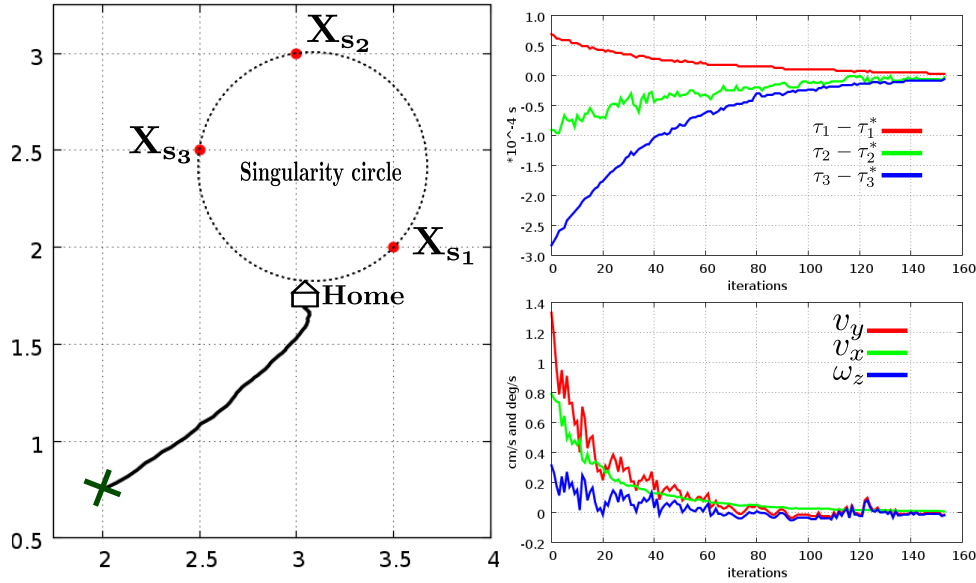


Figure 5.15: A typical homing task with moderate reverberation  $RT_{60} \approx 75$  ms

were on the same side of the microphones at the initial pose. Thus the orientation  $\theta_M$  varied from  $-30^\circ$  to  $30^\circ$  depending on the starting pose. With a home pose at  $\mathbf{M}(2.25, 1)$  and  $\theta_{M^*} = -5^\circ$ , the results summed up in Figure 5.16 show that with no reverberation the system always converges to that pose. This result confirms the relevance and the suitability of the auditory modeling. However, we distinguished two failure areas in presence of reverberation. The first one was located on the alignment of  $\mathbf{X}_{S2}$  and  $\mathbf{X}_{S3}$  and mainly occurs when the reverberation increases. In this area, the main difficulty is to identify each measurement. With variation in the estimation and outlier (sound echoes) values, a wrong ITD could be associated to a sound source. Likewise, a feature loss was observed on the right extremity of the test area since one of the ITDs was not correctly estimated. Consequently the system could fail to converge in those two areas especially with high reverberation.

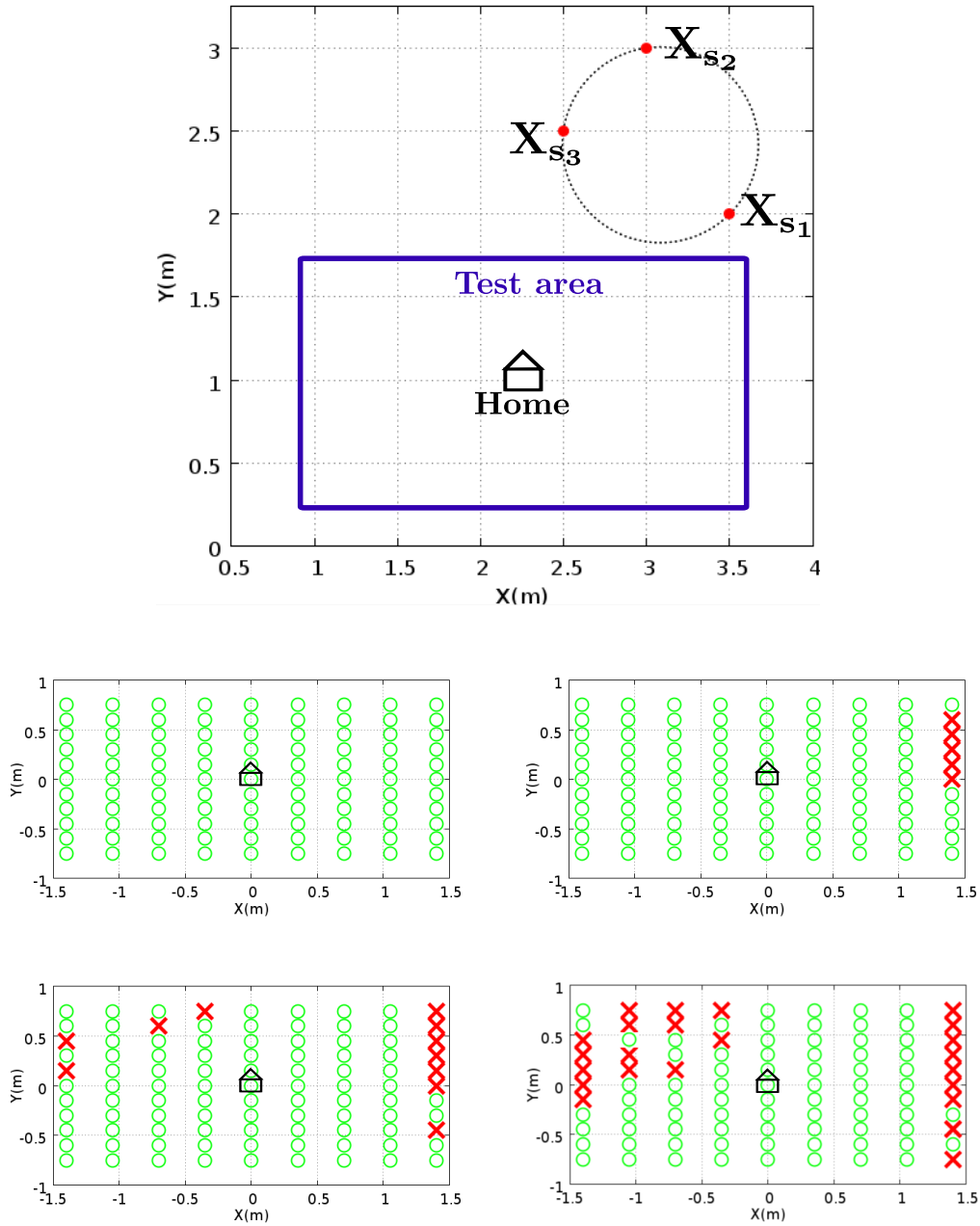


Figure 5.16: Homing task success in function of reverberation and initial pose: success(**O**) and failure (**X**) with from top-left to bottom-right with reverberation rate  $RT_{60} = \{0, 0.02, 0.05, 0.075\}$ .

### 5.3.2.4 Approximation of the interaction matrix

Several ways exist to approximate  $\mathbf{L}_\tau$  by  $\widehat{\mathbf{L}}_\tau$  and these different strategies affect the behavior and the stability of the system as shown in [MMR10]. Since the distances  $\ell_i$  are unknown in practice, a first strategy consists in using  $\ell_i^*$ , the distances at the desired pose, instead. As illustrated by Figure 5.17,  $\widehat{\mathbf{L}}_\tau = \mathbf{L}_{\tau(\tau, \ell^*)}$  gives a straighter and shorter trajectory. In addition, the desired measurements  $\tau^*$  can be used in the interaction matrix. In that case  $\widehat{\mathbf{L}}_\tau = \mathbf{L}_{\tau(\tau^*, \ell^*)}$  leads to a longer and unexpected trajectory but still converges to the home pose. These two solutions solve the problem of the unknown distance to the sound source. Moreover  $\widehat{\mathbf{L}}_\tau = \mathbf{L}_{\tau(\tau^*, \ell^*)}$  is constant and needs to be computed only once. A last strategy considers the combination of the real and the constant interaction matrix leading to  $\widehat{\mathbf{L}}_\tau = (\mathbf{L}_{\tau(\tau, \ell)} + \mathbf{L}_{\tau(\tau^*, \ell^*)})/2$ . The corresponding trajectory is improved compared to the constant interaction matrix, with the influence of  $\mathbf{L}_{\tau(\tau, \ell)}$ .

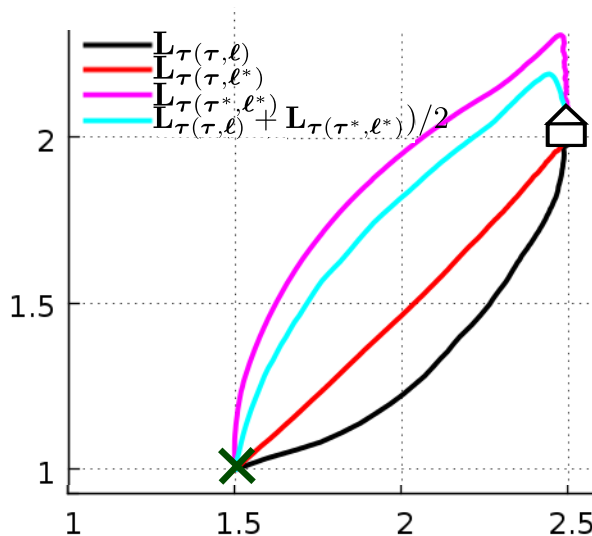


Figure 5.17: Homing trajectories with different interaction matrix approximations

## 5.4 Conclusion

In this chapter, we proposed control schemes based on either a single ITD or several ITDs in the configuration of multiple sound sources. These rely on the geometrical representation of ITDs. A given ITD expresses the sound source location through an hyperbola function. This hyperbola leads to two types of ambiguities that are the *left-right ambiguity* and the *front-back ambiguity*. The first ambiguity is solved from the sign of the measured ITD, but the *front-back ambiguity* remains unsolved, similarly to most binaural sound localization algorithms. Under far-field conditions, the ITD is directly related to the sound source direction, which is a common assumption in robot audition. This assumption corresponds geometrically to the branch of the hyperbola reaching its asymptotic value. However as the sound source is getting closer to

microphones, namely in the near-field, this assumption does not hold anymore, since the hyperbola and its asymptotes are not merged. This result explains that most of the localization techniques based on ITD only consider far-field configuration for accuracy purpose.

In a second phase, we focused more particularly on positioning tasks based on a single sound source. We then defined two interaction matrices that link the variation of a given ITD to the microphones motion. From the virtual linkage approach, the positioning task was specified as orienting the robot with respect to the source direction. Controllers based on these matrices were then designed while respecting the stability conditions related to the approximations used. Similar conditions of stability were extracted from both ITD interaction matrices. These conditions are related to the *left-right ambiguity* (solved by the measurement of the ITD), and the *front-back ambiguity*. As expected from Chapter 3.1.1, the sound source localization is not required. The stability conditions are not constraining for both of these controllers. There exists an infinite number of solutions and the stability can easily be ensured by setting a positive source distance in the first matrix or by ensuring that the ambiguities expressed above are solved, in the case of the second matrix. Thus, by ensuring these conditions, experimental results validated the robustness of our approach. First, we demonstrated by analyzing the two interaction matrices, that our approach is not influenced by the distance to the given sound source. More explicitly, the positioning task can be performed in the near-field or far-field without restriction, despite rough approximations of the interaction matrices. This result is a clear advantage in comparison to binaural localization techniques that are generally devoted to only one type of configuration. Furthermore, experiments conducted on a mobile robot in a real environment emphasized the robustness of the controllers. The robot is able to face and track an intermittent sound source that is moving. For this purpose, a basic tracking process was introduced. The tracking exploits the knowledge of the interaction matrix in order to predict the evolution of the tracked ITD. A more exhaustive evaluation also confirmed that, thanks to this tracking, the control framework could be robust to real-world conditions since the controller can cope with false and missing measurements. Punctual errors and approximations in ITD tracking do not compromise the achievement of the positioning task.

In the last segment, we extended the latter control scheme to the case of multiple sound sources, that are used as acoustic landmarks for robot navigation. When considering several sound sources our approach is related to bearing-only navigation. Several positioning tasks can be achieved controlling up to three DOF in the azimuth plane. In a real-world experiment with two sound sources, we have been able to correctly position a robot with respect to these sources. The robot successfully reached a pose that belongs to a circumscribed circle shaped by the sound sources. This circle characterizes the desired bearing conditions. A single pose of the microphones can be ensured with at least three sound sources as validated by simulations results. However, multi-source configurations require dedicated tracking algorithm in order to cope with adverse situations where competing and intermittent sound sources are present. This tracking is certainly the most critical aspect of this control framework. The main issue concerns the tracking and identification of the ITDs associated to

the actual source(s) rather than obtaining accurate intrinsic values.

Globally, the results exhibited in this chapter show robustness to inaccurate ITDs measurements. This property is confirmed in the next chapter, that demonstrates that the ILD framework and the current ITD framework could be freely applied on humanoid robots where the head scattering effect (*i.e.*, HRTFs) substantially modifies these cues.





## Chapter 6

# Application to humanoid robots

The last chapter of this thesis is about the context of humanoid robots. Until now, AS paradigm has been applied in a free-field context. Free-field context greatly eases auditory perception although real-world conditions are considered in this work. Nonetheless, all along the thread of this thesis several evidence showed the flexibility of AS. First the exact sound location is not needed to perform positioning tasks. The interaction matrices governing robots motions are actually using rough approximated parameters while preserving stability properties. Furthermore, approximations of acoustic models are also well tolerated. From a far-field model, positioning tasks in the near-field can be performed with the same accuracy as with a correct modeling. Eventually, positioning tasks can be performed from cues (ILDs) that cannot be directly related to a precise source location.

Thus, evidence converges towards the idea that accurate interaural cues (with respect to a ground truth source location) are not crucial to control robots motions. This idea let us foresee a potential applicability of AS on humanoid robots. The main difficulty arising when considering humanoid robots comes from the required accurate modeling of HRTFs, and acoustic conditions, in order to relate interaural cues to a sound location. However, since our approach is not attached to localize sound sources, thanks to the robustness of the control scheme, intuitively it can be envisaged that HRTFs modeling is not required for AS. This chapter aims to validate the latter idea. No new theoretical concepts are introduced here. We are rather interested in the application of the framework detailed in the previous chapters to the context of humanoid robots.

This chapter is then simply divided into two sections. In the first section, we briefly recall the context and challenges related to binaural hearing for humanoid robots. Following this review we give a comprehensive study on how the framework developed in Chapters 4 and 5 could be applied to humanoid robots. Thereafter, in the second section, experimental studies on two types of humanoid robots validate the versatility of the AS paradigm. All these experiments are achieved with a free-field modeling in a real world environment.

## 6.1 Versatility of aural servo paradigm

Binaural localization on humanoid robots is the closest configuration to biological auditory systems, but at the same time probably the most challenging configuration for robot audition. As explained in Chapter 2, in this configuration the combination of acoustic perturbations (reverberation, noise) and scattering effects induced by the head creates challenging conditions to be resolved for sound localization. Very few works, in view of the size of robot audition community, tried to address sound localization in this context. So far, the approaches considered to address this configuration are threefold. The first set of solutions are based on the knowledge of HRTFs. Pre-measured HRTFs are used to infer the sound location. This kind of approaches is generally performed under anechoic/controlled conditions since the influence of room acoustics substantially modifies measurements. A second path consists in modeling the scattering effects. For unknown robot HRTFs, the scattering theory, assuming a spherical rigid head is by far the most used model in robot audition. However this model does not account for acoustic conditions similarly to HRTF-based methods. Hence, these approaches suffer from the same limitations. The last type of approaches, based on learning methods, provides until now the best performances in realistic environment. But these methods are room/robot-specific. The learning process cannot be generalized to every kind of robots or acoustic configurations. Learning all kinds of acoustic environments is not affordable due to the high variability and dynamic nature of acoustic conditions. Currently, among the current projects related to robot audition only TWO!EARS (see Chapter 1.3) is addressing this challenging configuration.

From an AS perspective, in a first part, we focus on the question on how such approach could be adapted on humanoid robots. As stated in the scenario developed in Chapter 3, the core advantage of AS is certainly the fact that our approach is mainly based on the variation of auditory features. As long as the variation of the selected features is consistent with robot motion, the desired task can be correctly achieved. This characteristic has been exemplified in Chapter 5 where we showed that the control scheme could be applied independently in the near-field or far-field despite approximate acoustic models. Similarly in Chapter 4 we demonstrated that a robot could face a sound source despite ILD estimations that could not be interpreted as an azimuth angle. Until now, in robot audition and more specifically in sound source localization, this kind of approach relying on the variation of low-level features has not been fully exploited. As illustrated in the experiments performed all along this thesis, exploiting such characteristic increases the robustness to acoustic conditions. The variation of auditory cues could also be robust to the individual variability of sound cues perception.

In order to assess the latter statement, we conducted a set of experiments based on ILD features. The experiments consist in observing the variation of a given ILD with respect to the source motions from different robots. Let us first introduce the robots used in this experiment, that are respectively *Romeo* and *Pepper* from Softbank Robotics (see Figure 6.1). *Romeo* is a 1.4 m tall robot, equipped with four microphones and two pinnae. These microphones are embedded inside the robot

head as represented in Figure 6.2. *Pepper* is a wheeled holonomic robot that is 1.2 m tall, also equipped with four microphones on the top of its head. However, unlike *Romeo*, this robot does not have pinnae. The HRTFs of both robots are unknown and are certainly quite different to each other due to their dissimilar structure and microphones topology. For both robots, we considered only two microphones  $\mathbf{M}_1$

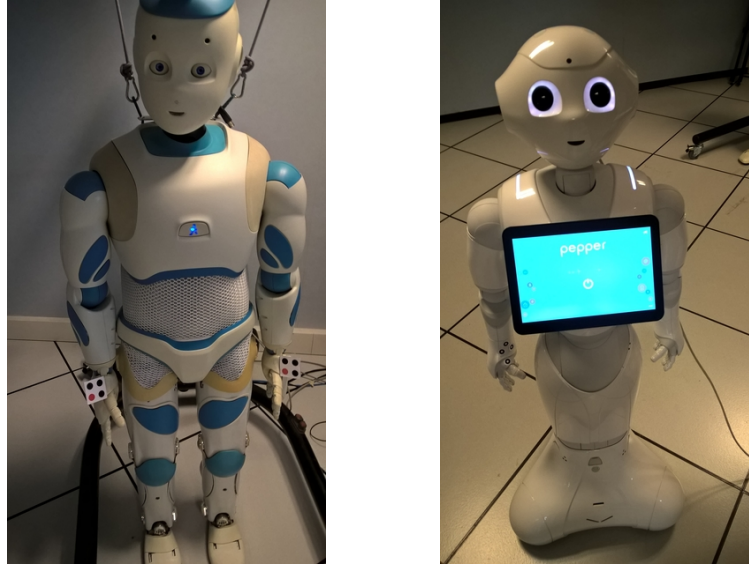


Figure 6.1: The two robots *Romeo* (right) and *Pepper* (left) used for measuring the dynamics of ILD cues

and  $\mathbf{M}_2$  separated by a distance  $d$ , as depicted in Figure 6.2. The inter-microphones distances are respectively  $d_{Pepper} \approx 0.07$  m and  $d_{Romeo} \approx 0.12$  m. For the experiments, a white Gaussian noise is continuously emitted from a loudspeaker. This loudspeaker is moved in a circular motion around robots head (*i.e.*, constant distance  $\ell$ ) so that the DOA  $\alpha$  varies from  $0^\circ$  to  $180^\circ$ . The ILD  $\rho$  computed from the measurements of the microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$  is measured all along the motion of the loudspeaker.

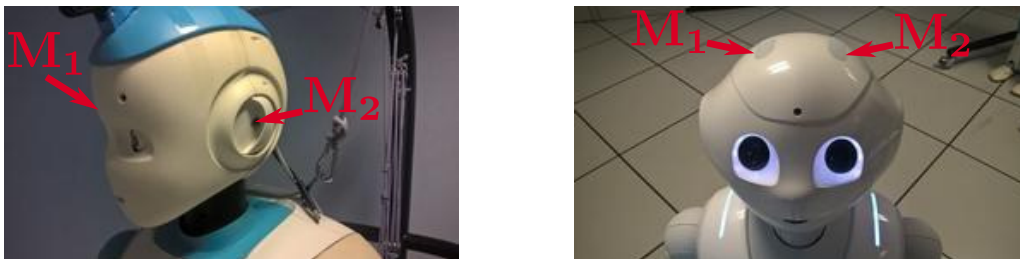


Figure 6.2: Auditory system of *Romeo* and *Pepper*

These experiments are conducted in a highly reverberant environment with  $RT_{60} >$

1 s. Furthermore, the same task is conducted in simulation, however, in anechoic free-field conditions, with microphones separated by  $d_f = 0.3$  m. The latter simulation serves as a baseline, for comparison with ideal conditions. The results are given in Figure 6.3 where the absolute value of  $\rho_{dB} = 20 \log_{10}(\rho)$  is plotted in order to facilitate the analysis.

First, it can be noticed that all configurations share the same "V" shape that reflects the symmetry of ILDs for sources at opposite locations with respect to interaural axis midpoint. An ILD value is maximal for eccentric positions of a given source, while its minimum value ( $\approx 0$  dB) is reached when this source is in the front. This property, shared by all configurations, is certainly the most interesting result for our approach. Despite different robot structures and acoustic conditions, the ILD variation is preserved. In contrast, the intrinsic values of  $\rho_{dB}$  are drastically changed depending on the auditory setup considered. Any approach based on localization would need to model or learn the influence of HRTFs on interaural cues. Hence, this kind of experiment confirms that independently of the chosen robot platform, the variation of such acoustic feature remains consistent. The adverse experimental conditions ( $RT_{60} > 1$  s), in the case of *Romeo* and *Pepper*, do not influence this property. The shape is the same for anechoic conditions and for both robots, despite a high level of reverberation. As a result, one can emphasize the robustness of our approach for real world conditions and more specifically to reverberation.

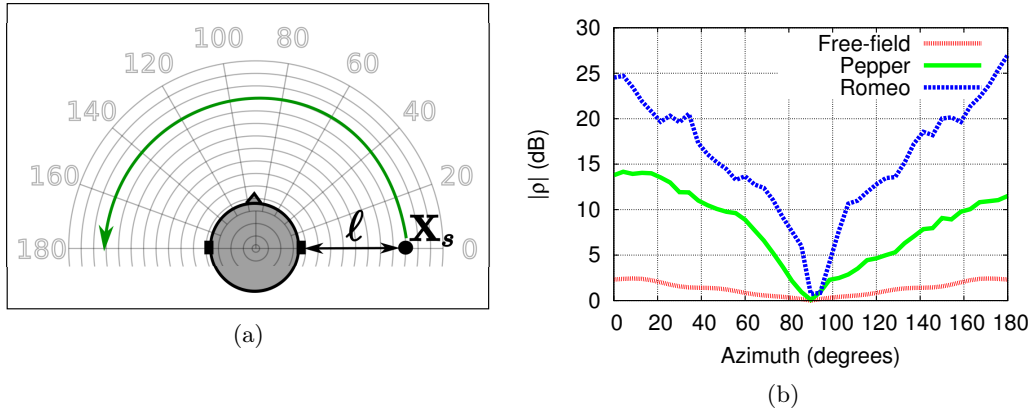


Figure 6.3:  $|\rho_{db}|$  measurements for a source at  $\ell = 0.8$  m, for *Romeo*, *Pepper* obtained in a real environment and for free-field microphones through simulation in anechoic conditions.

The details of the plotting in Figure 6.3 reveal some additional aspects of sound perception. Among the three configurations tested, the "V" shape is more pronounced on *Romeo*. For this robot, ILD variation is abrupt, while in free-field conditions this variation is more gradual. The head shadowing effect increases the variation of ILDs as detailed in Chapter 2. The results of *Pepper* are in-between, but far more pronounced than for free-field conditions. The difference between *Romeo* and *Pepper* can be explained by the different microphones topologies. The inter-microphone

distance is wider for *Romeo* and the microphones are located on the lateral position of the head. The sound attenuation is then more pronounced compared to *Pepper* which is endowed with microphones on the top of its head. From these results it can be inferred the following conclusion: head-mounted systems are better configurations than free-field systems for our approach. The head shadowing effect enhances ILD variation and any minimal variation of ILD (source motion) would be more "perceptible". As a consequence, the far-field limitations of ILD-based positioning tasks, described in Chapter 4, should be attenuated on humanoid robots. Furthermore, with enhanced ILDs, the effect of reverberation is reduced, especially for far-field conditions. This outcome is opposed with the commonly acknowledged idea in robot audition which states that free-field configurations are easier to process.

Hence, from this simple evaluation, we deduce that the experiments performed in Chapter 4 can be performed on humanoid robots, with at least the same accuracy and without any additional modeling. Tasks consisting in facing a sound source could be performed on humanoid robots with a free-field model. It should also be noted that the same kind of results can be obtained when considering ITD cues. In this case, the minimum absolute ITD value should be obtained when facing the sound source, while eccentric positions leads to higher absolute ITD value. Therefore, the experiments depicted in Chapter 5 are also applicable on *Romeo* and *Pepper*. The experimental results in the following sections support these conjectures.

## 6.2 Experimental validation

### 6.2.1 Gaze control: facing a sound source

As a first case of study, this set of experiments consists in orienting the head of a robot towards a sound source. Despite using humanoid robots, the following experiments are based on free-field assumptions. More exactly we consider the modeling expressed in Chapter 4.1 on page 80 for the ILD case and in Chapter 5.1 on page 112 for the ITD case. The control inputs are given by

$$\dot{q}_{itd} = -\lambda \frac{1}{\sqrt{\left(\frac{d}{c}\right)^2 - \tau^2}} (\tau - \tau^*) \quad (6.1)$$

when controlling the robot through the ITD  $\tau$ . From the ILD  $\rho$  the control input is given by

$$\dot{q}_{ild} = -\lambda \frac{\ell^2 + \frac{d^2}{4} - d\hat{x}_s}{\hat{y}_s d(\rho + 1)} (\rho - \rho^*). \quad (6.2)$$

In these two control schemes, only the angular velocity  $\omega$ , that sets the orientation of the robot head, is controlled.

The experimental setup consists in a loudspeaker continuously emitting a white Gaussian noise. The room acoustics corresponds to the conditions described in the previous section. The room has a high reverberation rate  $RT_{60} > 1$  s and typical diffuse noises, related to the ventilation or computer fans, are present in the scene. We consider beforehand ILD-based positioning task with the control input given by

(6.2). For this task, we set  $\rho^* = 1$ . Furthermore, the approximated parameters appearing in the control scheme are set to  $\hat{x}_s = 1 \times \text{sign}(\rho - 1)$  and  $\hat{y}_s = 1$ . The first experiment is carried on *Romeo*, with the use of two microphones  $\mathbf{M}_1$  and  $\mathbf{M}_2$  as depicted in Figure 6.2. No knowledge of HRTFs nor modeling of the scattering effect of the robot, are considered in the following results presented in Figure 6.4. As predicted in Section 6.1, the head turning task is correctly achieved despite a

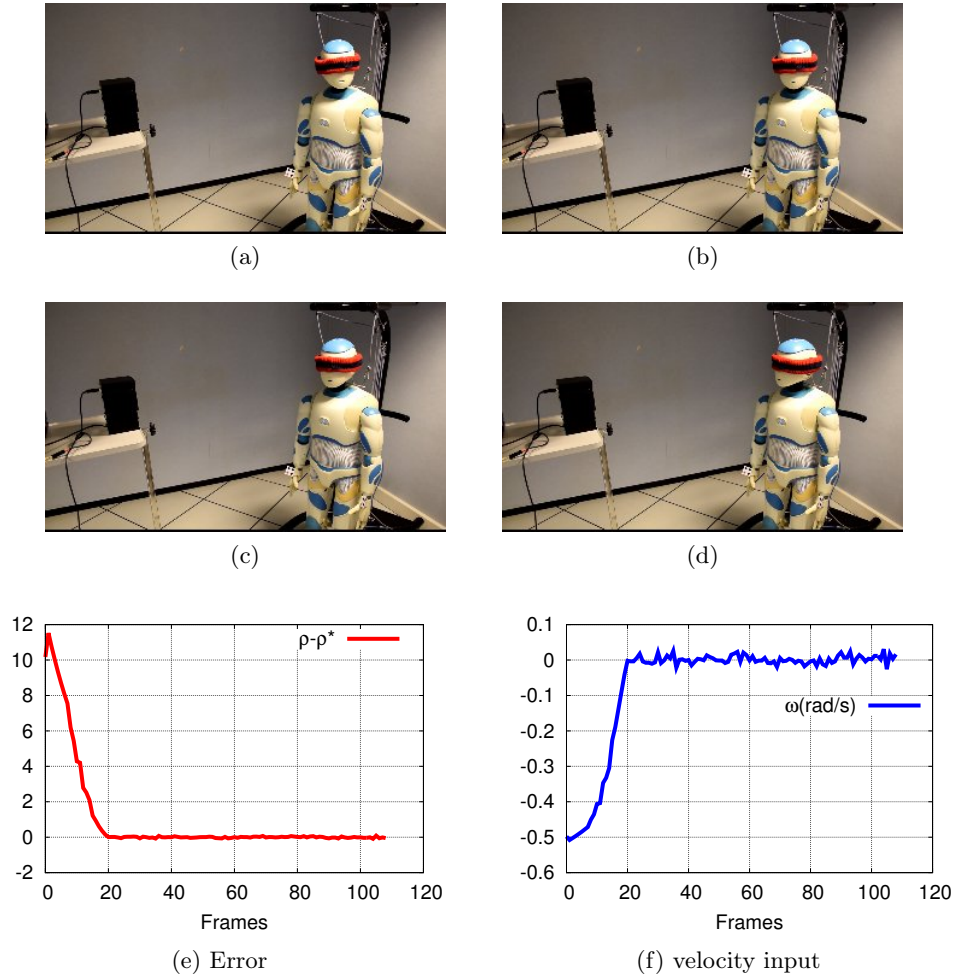


Figure 6.4: Head turning task from initial pose in (a) to final pose in (d) using ILDCues on *Romeo*.

model based on free-field conditions. The error curve follows an exponential decrease pattern while the robot accurately faces the sound source. This experiment has also been performed on Pepper in Figure 6.5. The same control scheme was used. Although the curves are jittery for both experiments when reaching the desired pose, the task is correctly achieved in both cases. Filtering strategies could also be applied on the measurements in order to smooth robots motions. These experiments confirm the versatility of AS that can be applied to different humanoid robots without any



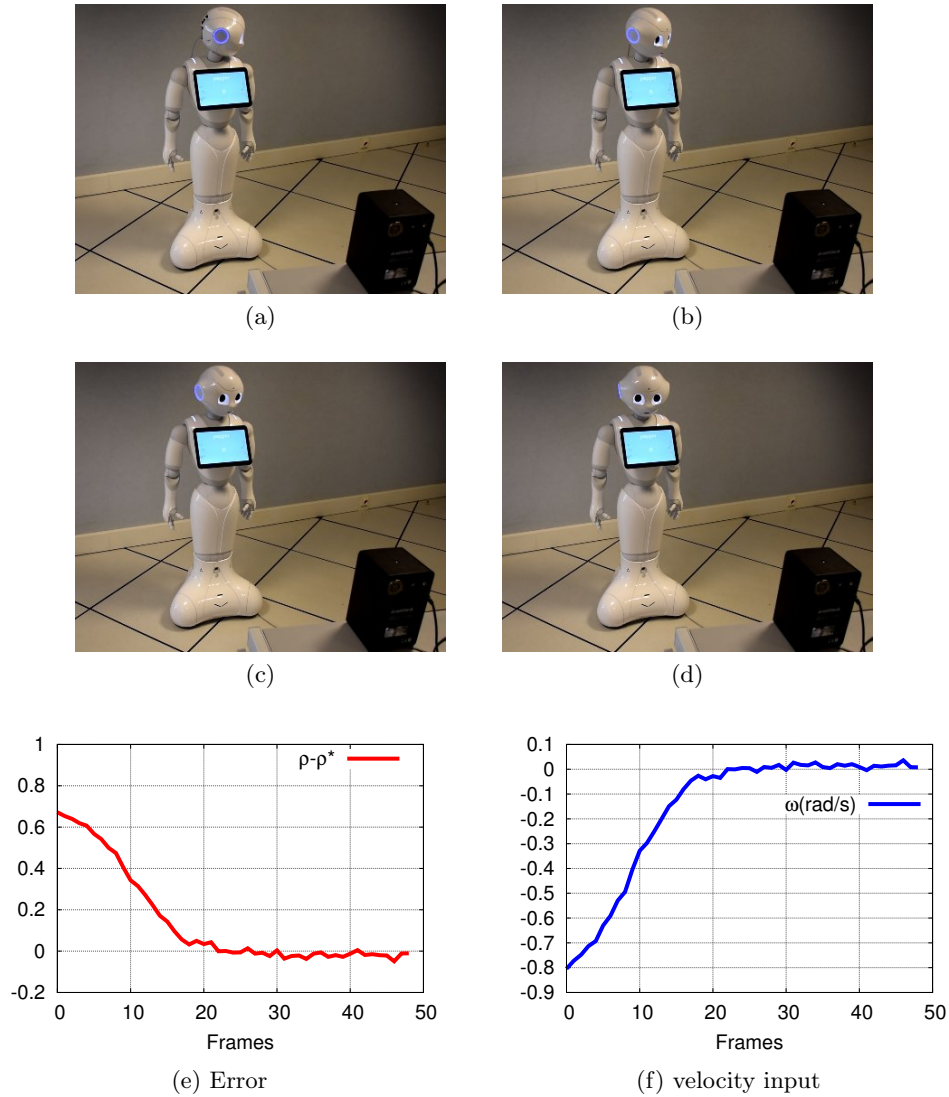


Figure 6.5: Head turning task from initial pose in (a) to final pose in (d) using ILD cues on *Pepper*.

additional modeling.

A second set of experiments was conducted, by using ITD measurements as input features for the positioning task. In this case, the control scheme for controlling robot motion is given by (6.1). This experiment was performed on *Romeo* with a loudspeaker emitting a continuous speech signal. Two external microphones were used instead of the embedded ones, because of the high level of internal noise in this robot. GCC-PHAT algorithm is not particularly suitable for such level of noise. Despite the use of external microphones, the head scattering effect and the high level of reverberation in the scene are still influencing ITD measurements. Once again, the task is correctly achieved (see Figure 6.6) while using a non-stationary signal.



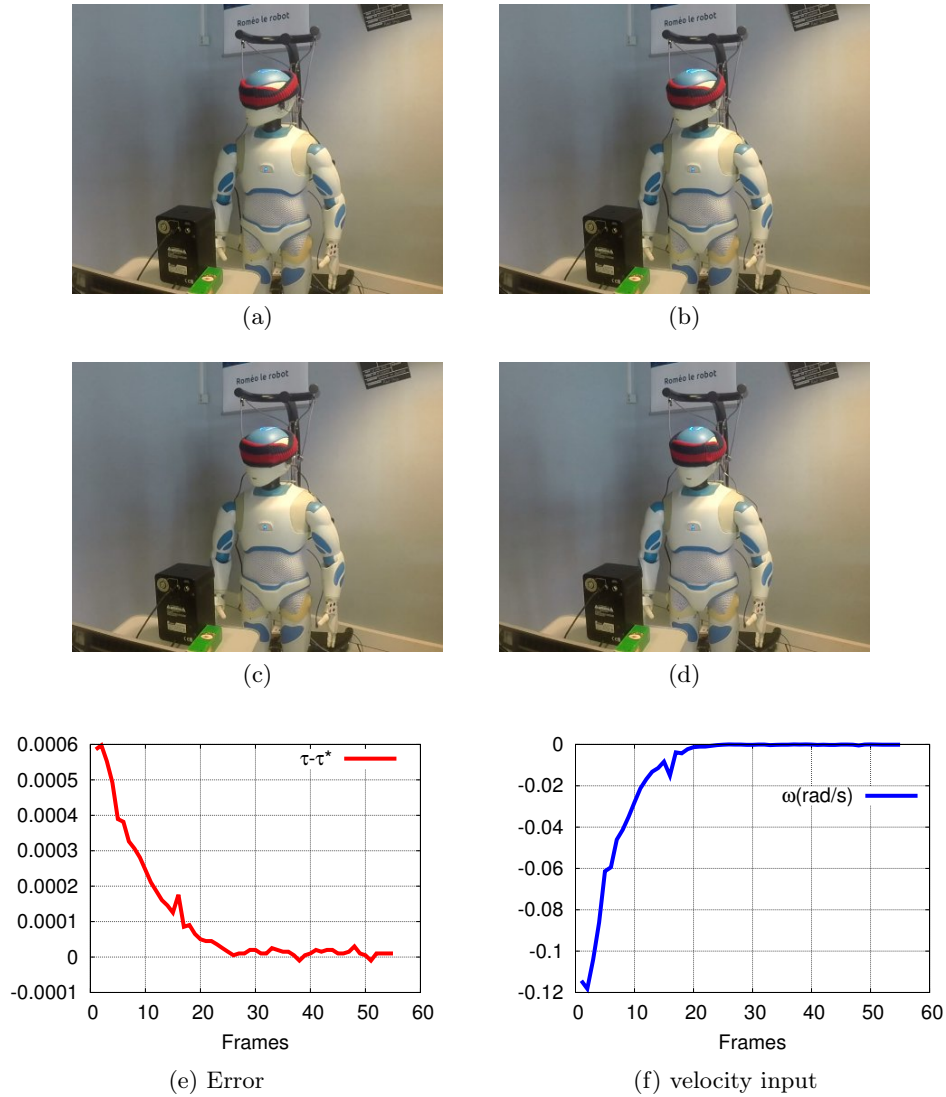


Figure 6.6: Head turning task from initial pose in (a) to final pose in (d) using ITD cues on *Romeo*.

The control scheme is based on the AEG model where the azimuth angles are directly related to ITDs. For head-mounted systems, this model is obviously erroneous when considering localization approaches, however, it is correct when considering the variation of ITDs. Hence, AS greatly simplifies auditory perception models, since AEG models can be used for various robotic platforms, for near-field and far-field distances, without any particular tuning.

### 6.2.2 Tracking a moving sound source

The control scheme is also able to track a moving sound source. In the experiment depicted in Figure 6.7, the loudspeaker is continuously emitting a speech signal. The task consists in maintaining a sound source in front of *Romeo* head. ITD cues are used in this experiment with (6.1), but the same kind of results can be obtained with ILDs.



Figure 6.7: Tracking a continuously moving sound source from (a) to a final pose in (d) using ITD cues on *Romeo*.

However, for ITD-based tasks, the loudspeaker is moved more slowly than for ILD-based tasks. The tracking step of ITDs estimation limits the loudspeaker motion speed. On the other hand, since ILD-based tasks do not require any tracking, the robot is able to deal with faster motions of the sound source.

This property is confirmed by the second set of experiment conducted on *Pepper*. In this case the robot has to deal with rapid changes of the source position. This experiment corresponds somehow to achieve a sequence of positioning tasks as depicted in Figure 6.8. Hence the task error follows a sequence of exponentially decreasing patterns.

These two experiments were correctly achieved but we particularly noticed the sound source directivity effect. This effect is more visible when considering ILD-based control. As already outlined in Chapter 4, the sound source directivity modifies admissible poses for completing a given task. More precisely, since sound waves

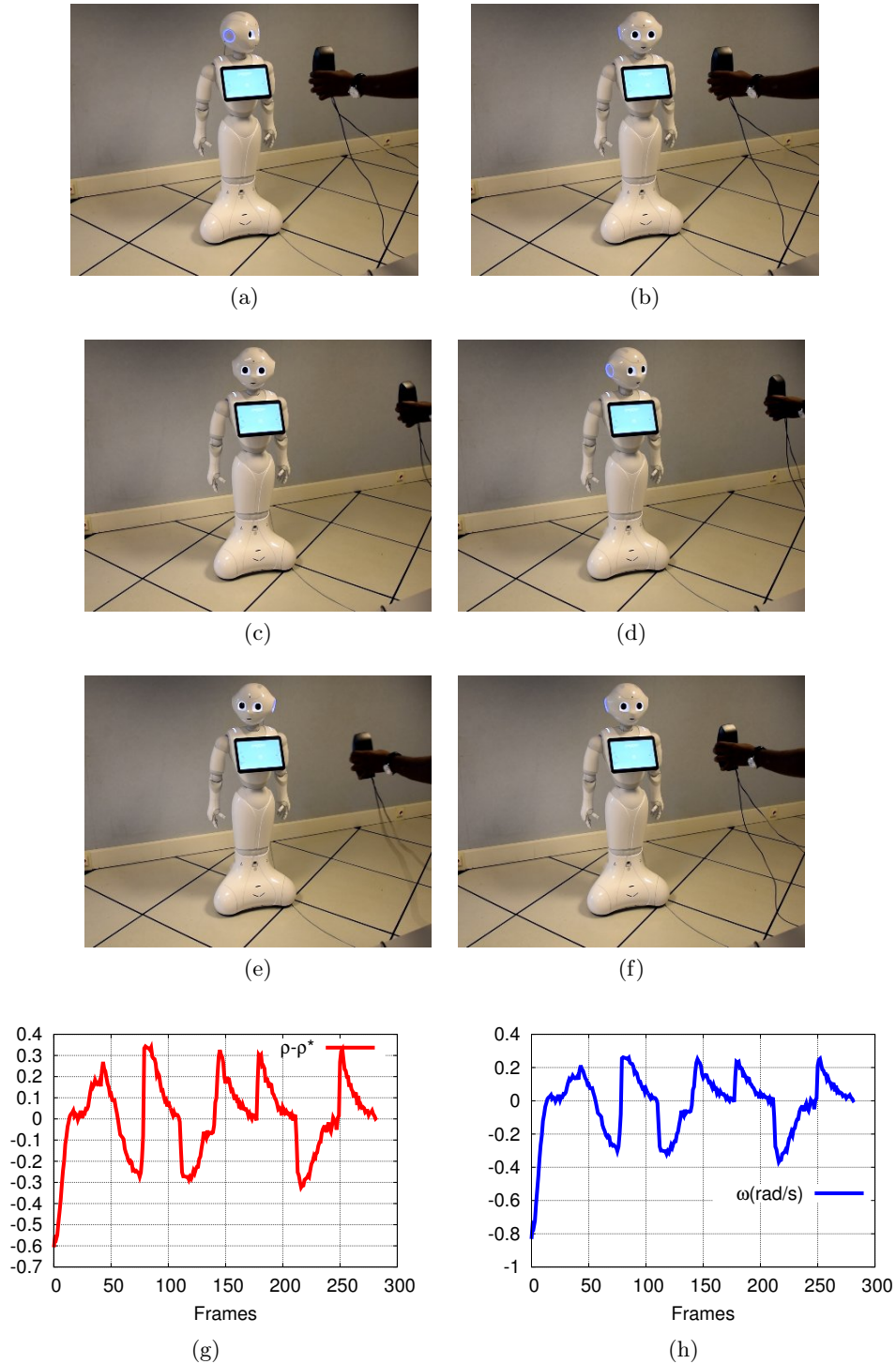


Figure 6.8: Experiment with rapidly changing source positions using ILDs that is solved as a sequence of conventional positioning tasks.

are not emitted uniformly in all directions, the task does not strictly aim to face the sound source. The task rather consists in facing the sound beam emitted by the loudspeaker (see Figure 6.9). The controller correctly converges towards an appropriate configuration but perceptually the robot pose seems inaccurate.

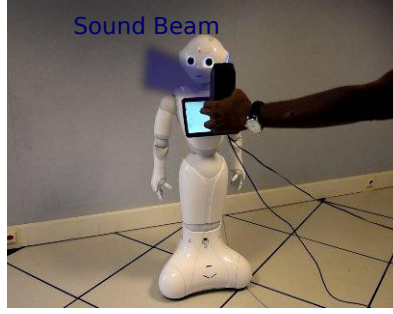


Figure 6.9: The effect of sound directivity: *Pepper* is facing the sound beam instead of the speaker

All along this manuscript, we mainly considered point sources uniformly emitting spherical waves, which is a commonly-admitted assumption in robot audition. In practice this assumption does not hold. All the sources emitting in realistic environment follow more complex radiation patterns governed by their intrinsic directivity property. A more thorough study is certainly required to include this notion in the control modeling.

### 6.2.3 Controlling *Pepper* with ILD and the energy level

For this last set of experiments, we aimed to apply the framework developed in Chapter 4.3, that considers ILDs and the energy level as input features. Here we are not attached to position the head of *Pepper*, but rather its holonomic base, that can be controlled with  $\mathbf{u}_M = (v_x, v_y, \omega_z)$ . In this case, the interaction matrix  $\widehat{\mathbf{L}}_{\rho E}$  combining the ILD  $\rho$  to the energy level  $E_M$  can be directly used to control the robot.  $\widehat{\mathbf{L}}_{\rho E}$  is given as:

$$\widehat{\mathbf{L}}_{\rho E} = \begin{bmatrix} \frac{2\widehat{x}_s(\rho-1)-d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{2\widehat{y}_s(\rho-1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} & \frac{\widehat{y}_s d(\rho+1)}{\widehat{\ell}^2 + \frac{d^2}{4} - d\widehat{x}_s} \\ \frac{2E_M \widehat{x}_s}{\widehat{\ell}^2} & \frac{2E_M \widehat{y}_s}{\widehat{\ell}^2} & 0 \end{bmatrix}, \quad (6.3)$$

while the control scheme related to the microphones velocity is given by

$$\mathbf{u}_M = -\lambda \widehat{\mathbf{L}}_{\rho E}^+ \mathbf{e}, \quad (6.4)$$

with  $\mathbf{e} = [\rho - \rho^*, E_M - E_M^*]^\top$ . In these experiments, the loudspeaker is emitting a white Gaussian noise. The task consists in approaching the loudspeaker from a random pose of the robot. For this purpose the desired ILD is set to  $\rho^* = 1$ .  $E_M^*$  is measured experimentally when the robot is at  $\ell \approx 0.3$  m from the sound source. The results are exposed in Figure 6.10. The robot is able to reach a pose satisfying

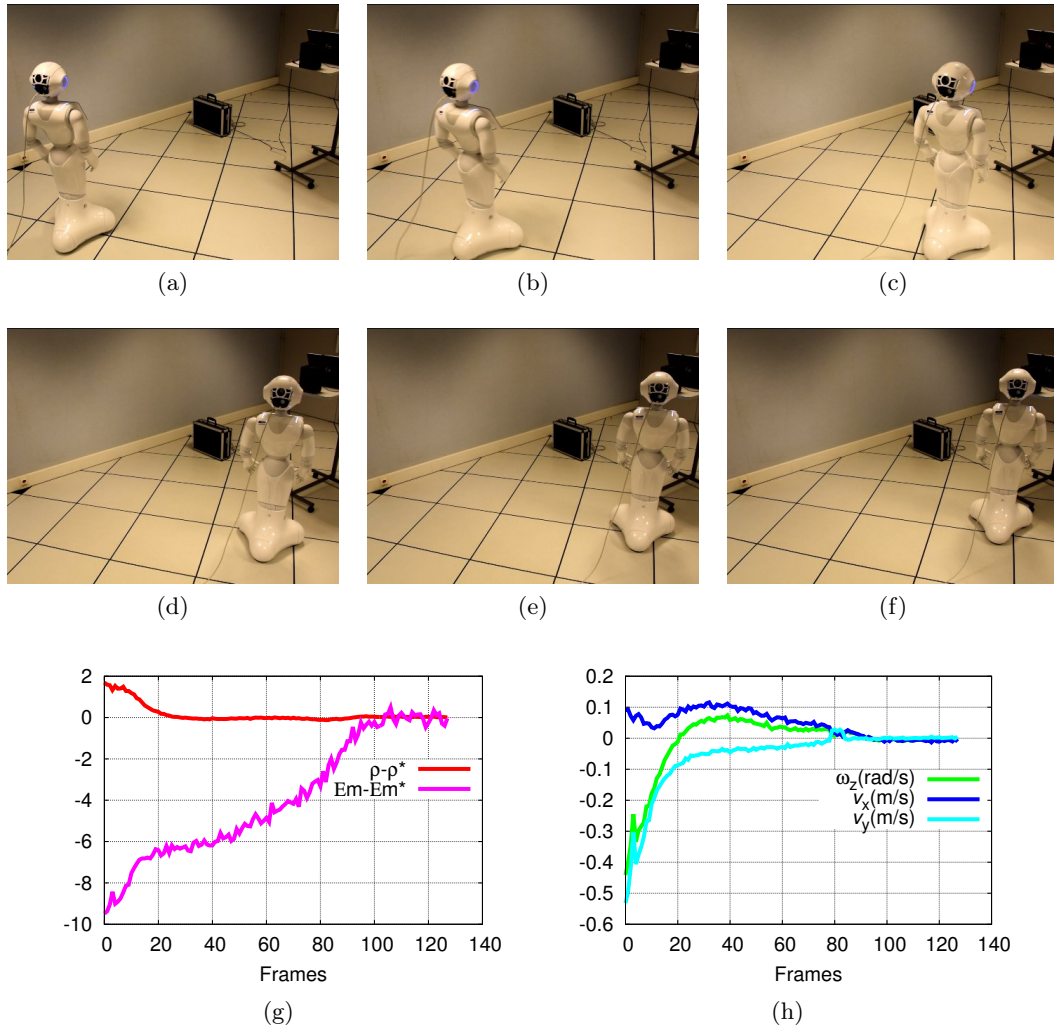


Figure 6.10: *Pepper* approaches the sound source and reaches a pose satisfying  $\rho = \rho^*$  and  $E_M = E_M^*$ .

the given auditory conditions. The behavior of the robot is, nonetheless, not very natural. The control scheme generates lateral motions that would be unsuitable for more realistic scenarios involving human-robot interaction for instance. Rotation motion would certainly give more naturalness to the approaching task. This behavior is due to the control that is uniquely applied to *Pepper* base. Thus, it should be interesting to control at the same time the head (rotation) and the base of *Pepper* to get a more natural and smooth motion. Despite this, from the latter application, the robot can also follow a moving sound source as illustrated in Figure 6.11.



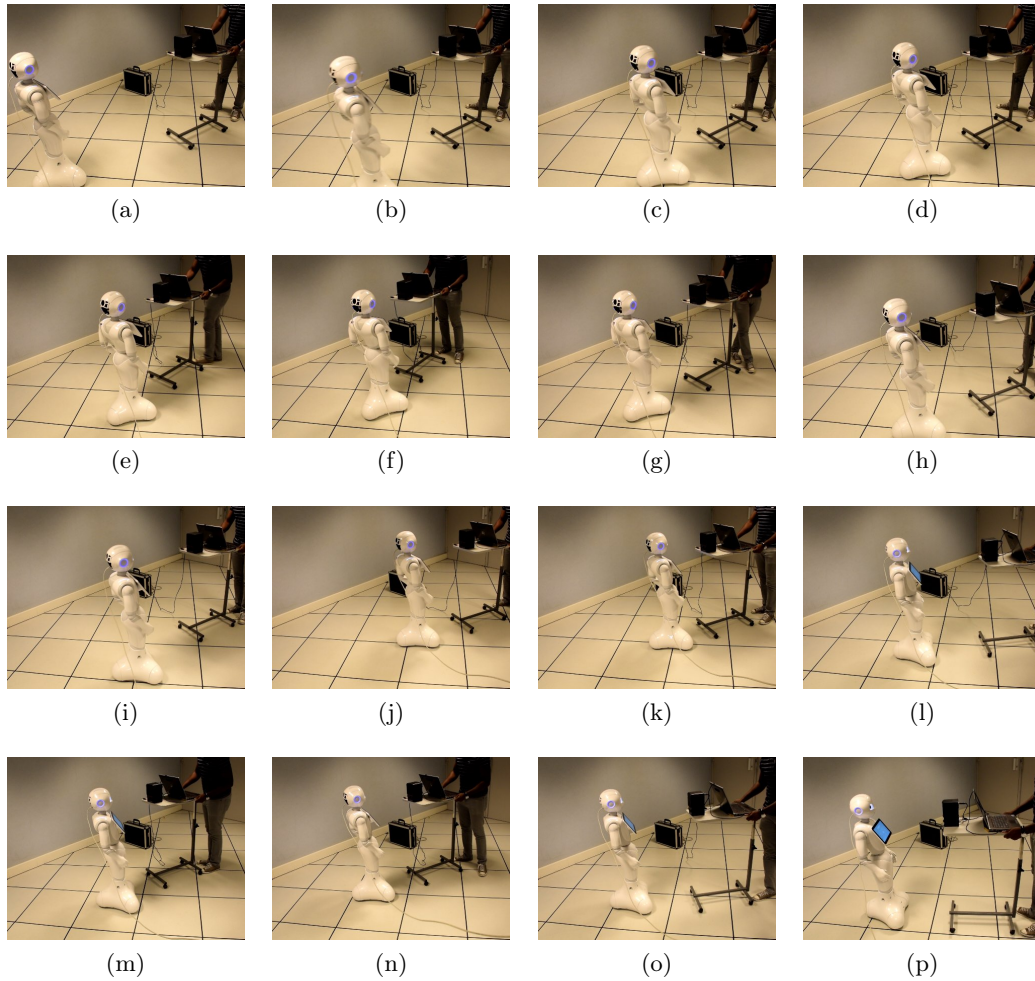


Figure 6.11: *Pepper* follows the sound source

### 6.3 Conclusion

This chapter confirmed the versatility of AS paradigm. This paradigm is mainly based on the variation of the selected auditory features. This property is robust and flexible to changes of acoustic conditions and of binaural configurations. To demonstrate such property, we showed, in a first phase, that the same variation governs ILD measurements for free-field anechoic conditions, and for head mounted systems in highly reverberant conditions. The same ILD shape could then be observed on two humanoid robots *Pepper*, *Romeo* and in simulated environments. From this result, it could be inferred that the frameworks developed in Chapters 4 and 5 considering free-field conditions can be applied on head mounted systems without any additional modeling.

Hence, in a second phase, we validated experimentally these frameworks on *Romeo* and *Pepper*. Despite the free-field modeling, convincing results could be

obtained on humanoid robots, in realistic environments. A first set of experiments were about controlling robots gaze from ILD and ITD cues. The head scattering effect did not affect the task accuracy independently of the selected feature. Thereafter, a full motion control based on ILD and energy features was demonstrated to work accurately on *Pepper*. In this experiment, the translation and rotation velocities were controlled to approach a sound source. For all these experiments, the case of a moving sound source was also successfully addressed.

As the final chapter of this thesis, the experimental results depicted above sum up all the potential benefits of applying AS for real world applications. AS can be defined as a straightforward and simple paradigm that allows to control robots despite approximations in the auditory scene modeling and/or robot structure, and adverse acoustic conditions. Of course only basic scenarios and tasks have been addressed in this chapter, but these results unveil solid foundations for more complex and realistic scenarios.

# Conclusion

## Conclusions

As a recent field of research, robot audition encompasses all topics related to the understanding of and interaction with an acoustic scene. Sound source localization among others concentrated most of the efforts of the robot audition community. Localization paradigms initiate most of the applications based on the sense of hearing. More specifically robot motion and behavior control are mainly based on localization outputs. However, unlike many other fields related to perception (vision, touch/range sensing), applications in real-world context remains rare. The main bottleneck lies in controlling robots from unreliable, inaccurate features or even sometimes from features that cannot be related to an accurate sound location. These features are highly fluctuating and depend principally on environment acoustics and robot structure. Several parameters may change the sound perception. Reverberation, in indoor environment, and noise substantially alter auditory features for localization. Moreover acoustic waves heading towards microphones are modeled differently in the far-field or in the near-field. In the far-field it is easier to model auditory features but on the other hand the effect of reverberation is more important. Alterations may also be generated by robots structure. The scattering effect induced by heads, when considering anthropomorphic robots, and ego-noises are parameters to be included in the "equation" of sound localization. Eventually, when considering dynamic scenes, where robots and/or the sound source of interest move, the degree of complexity of the localization process greatly increases. In view of all these parameters, motion control from sound localization in realistic and dynamic environments is a task that has not been achieved yet when considering binaural configurations as exposed in this thesis.

Hence, this thesis investigated the use of a sensor-based approach in order to control the motions of a robot. Sensor-based control corresponds to a feedback loop directly relating robot motion to low-level auditory features. Unlike positioning methods based on localization, here, the sensorimotor loop is closed and voluntary motions of the robot are driven by auditory measurements. The resulting paradigm, that we termed *aural servo*, allows positioning a robot with respect to given conditions related to measured auditory features. More interestingly this approach is performed without source localization. This latter characteristic certainly represents a key advantage of *aural servo*. Instead of focusing on the localization problem that is extremely complex to solve in realistic environments, our approach focuses primarily



on the variation of the measured auditory features. By discarding the localization concept to the positioning problem, *aural servo* is more robust to parameters related environments or robots structure.

In this thesis, we introduced more specifically control models based on the ITD, the ILD and the absolute level of energy. From these models built upon the variation of these features with respect to robot motion, various positioning tasks were performed. With both ILDs and ITDs, tasks that consist in controlling the orientation of a robot and/or approaching a sound source are performed under realistic conditions. More impressively, the benefits of these frameworks have been emphasized by solving issues related to the auditory perception. When considering ILDs, the front-back ambiguity is inherently solved by the control scheme. For ITDs, we demonstrated the robustness of the control scheme towards approximations in the distance of observation. Eventually, we proved that a range positioning using the absolute level of energy could be performed accurately under reverberant conditions. Furthermore these approaches are suitable for dynamic scenes. This problem is seldom addressed in robot audition. In this manuscript, from these approaches, we demonstrated the ability of our approach to control in real-time the orientation of a robot in order to face a moving sound source. It has also been demonstrated that a robot could accurately track a moving sound source under various changing acoustic conditions. More importantly, most of the concepts developed in this work are experimented on robots, in realistic acoustic environments. Besides stressing the suitability of *aural servo* for real environments, these experiments validate the versatility of the paradigm. The experiments have been carried, in the same way, on mobile robots equipped with free-field microphones or on humanoid robots in which microphones are integrated inside the head.

To the best of our knowledge, this work is one of the first attempt to characterize the motion of a robot with respect to auditory information, in an explicit control framework. This manuscript is also supported by a large set of experiments in real world conditions, which is a core contribution of this thesis. Beyond this work, we would like to envision this approach as a general framework that can be applied to any type of robots or acoustic conditions. However, as a pioneering work, we are aware that this work only scratches the potential benefits of sensor-based control in robot audition. The methods presented in this thesis could be enhanced in order to interact, navigate, and understand a given acoustic scene. In pursuance of this goal, this work should be complemented by the wide variety of methods developed for sound source localization, in order to handle intermittent sound sources, tracking and evaluating the number of active sources. Indeed, the path taken by this thesis is oriented towards the simplest and the most straightforward configurations that allows to control robot motion from auditory feedback in real world conditions. As a result, this choice certainly emphasizes the actual benefits of using *aural servo* but on the other hand the solutions proposed in this manuscript might not be optimal and should be polished for more adverse and challenging scenarios.

## Perspectives

### **Redefinition of the sound localization problem**

More than an alternative approach to sound localization for robot motion control, we think that our approach presages a redefinition of interactions based on auditory perception. Until now, the common approach consists in making the interaural cues robust, considering realistic environments and a particular auditory system. The localization process and the motion control stemmed from it solely depend on the accuracy of the extracted cues. This thesis showed that from inaccurate auditory cues, positioning tasks could be performed independently of the auditory system used. Hence, by adding *aural servo* in the mechanism of auditory interaction, the part devoted to localization is relaxed from accuracy constraints, since *aural servo* can cope with inaccurate measurements. Such mechanism resembles to human auditory system. As already explained in the introduction of this thesis, when using only interaural cues, humans are not so accurate in localizing sound compared to other species. However, by using head motions, visual cues or memory of previous auditory experiences, human auditory system reveals all its potential and is then fully exploited.

Similarly, here, we could imagine a global framework where auditory cues are extracted, tracked, identified and fed to an *aural servo* mechanism in order to interact with a given sound source. In this way, addressing the cocktail-party problem could be performed by combining *aural servo* with a source activity detection algorithm, methods for inferring the number of sound sources, robust tracking and an attention mechanism. These methods have already been studied and developed in signal processing/robot audition literature in the context of sound localization. Most of these methods could be used in synergy with the ILD-based or ITD-based positioning tasks.

### **A unified framework**

In this work, ILDs and ITDs have been processed separately. As a logic extension, combining ILDs and ITDs in an unified framework can be envisaged. As explained in the first chapter, human auditory system combines several auditory features in order to localize potential sound sources. In our context, although ILD-based methods and ITD-based methods are redundant with respect to the positioning task, these methods exhibit complementary characteristics. ILD-based methods are particularly adapted to single source tasks, and do not require any tracking. On the other hand, ITD-based methods can handle multi-source configurations and non stationary/intermittent signals such as speech. Nonetheless, when using ITDs, the bottleneck concerns the sound source identification and tracking. Combining these two features could then overcome the limitations of both methods. Multi-modal control and hybrid sensor-based control are certainly paths to explore in order to develop such a framework. By extension, we can then imagine using all auditory cues involved in sound localization, in a control scheme positioning a robot with respect to the azimuth, elevation and range information.

### Integration with other modalities

*Aural servo* could also be extended by using other sensory modalities such as vision. Interactions as performed by humans generally combine several senses. The sense of hearing can complement a limited field of view, while vision is useful to identify brief and/or intermittent sound sources. This could be exploited for robots. In this thesis, we considered sound signals long enough to allow converging towards a desired pose. This assumption it is not always realistic. A thorough study should then be conducted in order to define the optimal time to convergence of such approaches in comparison to the mean duration of a sound signal. In this case, features prediction or techniques such as visual servoing (when the source is visually identifiable) could be used if the controller has not converged at the end of the sound signal.

### Extension to other auditory cues

One of the benefit of our approach lies in the fact that we do not need to localize sound sources. This interesting property opens perspective about using auditory cues that are for now neglected since they cannot easily be linked to a source location. For instance, the direct-to-reverberant ratio could be exploited in this manner to position the robot at a desired range to a given source. Spectral notches, related to HRTFs and sound elevation could be also exploited in the same manner, so that the control scheme is not limited to the azimuth plane. In the same vein, dense methods could be developed to directly interpret sound signals and skip the stage of auditory cues estimation. This approach could be really useful in the case of ITDs by allowing to skip any tracking or identification step. An explicit method consists in deriving the robot motion directly from the cross-correlation function, so that these tracking and identification steps are avoided. Such approach has already been used with vision through mutual information [DM11] or photometric moments [BCM13].

During this thesis several paths and ideas of applications have been explored. Among them, a navigation system based on echoes was an interesting application that could not be fully developed, principally for a lack of time. In this configuration, echoes are used as input features of a sensor-based control framework. From these echoes used as a range sensing, a robot is able to position itself with respect to walls similarly to bats.

### Extension of the field of applications

As underlined in the multi-sources simulations in Chapter 5, bearing-only navigation can be performed with sound and more particularly ITDs. Unfortunately, this method could not be fully experimented during this thesis. Applications of this method in the field of multi-robot control or autonomous navigation could be particularly interesting to expand the field of robot audition. We do believe that this kind of approaches could be implemented without a lot of effort. Moreover, these approaches allow connecting robot audition to fields with a wider exposure. This connectivity could be beneficial for robot audition by increasing the awareness of the robotic community about hearing sense and the ensued problematic.

Furthermore *aural servo* is not limited to binaural configurations. Array-based configurations could also be considered in a dedicated control framework. Such application could be a consistent alternative to usual localization techniques that are principally based on the fusion of information. This fusion step does not appear

## Conclusion

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in *aural servo* paradigm. Hence, better processing time should be obtained with similar flexibility and robustness as the tasks described in this manuscript.

However, on the path of extending our approach, more evaluations are required. The development of appropriate simulation tools is essential. Currently most acoustic simulated environments are designed for punctual and static situations. These tools are particularly slow when simulating high levels of reverberation. As a result, most of the simulations conducted in this manuscript are based on moderate reverberation, since a higher level of reverberation was not affordable in term of processing time. A real-time simulation tool would be mandatory to process deeper analysis of control frameworks.



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## Resume

Cette thèse s'intéresse au développement de lois de commande basées sur la perception auditive. Dans le domaine de l'audition robotique, le contrôle du robot à partir d'informations auditives est généralement basé sur des approches de localisation de source sonore. Cependant, la localisation de source en conditions réelles est une tâche complexe à résoudre. En environnement intérieur, les perturbations causées par le bruit, la réverbération ou même la structure du robot peuvent altérer le processus de localisation. Cette tâche de localisation devient encore plus complexe si la source et/ou le robot sont en mouvement. Aujourd'hui, en se restreignant aux systèmes binauraux, la localisation sonore en environnement réel n'est pas encore réalisable de manière robuste. À l'opposé, nous développons dans cette thèse une commande référencée capteurs, l'*asservissement sonore*, qui ne nécessite pas de localiser la source. Le mouvement du robot est directement relié à la perception auditive : une tâche de positionnement est réalisée par une boucle de commande, où le mouvement du robot est régi par la variation d'indices sonores de bas niveau. Les résultats expérimentaux dans différentes conditions acoustiques et sur différentes plates-formes robotiques confirment la pertinence de cette approche en conditions réelles.

**Mots-clés:** Robotique, Traitement du son, Asservissement sonore, Commande référencée capteur, Audition robotique, Différences interaurales.

## Abstract

This thesis concerns the development of a control framework based on auditory perception. In general, in robot audition, the motion control of a robot using the sense of hearing is based on sound source localization approaches. However, sound source localization under realistic conditions is a significant challenge to solve. In indoor environment perturbations caused by noise, reverberation or even the structure of the robot may alter the localization process. When considering dynamic scenes where the robot and/or the sound source might move, the degree of complexity of source localization raises to a higher level. As a result, sound source localization considering binaural setup is not achievable yet in real-world environments. By contrast, we develop in this thesis a sensor-based control approach, *aural servo*, that does not require to localize sound sources. The robot motion is straightly connected to the auditory perception: a positioning task is performed through a feedback loop where the motion of a robot is governed by the variation of low-level auditory features. Experimental results in various acoustic conditions and robotic platforms confirm the relevance of this approach for real-world environments.

**Keywords:** Robotics, Audio signal processing, Aural servo, Sensor-based control, Robot audition, Interaural differences.